Wire-Cell TPC Responses, Simulation, Signal Processing and Implications for Vertical Drift Designs

Brett Viren and Andrea Scarpelli

January 21, 2021

Topics

- Concepts in responses for simulation and signal processing.
- Requirements and progression of LArTPC detector response models.
- Review of performance of Wire-Cell Toolkit's implementations for wire detectors.
- Challenges for responses for **strips+holes** anodes and performance of Wire-Cell Toolkit with 50-L detector data.

\mathcal{R} : modeling LArTPC ionization response

For $x \in "det$ " (real detector) or "*sim*" (detector simulation):

- S_x drifted ionization charge distribution ("signal"), \mathcal{R}_x detector response in anode to drifting electrons, \mathcal{N}_x non-signal related "noise",
- \mathcal{M}_x a **measurement** (eg ADC waveforms on channels).

Simulation is a **convolution** (with \mathcal{R}_{det} or \mathcal{R}_{sim}):

$$\mathcal{M}_x = \mathcal{N}_x + \mathcal{R}_x \circledast \mathcal{S}_x$$

Signal processing is (mostly) a **deconvolution** with \mathcal{R}_{sp} to get **reconstructed signal**:

$$\mathcal{S}'_x = F_{sp} \circledast \mathcal{R}_{sp}^{-1} \circledast \mathcal{M}_x$$

(more details in backups)

Requirements on responses

We then must **choose** \mathcal{R}_{sim} and \mathcal{R}_{sp} and wish to minimize the **per-event difference** between reconstructed and "true" ionization signal in the *sim*:

$$|\mathcal{S}'_{sim} - \mathcal{S}_{sim}|$$

and simultaneously minimize an **ensemble difference** between reconstructed signal over **similar event samples** from *det* and *sim*:

$$|\langle \mathcal{S}'_{det}
angle - \langle \mathcal{S}'_{sim}
angle|$$

This obviously implies we want:

$$\mathcal{R}_{det} pprox \mathcal{R}_{sim} \sim \mathcal{R}_{sp}$$

IOW, we want \mathcal{R}_{sim} as close to reality as computational power allows and \mathcal{R}_{sp} as close to reality tempered by our limited basis of measurement \mathcal{M} (ie, channel-level info).

Historical Progress of Response Sophistication

$$\mathcal{R}^{1D} \to \mathcal{R}^{2D} \to \mathcal{R}^{2.5D} \xrightarrow{?} \mathcal{R}^{3D}$$

LArSoft \rightarrow Wire-Cell Toolkit (with wires) \rightarrow WCT (with strips+holes) \rightarrow ???

1D response model: $\mathcal{R}^{1D}(x)$

- Response depends on 1D coordinate (drift direction).
 - Sim and SP assume current only in wire nearest to drifting electron.

Pros:

- Computationally fast and algorithmically easy.
- Still available for use in LArSoft SP and sim.
- For sim, strongly minimizes: $|\mathcal{S}_{\textit{sim}}' \mathcal{S}_{\textit{sim}}| pprox 0$

Cons:

- For sim, the $|S'_{sim} S_{sim}| \approx 0$ minimum is "too perfect".
- For data, avg reco signal differences $|\langle S'_{det} \rangle \langle S'_{sim} \rangle|$ are large.
 - The "long range induction" effects can not be ignored.
 - MicroBooNE data demonstrated this model is too simplistic.

Wire-Cell 2D response model: $\mathcal{R}^{2D}(x, \rho)$

- Response depends on 2D coords (drift + pitch directions).
 - May induce current on range of nearby wires (1 ± 10) or strips (1 ± 5) .
 - \mathcal{R}_{sim} varies w/in one *wire region* (10 sub-pitch bins).
 - Average over each wire region: $\langle \mathcal{R}_{sim}^{2D}(x,\rho) \rangle |_{\rho} \rightarrow \mathcal{R}_{sp}^{2D}(x)$

Pros:

- Well minimizes both $|S'_{sim} S_{sim}|$ and $|\langle S'_{det} \rangle \langle S'_{sim} \rangle|$.
- Validated, optimized implementation in Wire-Cell Toolkit.
 - Now established as default in most LArSoft uses.
- On average, works well on some non-2D geometries (eg wires).

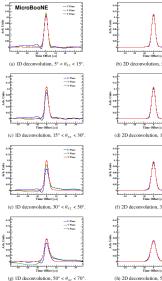
Cons:

- Field response calculations more difficult than 1D, but reasonable.
- Sim and sigproc algorithms more complex, somewhat slower than 1D.
- The more 3D the geometry \Rightarrow more imperfect is a 2D model.
 - Particularly, strips + holes stress the model.

1D SP vs Wire-Cell 2D SP on MicroBooNE det data

- V Film

-Y Plane



(b) 2D deconvolution, $5^\circ < \theta_{xz} < 15^\circ$ <u>a yan yan yan ya</u> Time Offset [as] (d) 2D deconvolution, $15^\circ < \theta_{re} < 30^\circ$ (f) 2D deconvolution, $30^\circ < \theta_{xx} < 50^\circ$. (h) 2D deconvolution, $50^\circ < \theta_{vz} < 70^\circ$.

Plotted: average reconstructed ionization signal $\mathcal{S}_{dat}^{\prime 1D}$ and $\mathcal{S}_{dat}^{\prime 2D}$ vs sample time for ensemble of tracks in four angle bins.

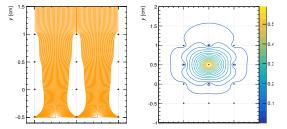
2D Wire-Cell signal processing is able to correctly recover identical average track reco signal $S_{det}^{'2D}$ independently from each wire plane.

Exploit LArTPC technology for tomography!

Ionization Electron Signal Processing in Single Phsae LArTPCs II. Data/ Simulation Comparison and Performance in MicroBooNE MicroBooNE Collaboration arXiv:1804.02583, JINST 13, P07007 (2018).

2D Field Response Calculations - Wires

- 2D slice across 3D geometry
 plus some fictional alignment!
- Wires are infinite, parallel and uniform, and there are no edge effects.
- Drift paths in applied \vec{E} -field.
- Per-conductor weighting field.
- 126 drift paths per plane: 1 ± 10 wires, 6 "impact positions" per wire at ¹/₁₀ th pitch, exploit translation and mirror symmetries.

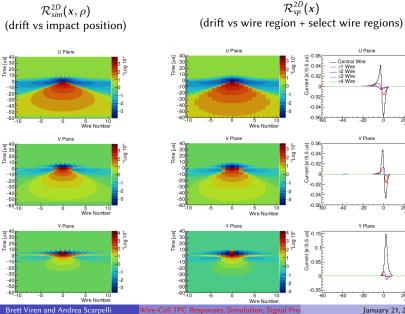


 2D (x, ρ) drift paths and example U-wire weighting field for ProtoDUNE-SP calculated by drifires/Garfield++ (we are migrating from the venerable GARFIELD).

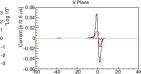
Instantaneous induced current: $I(t_i) = q \vec{W}(\vec{r}_i) \cdot \vec{v}(\vec{r}_i)$; $\vec{r}_i = \vec{r}(t_i)$; $t_i = t_0 + i\Delta t$

- W: weighting field is E-field with conductor-of-interest at 1V, all else at 0V.
- v: drift velocity along path r calculated from LAr physics and solving for applied E-field.
- q: an infinitesimal element from the distribution of drifted ionization electrons at the start of a drift path.

Wire-Cell 2D response for MicroBooNE



Time [µs]



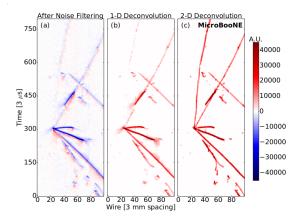
40 Time [µs] January 21, 2021

20

Time [us]

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More Wire-Cell NF/SP performance on MicroBooNE





- (a) WCT noise filtered,
- (b) 1D SP and
- (c) Wire-Cell 2D SP

Similar performance on ProtoDUNE-SP.

Ionization Electron Signal Processing in Single Phase LArTPCs I. Algorithm Description and Quantitative Evaluation with MicroBooNE Simulation MicroBooNE Collaboration arXiv:1802.08709, JINST 13, P07006 (2018).

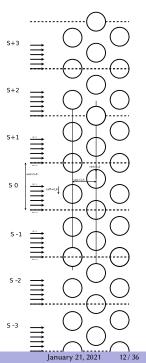
Challenges of strips+holes for $\mathcal R$

- Strips+holes strongly violate 2D model
 - non-parallel strips between the planes
 - (a "feature" shared with wires)
 - non-uniform along their length.

Extra challenges:

- ind/col hole patterns differ between strips-in-plane and strips-across-planes
- hole-pattern has some finite repetition distance
 - ▶ 50-L has a 2-hole repetition,
 - longer repetition for some 3-view designs.
- Small bonus: fields drop faster with strips than wires
 - 1 ± 10 wires $\rightarrow 1 \pm 5$ strips.

50-L detector collection stips+holes \longrightarrow



Wire-Cell TPC Responses, Simulation, Signal Pro

2.5D model for strips+holes $\mathcal{R}^{2.5D} \rightarrow \mathcal{R}^{2D}_{sim,sp}$

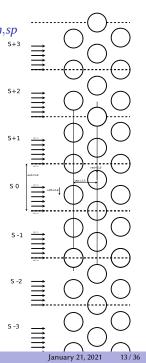
"2.5D" trick

- Construct **slices** across strips spanning the *repetition distance*.
- Each slice defines one \mathcal{R}^{2D} problem domain.
- Calculate per-slice R^{2D}_{slice},
- Take **average** over slices to get $\mathcal{R}^{2D}_{sim,sp}$
- New problems for calculating \mathcal{R}^{2D}_{slice} :
 - How best to define and combine slices?
 - How many slices are needed?
 - How to exploit symmetry to reduce calculation?
 - How wrong is this on average and in detail?
 - tests ongoing

Chosen slices shown as vertical lines \longrightarrow

(details how we use GARFIELD to perform calculation in backups)

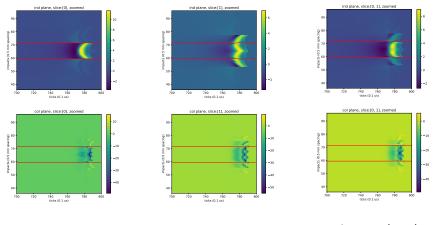
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50-L test detector: $\mathcal{R}^{2.5D} \rightarrow \mathcal{R}^{2D}_{sim}$

The PCBro (PCB anode readout) package processes GARFIELD output to form the per-slice responses to produce WCT-compatible $\mathcal{R}_{sim,sp}^{2D}$ as linear color scale.

(induction plane, slices 0, 1 and average) (collection plane, slices 0, 1 and average)



 Slice 0.
 Slice 1.
 Slices 0.5 * (0 + 1).

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 Wire-Cell TPC Responses. Simulation. Signal Pro
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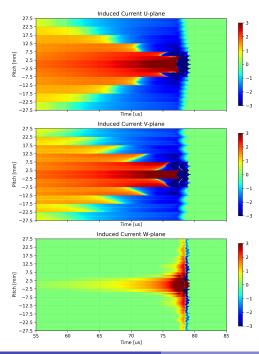
Next 3 slides

tl;dr: focus on bottom two plots.

To better view tails we use " $\pm \log 10$ " scale for 50-L's \mathcal{R}_{sim}^{2D} . Each slide shows a slice-average or a specific slice:

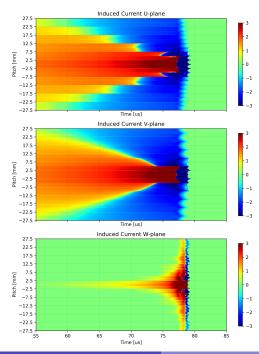
- highlight slice 0
- a highlight slice 1
- average over both slices

50-L detector only has 2-views: induction + collection. The "U" and "V" labels indicate different forms for the **induction** plane info. The "W" is always the **collection** plane info.



Slice 0 50-L \mathcal{R}_{sim}^{2D}

- U is average over both induction slices
- V is induction slice 0
- **③** W is collection slice 0

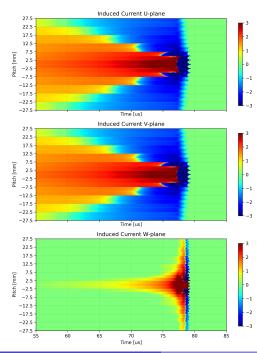


Slice 1

50-L \mathcal{R}^{2D}_{sim}

- U is average over both induction slices
- V is induction slice 1

③ W is collection slice 1

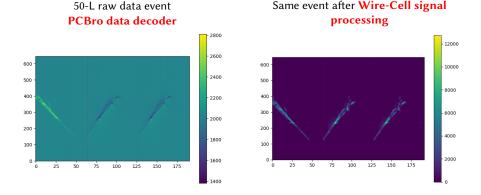


Average

- 50-L \mathcal{R}^{2D}_{sim}
 - U is average over the two induction slices.
 - V is average over the two induction slices.
 - (the two are identical)
 - W is average over the two collection slices.

50-L raw data and WCT sigproc with $\mathcal{R}_{sp}^{2.5D}$

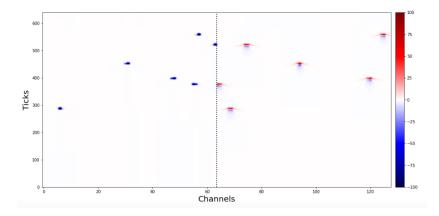
The PCBro package also provides a 50-L raw data decoder and hooks to run **Wire-Cell Toolkit signal processing** on 50-L data.



Note, the "double induction" plane is merely duplicate to fit nominal WCT assumptions of 3 planes. The PCBro package uses this "extra" plane to test different field responses in the same job. In production processing, we need not waste CPU on the duplication.

50-L WCT simulation of ³⁹Ar "blips"

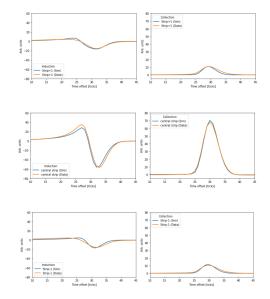
Andrea Scarpelli collection plane | induction plane



Samples ³⁹Ar energy spectrum scanned from arXiv:1705.05726v1.

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50-L data / WCT sim comparison - ³⁹Ar "blips"



Andrea Scarpelli and Serhan Tufanli

average raw waveforms

50-L data (orange) and Wire-Cell 2D sim (blue) using a 2-slice $\mathcal{R}^{2.5D}$. Separated by induction (left) and collection (right) planes and for central strip (middle) and central ± 1 strips (top/bottom).

3-view strips+holes

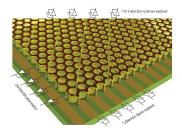
At least 3 views are needed to well exploit the tomographic power of LArTPC!

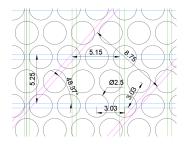
Various design, installation and **response** challenges for a 3-view **strips+holes** detector.

One possible design adds "diagonal" strip with "skewed" hole pattern. \longrightarrow

- \mathcal{R} may require N > 2 slice average: "2.7D ... 2.9D"
 - Have scheme to produce data-driven optimal weighting of slices. Non-trivial, but there if we need it.
- Strip angle is such that no slice goes only through hole diameters.
 - Further biasing of the "2.x D" approach?
- Even with 3D, prefer to maximize regularity of patterns.

Design wish: 3-view isosceles or hexagonal strip+hole pattern. A novel hexagonal design by Bo looks nice!





From Bo's presentation at the recent DUNE collab call

Summary and some next steps

- Wire-Cell's use of $\mathcal{R}^{2D}_{sim,sp}$ improves over \mathcal{R}^{1D} as demonstrated with MicroBooNE and ProtoDUNE.
- The $\mathcal{R}^{2.5D}$ trick brings Wire-Cell support to 50-L strips+holes det.
 - Eyeball SP event display and average raw waveform tests look okay.
 - These average test metrics may hide some variational problems.
- Unclear (yet) "2.5D" trick is enough esp. for 3-view designs.
 - Require WCT SP/sim and test detector data to confirm.
- The more regular the strip+hole pattern the better!
 - isosceles/hexagonal 3-plane designs for vert. drift det, please!
 - a very new design already in this direction!
- **Precision tests** to check for correct **variations** *vs* position even more important for strips+holes.

In general: we will continue to improve support in Wire-Cell Toolkit for strip+hole test detectors and for the eventual DUNE VD module(s)!

 \mathcal{FIN}

backups

Conceptual LArTPC Simulation

Real detector (and its simulation) produces an event via a convolution of:

S an ionization charge distribution ("signal") with \mathcal{R} a detector response to drifting electrons, plus \mathcal{N} all non-signal related "noise", producing \mathcal{M} a measurement (eg ADC waveforms on channels).

$$\mathcal{M}_x = \mathcal{N}_x + \mathcal{R}_x \circledast \mathcal{S}_x$$

With x ="*det*" (real detector) or "*sim*" (detector simulation).

knowns $\mathcal{M}_{det}, \mathcal{M}_{sim}, \mathcal{S}_{sim}$ and \mathcal{N}_{sim} (modeled), \mathcal{R}_{sim} (but imperfect). unknowns $\mathcal{R}_{det}, \mathcal{S}_{det}$ and \mathcal{N}_{det} .

Conceptual LArTPC Noise Filtering

Noise filtering is a transformation F_{nf} on the measurement:

$$\mathcal{M}_x \to \mathcal{M}'_x = F_{nf}(\mathcal{M}_x)$$

designed to strongly remove **excess** or **external** noise and potentially reduce **inherent noise** leaving residual noise *n_x*:

$$F_{nf}(\mathcal{N}_x) \to n_x \ll \mathcal{N}_x$$

while attempting to leave the signal term approximately invariant:

$$F_{nf}(\mathcal{R}_x \circledast \mathcal{S}_x) \approx \mathcal{R}_x \circledast \mathcal{S}_x$$

Conceptual LArTPC Signal Processing

Signal processing attempts to recover a good approximation of S_x from \mathcal{M}_x .

It uses a **deconvolution** of the measure with a response and with added filters F_{sp} to further suppress the residual noise term.

$$\mathcal{S}'_{x} = \mathit{F_{sp}} \circledast \mathcal{R}_{sp}^{-1} \circledast \mathcal{M}_{x}, x \in \{sim, det\}$$

We may not use the detailed \mathcal{R}_{sim} here as it is in terms of the detailed, inaccessible "true" signal coordinates so we use an **average** \mathcal{R}_{sp} (ie, per-channel vs sample time matching \mathcal{M}_x).

Note: for induction channels, \mathcal{R}_{sp}^{-1} diverges at DC, thus **amplifies residual**, low-frequency noise. To counter, F_{sp} includes special algorithmic high-pass "filters" called *signal region of interest* (signal-ROI) and *local baseline correction*.

Conceptual LArTPC Responses

We then must **choose** \mathcal{R}_{sim} and \mathcal{R}_{sp} and wish to minimize the **per-event difference** between SP and "true" ionization signal in the *sim*:

$$|\mathcal{S}'_{sim} - \mathcal{S}_{sim}|$$

and simultaneously minimize an **ensemble difference** between SP signal over similar samples from *det* and *sim*:

$$|\langle \mathcal{S}'_{det}
angle - \langle \mathcal{S}'_{sim}
angle|$$

This obviously implies we want:

$$\mathcal{R}_{det} pprox \mathcal{R}_{sim} \sim \mathcal{R}_{sp}$$

IOW, we want \mathcal{R}_{sim} as close to reality as computational power allows and \mathcal{R}_{sp} as close to reality tempered by our limited basis of measurement \mathcal{M} (ie, channel-level info).

The Test Metric Caveat

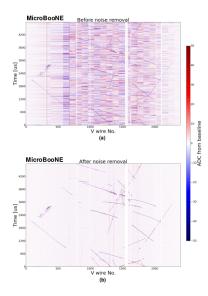
So far we use metrics sensitive to response averages over space.

- Demonstrates WCT \mathcal{R}^{2D} good on average for SP(*sim*) \approx SP(*det*).
 - Does control for 3D direction, important variation for SP efficiency!
- May not be sensitive to **imperfect detailed variations** in $\mathcal{R}^{2D}_{sim,sp}$
 - Eg, is there $SP(sim) \neq SP(det)$ bias/resolution at specific locations?
 - Particularly strip+holes have large variations along strip direction.
 * (will show)
- Examples of more **precise metrics** to apply in future *det* vs *sim*:
 - Signal matching between planes with ³⁹Ar or other "blips".
 - Detailed comparison of dE/dX with tracks from full 3D reco.

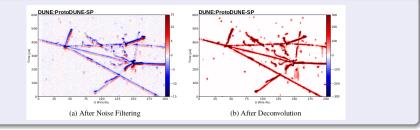
More Wire-Cell performance on MicroBooNE: NF

Wire-Cell software **noise filter** applied to MicroBooNE data event.

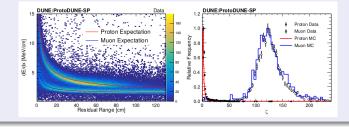
Noise Characterization and Filtering in the MicroBooNE Liquid Argon TPC MicroBooNE Collaboration arXiv:1705.07341, JINST 12 P08003 (2017).



Wire-Cell signal processing on ProtoDUNE-SP data



Wire-Cell simulation/data comparison dE/dX for μ and P on PDSP

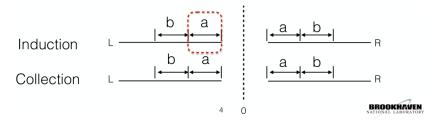


First results on ProtoDUNE-SP liquid argon time projection chamber performance from a beam test at the CERN Neutrino Platform. arXiv:2007.06722, JINST 15 (2020) P12004.

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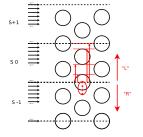
The 2.5D trick applied to GARFIELD



GARFIELD setup for 2D single-hole (Yichen Li)

Clever work-around to limitations of 2D and GARFIELD:

- Single-hole, 2D geometry with strips constructed as array of hundreds of "micro wire" sensing conductors
 - the L/R and A/B blocks work around some GARFIELD technical limits.
- Manual labor intensive post processing
 - Catalog maps of drift paths on slice to single-hole geometry.
 - Catalog micro wire selection criteria for each drift path.
 - ► Longer the hole pattern repetition distance ⇒ more the effort.



50-L data/sim checks t.b.d.

Other sources:

- 1 MeV e- Bismuth source "blips".
- MIP tracks in different direction bins.

Address "Test Metric Caveat" with more precise det vs sim comparisons:

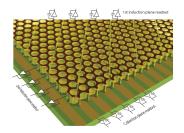
- SP ind/col ratio for "blips", ideally = 1.0
- Invariant values (eg, raw ind waveform integral = 0.0)
- Raw waveforms from "blips" as $f(\rho)$?

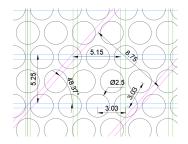
Skewed hole pattern: 2.5D? 2.7D?

50-L: **average** tests look good for $\mathcal{R}_{sim}^{2.5D}$, **eyeballing** SP event display looks good for $\mathcal{R}_{sp}^{2.5D}$. But, 50-L has fairly **regular hole pattern**. **New worry**: 3-view's "diagonal" strip with "skewed" hole pattern. \rightarrow

- May need *N* > 2 slice average: "2.7D ... 2.9D"
 - Have scheme to produce data-driven optimal weighting of slices. Non-trivial, but there if we need it.
- Strip angle such that **no slice** goes only through hole diameters.
 - Further biasing of the "2.x D" approach?
- Even with 3D, prefer to maximize regularity of patterns.

... **Design wish**: rectangular 2-plane or isosceles/hexagonal 3-plane!





From Bo's presentation at the recent DUNE collab call

So, why not \mathcal{R}^{3D} ?

Some work in 3D exists.

- Mostly for "near by" fields or "far fields" but for wires.
- Far field 2D calculation takes minutes-hours, 3D calculations take hours-days (using BEM, effectively impossible with FEM).
- Strips+holes need finer meshing ⇒ more processing (than wires).
- Longer hole repetition distance \Rightarrow more processing (than 50-L).

And, given \mathcal{R}^{3D} sim must contend with an explosion of data.

- *R_{sim}* drift paths per plane (some estimate/guesses)
 - 2D wires: 126.
 - 2.5D strips: O(100) for 50L, O(1,000) for skewed hole pattern
 - ★ but at least results in "standard" R^{2D}_{sim,sp}!
 - 3D: O(50,000), and worse: old simulation must be thrown out.

3D simulation

- Same concepts as 2D sim, but need all new algorithms/code.
- 2D sim exploits 10-way interlacing across common, wire-relative impact positions in order to use 2D FFT + 10-way sum for fast convolution.
 - How to even apply this trick in 3D? More variety along the strip direction will increase interlacing ⇒ harder/slower calculation.



From Bo's talk again



LARF

