

Cryo DC-DC boost converter development

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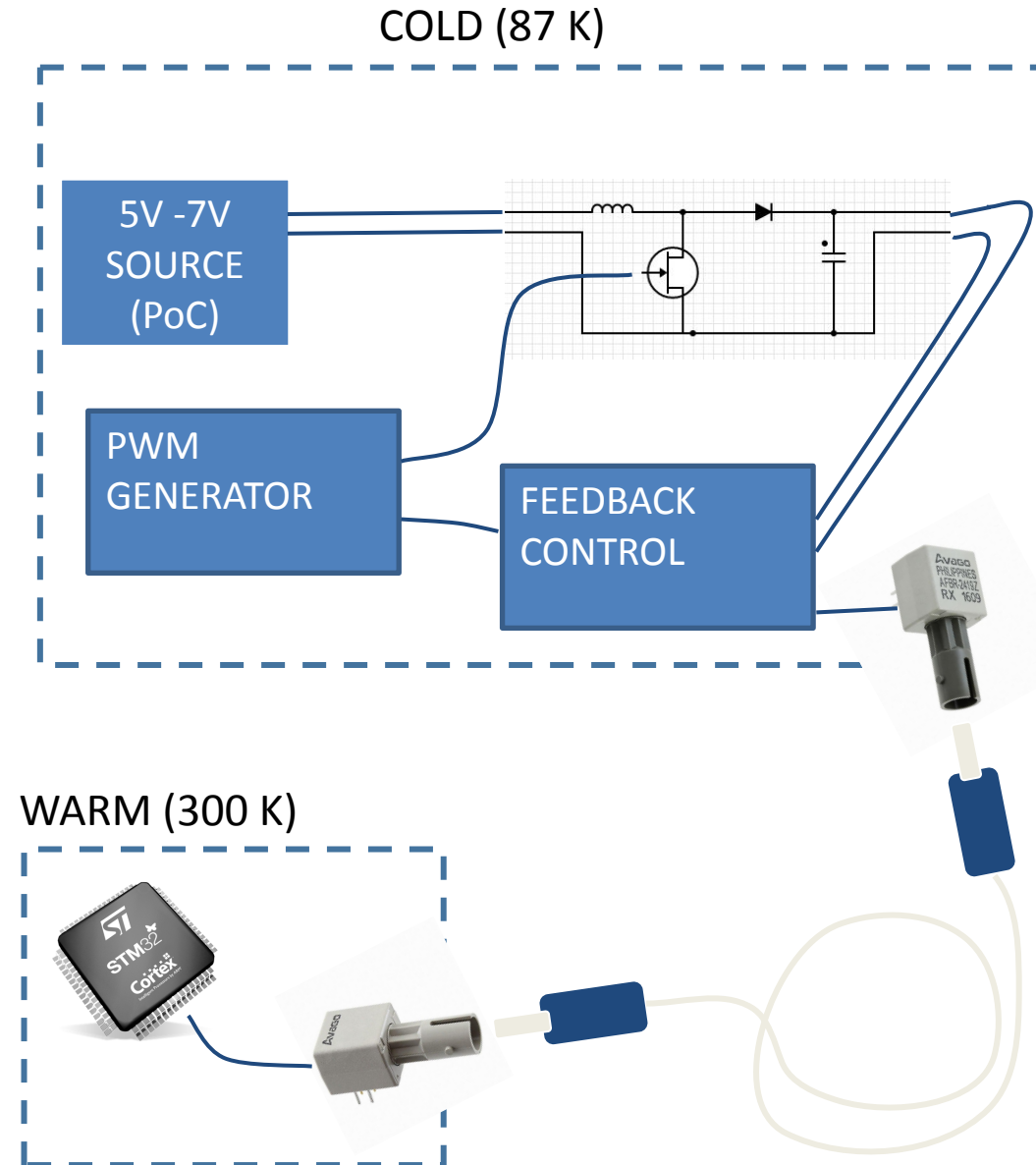


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- GaAs Power optical Converter (PoC)
- DC-DC boost design
 - Component characterization
 - Performance
- Control
 - Design
 - Component qualification
- Conclusions

System overview

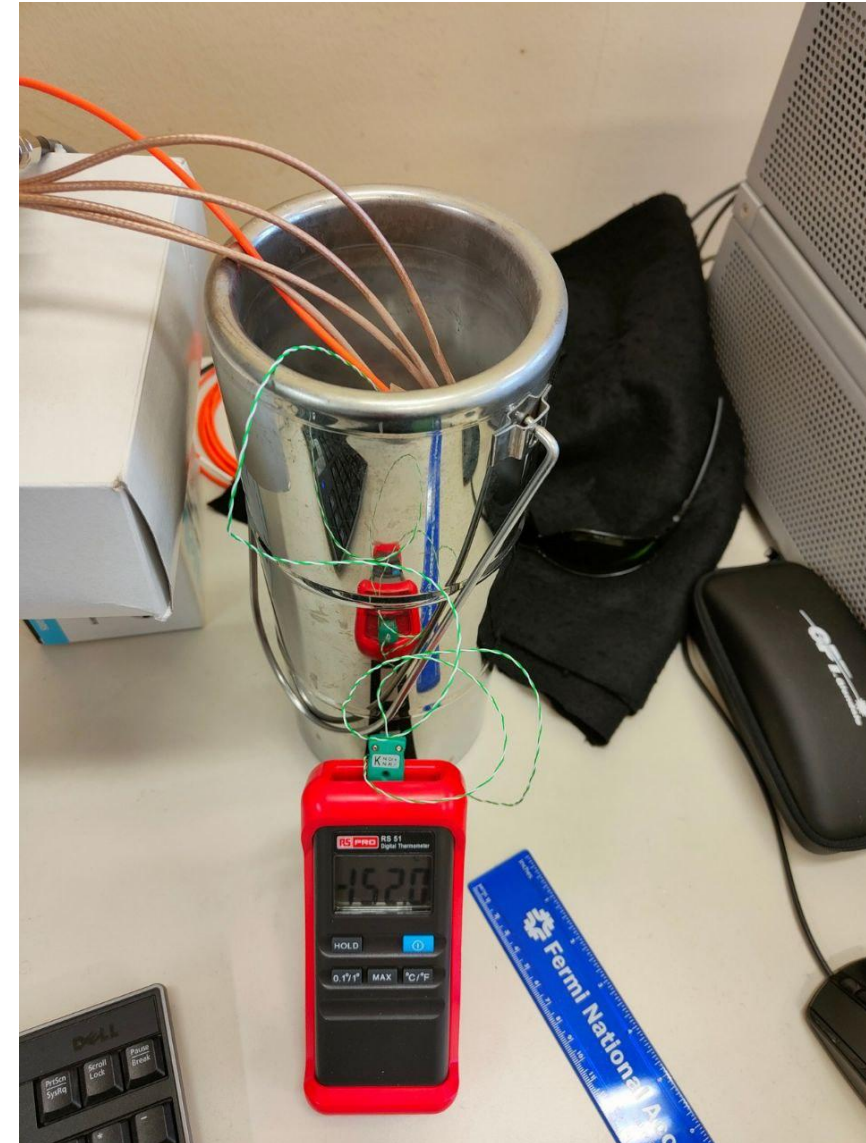
- The system evolved with the needs / possibilities in the DUNE collaboration
- Produce the HV for SiPM bias at cold by the means of a DC-DC boost converter
- A **power supply provided by a PoC** (Power optical Converter) with a 5V - 7V range will give voltage/current for the DC-DC converter
- **A typical boost configuration with a MOS** transistor, to increase the output voltage up to 50V
- A **PWM generator drives the MOS** switch with two possible controls:
 - Inner feedback setting the output voltage at a nominal point (e.g. 48 V)
 - Optical input to change the setting voltage within few volts



GaAs PoC

In order to power the DC-DC system a PoC is used.

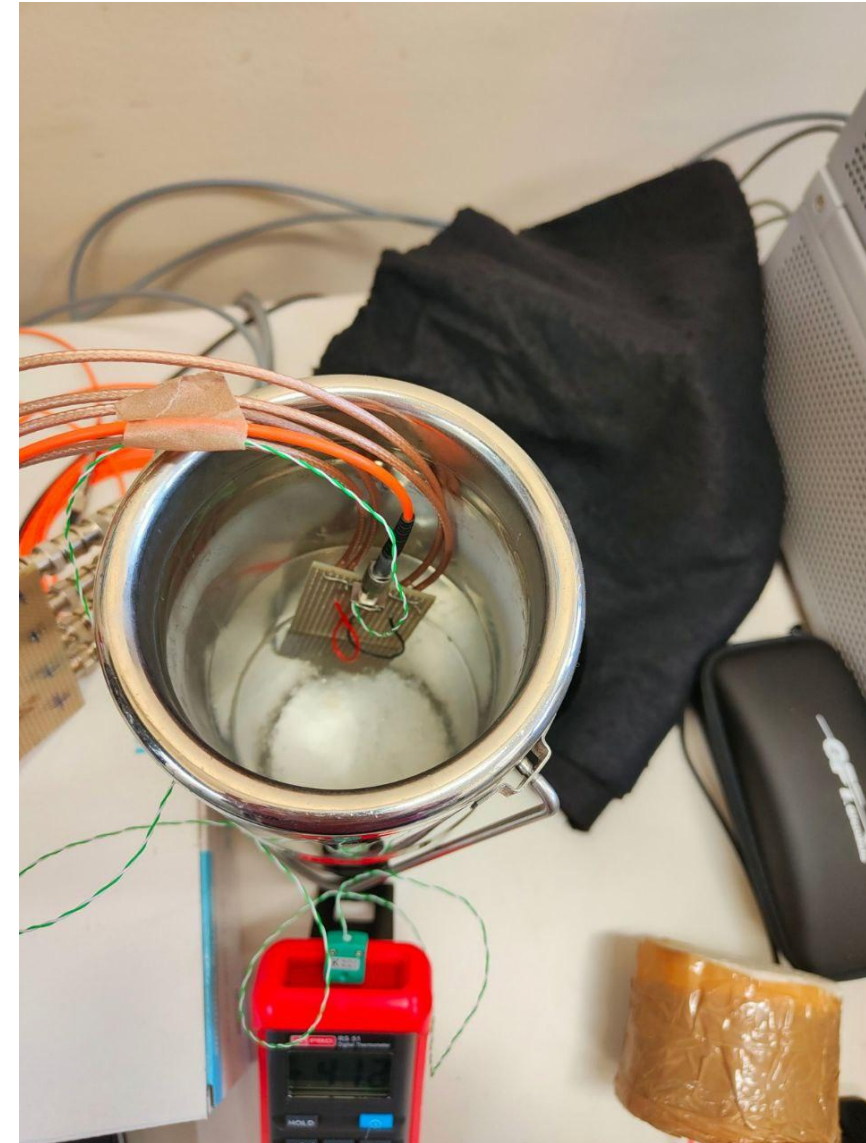
- We tested a **Broadcom AFBR-406L** based on GaAs technology
- Powered through a **808 nm Lumics laser** (8W max)
- We used a **multimode 200 μm core fiber** connected to the pig-tailed laser with an SMA/FC-PC connector, and to the PoC on the other side
- A **B1505A Semiconductor Analyzer** is used to fix the output voltage and **measure the current**
- The PoC is slowly cooled down checking its temperature with an RTD sensor.



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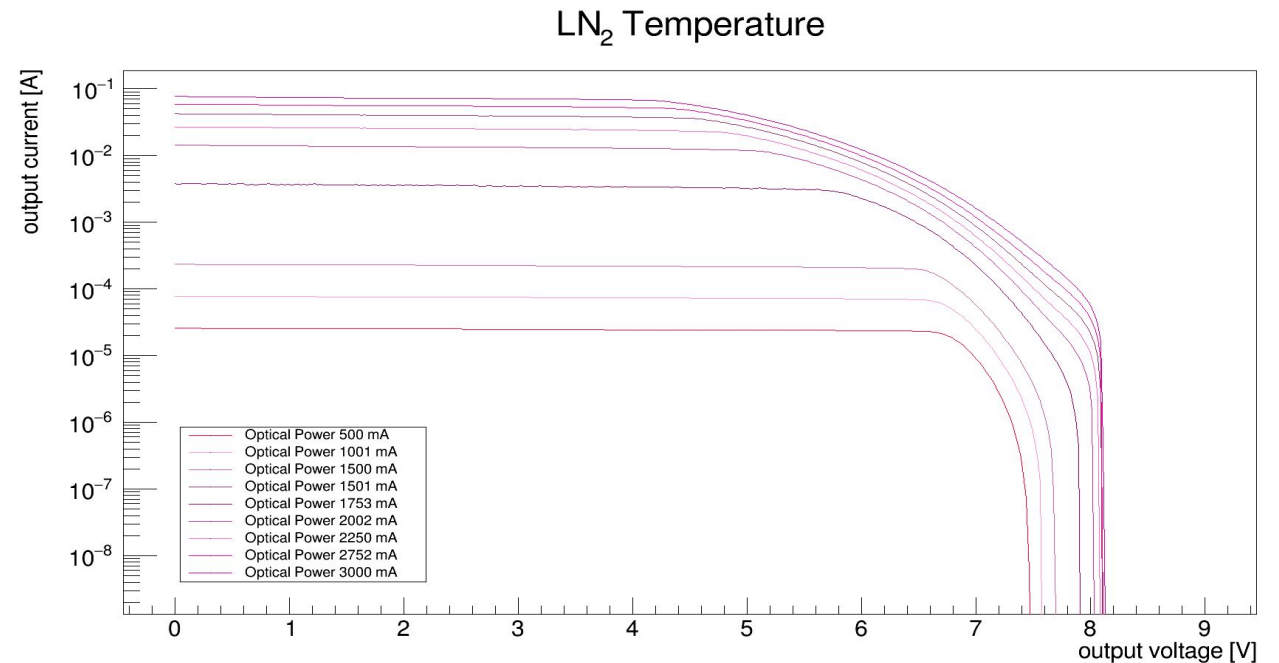
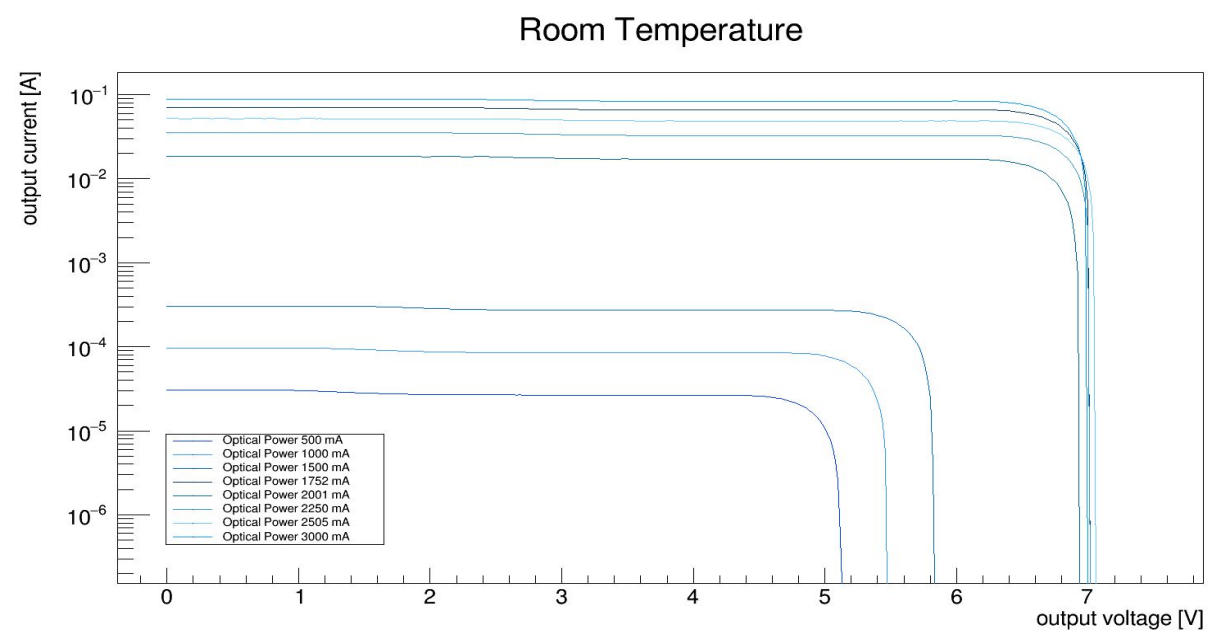
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GaAs PoC

In order to power the DC-DC system a PoC is used.

- The **curves are labelled** according to the **laser input current** ($P_{opt} \propto I$)
- As reference $3A \rightarrow 1.5 W_{opt}$
- Measurements are take both at RoomT (temperature not fixed but monitored during measurement) and at LN_2 temperature
- A slightly increase in the “open” voltage is observed going **from 7 V to 8V** for 3000 mA between Room and LN_2 temperature



DC-DC boost converter

We selected and characterized many components to check their operability at cryogenic temperature, their characteristics and to select the most suitable.

- MOSFET: *NTF3055L108T1G*, **ZXMN10A07FTA**, VN2460N8, 2N7002H6327XTSA2, 2N7002CK
 - *Lowest on resistance*
- Switching Diode: **BAV16W**, BAS16LD
 - *Very similar device, slightly better max current*
- Inductor: B82442T1125K000 (680 μ H), **B8244T1106K050 (10 mH)**
 - *Good to mitigate ripple*
- Capacitor: **C1812C104J1GACTU (COG 100 nF)**
 - *COG for cryogenic use*

DC characterization of MOSFET and diode

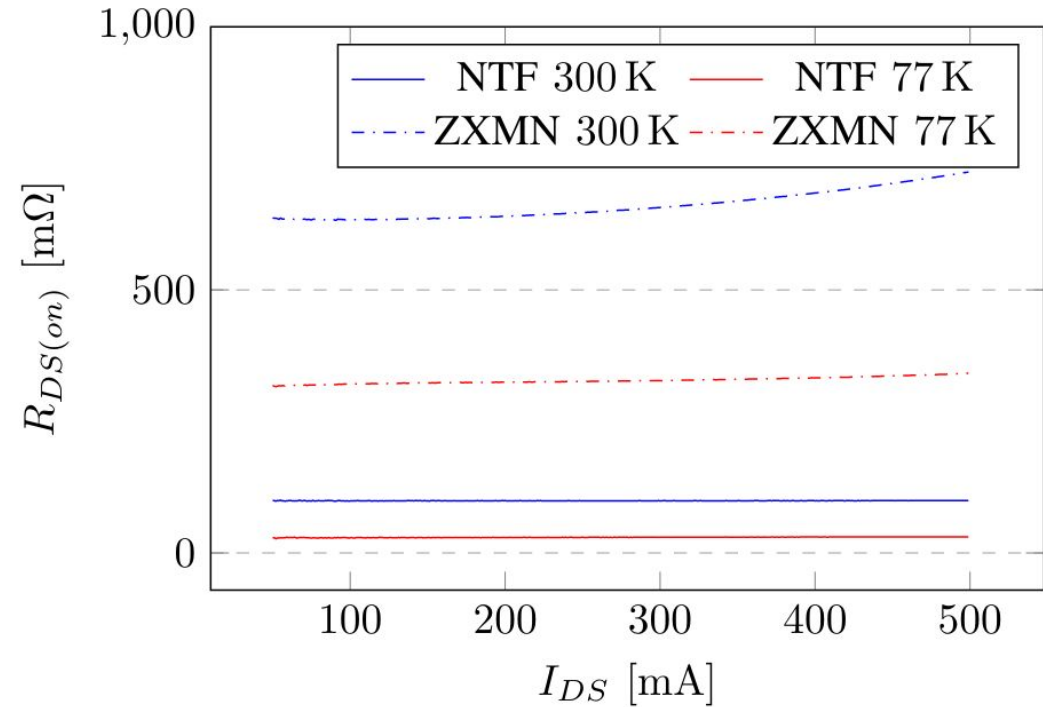
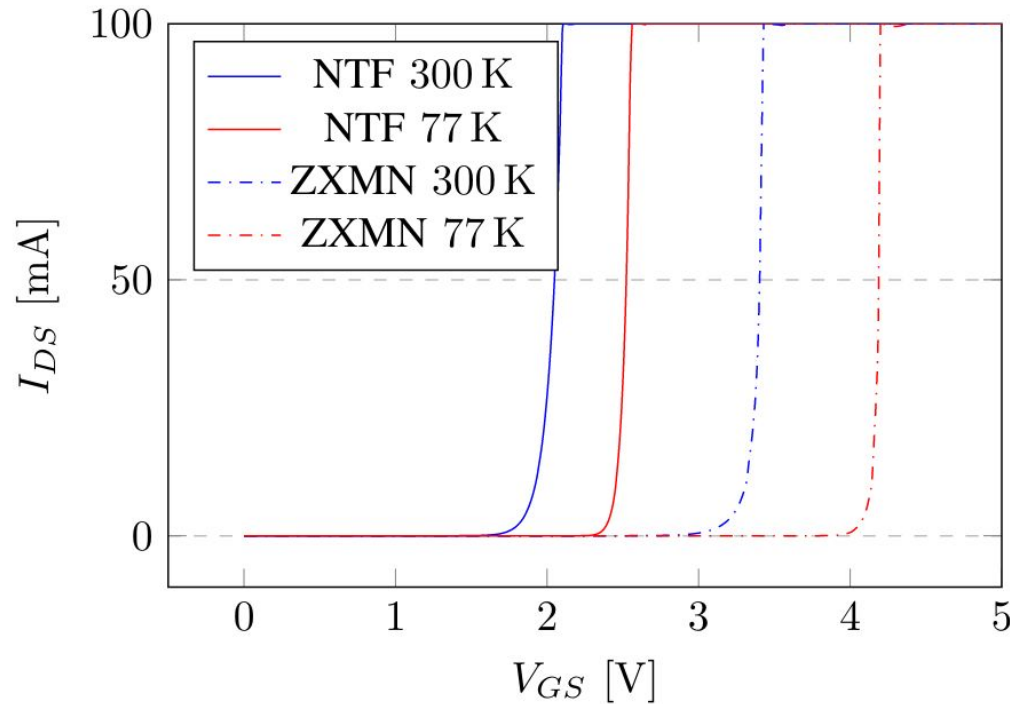
All **MOSFETs** have been characterized at Room Temperature and a LN2 temperature, with a B1505A Semiconductor Analyzer:

- I_D vs V_{GS} (at various V_{DS}) to measure the **threshold**
- V_{DS} vs I_D (at various V_{GS}) to measure the **on resistance** $R_{DS(on)}$

We will just show results for a couple of them, the most suitable for the application.

For the **diode** only the forward I-V curve was measured at both Room Temperature and LN2 temperature to measure the **junction voltage** V_D .

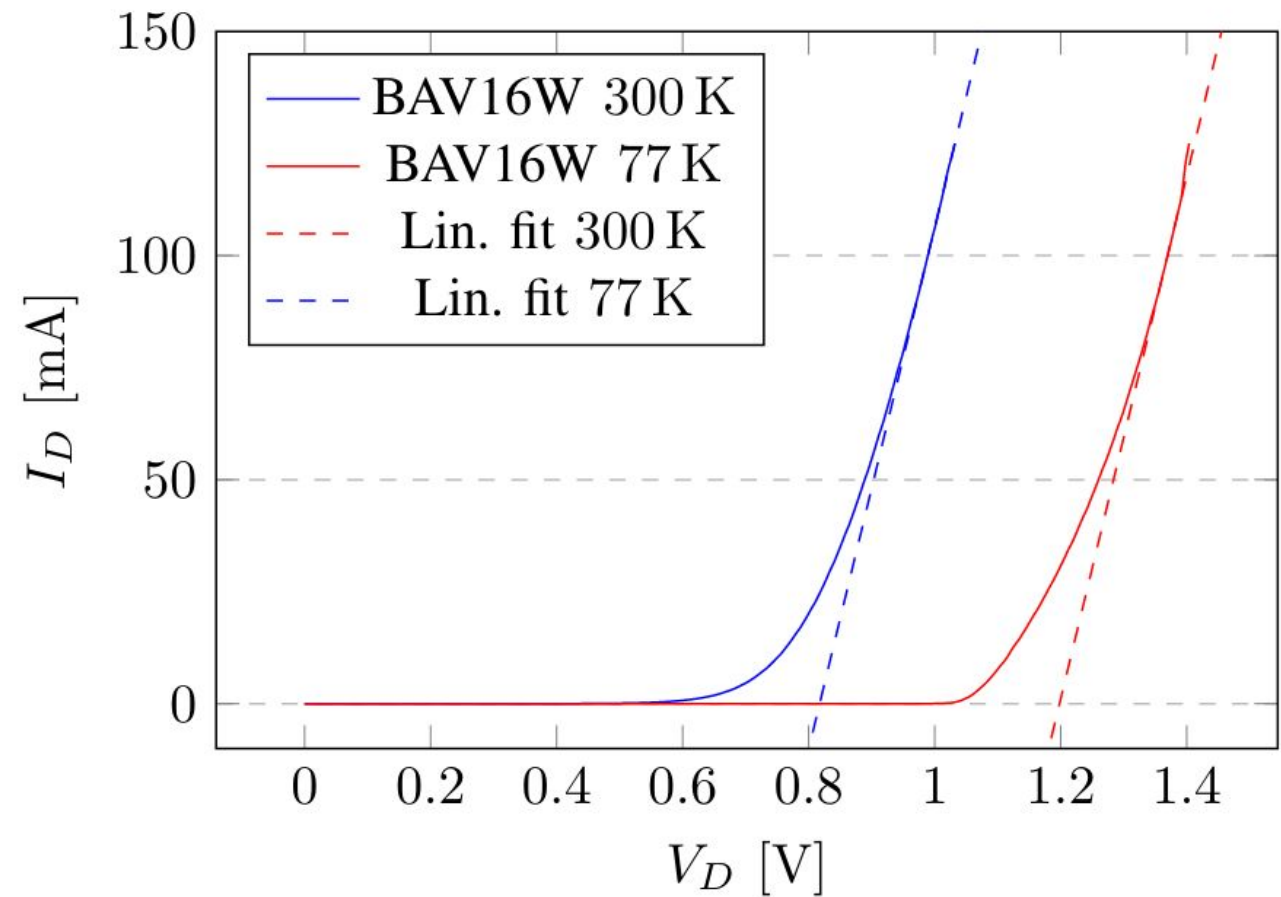
MOSFET results



- Drain current vs Gate-to-Source voltage: at 77 K the **threshold increases** of about **0.6 V**
- **On resistance reduces** considerably at 77 K, for NTF resulting in 30 m Ω .

Diode results

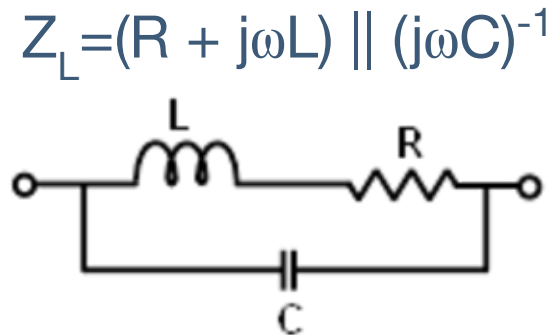
- Diode current vs applied voltage in forward bias mode. The **junction voltage increases** of about **0.4 V**
- The two selected diodes model exhibit the same behavior.



AC characterization of inductor

- The **10 mH inductor** has been tested as it is the most suitable for the application
- An **HP 4395A Impedance Analyzer** has been used in Reflection mode to analyze the impedance as a function of frequency

Model:



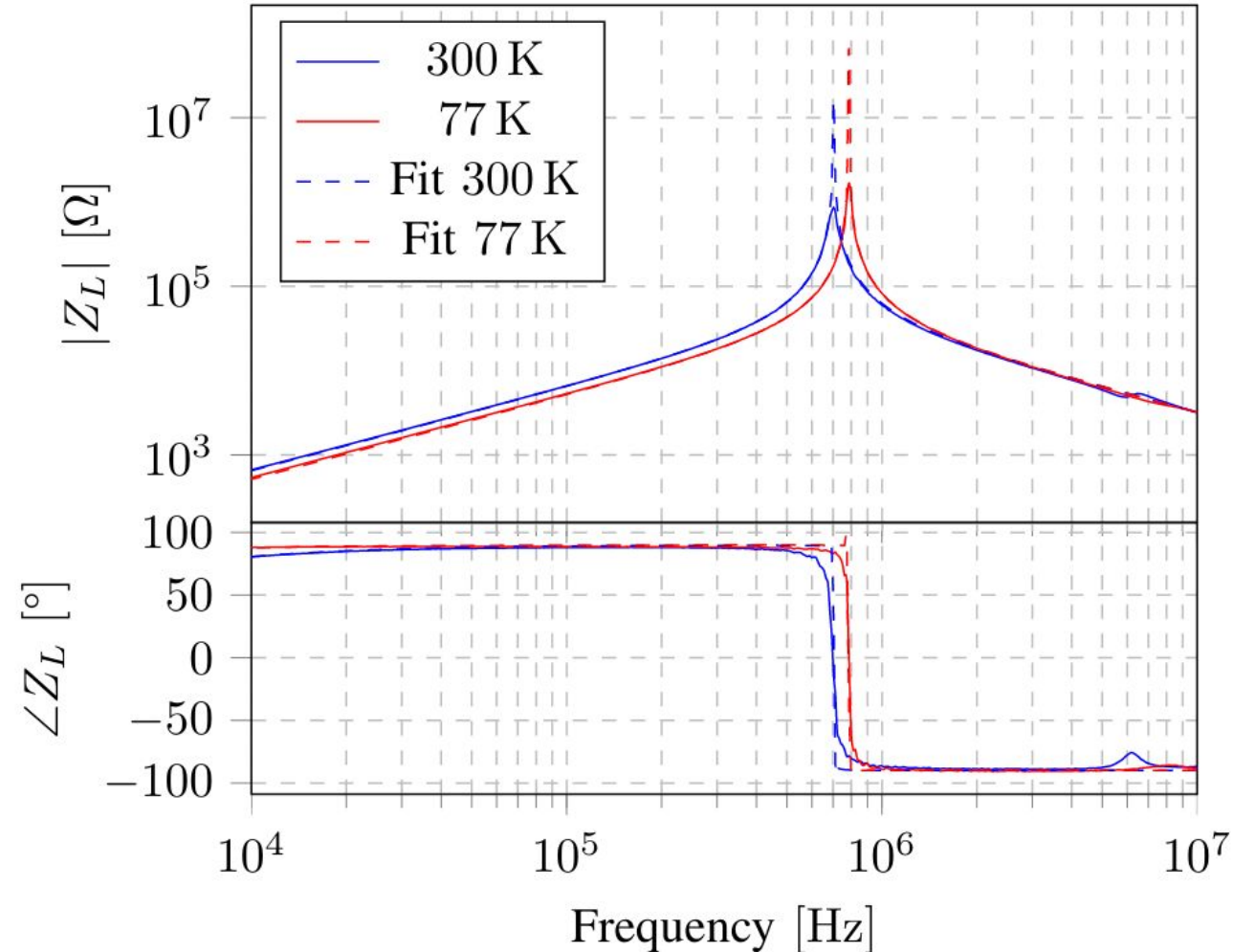
Results are fitted and parameter extracted

Inductor results

- Fitted parameters are:

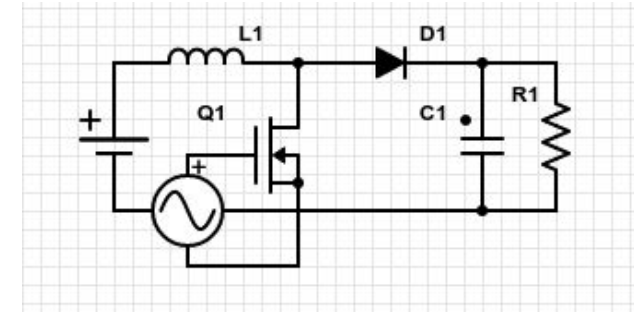
T [K]	R [Ω]	L [mH]	C [pF]
300 K	106	10	5
77 K	20	8	5

- Resonance frequency is 700 kHz (**800 kHz**) at 300 K (**77 K**), thus $f_{sw} = 100$ kHz is a good value
- Series resistance drops drastically at 77 K, giving much better performance to the DC-DC converter



DC-DC boost prototype test bench

- A matrix board is equipped with L1, D1, R1, C1:
 - **Load** is a **10 k Ω** resistor
- The Q1 (**NTF**) **transistor** can be changed to test all models
- **DC input** provided by a **linear supply** (AimTTi PL303QMD-P)
- The input current is monitored with a multimeter (HP 971A)
- The **control signal** is produced by a **Pattern Generator** (HP HP 81104A), High-level = 5 V, Low-Level = 0 V and rise/fall time = 3 ns with **100 kHz of period**.

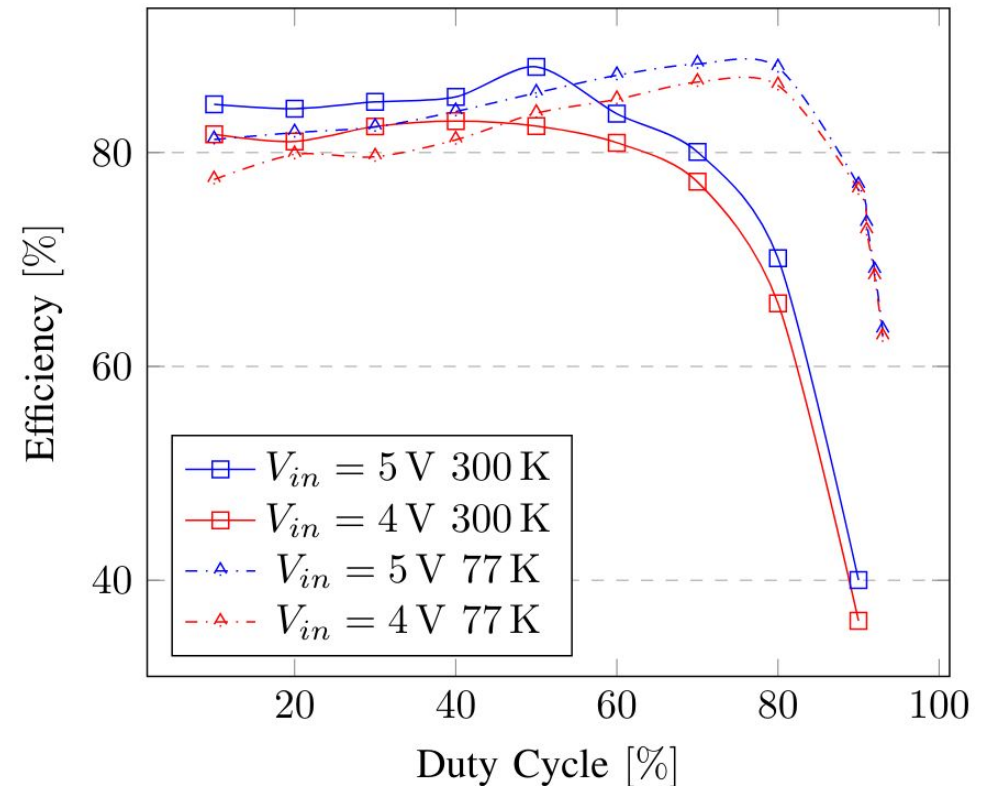
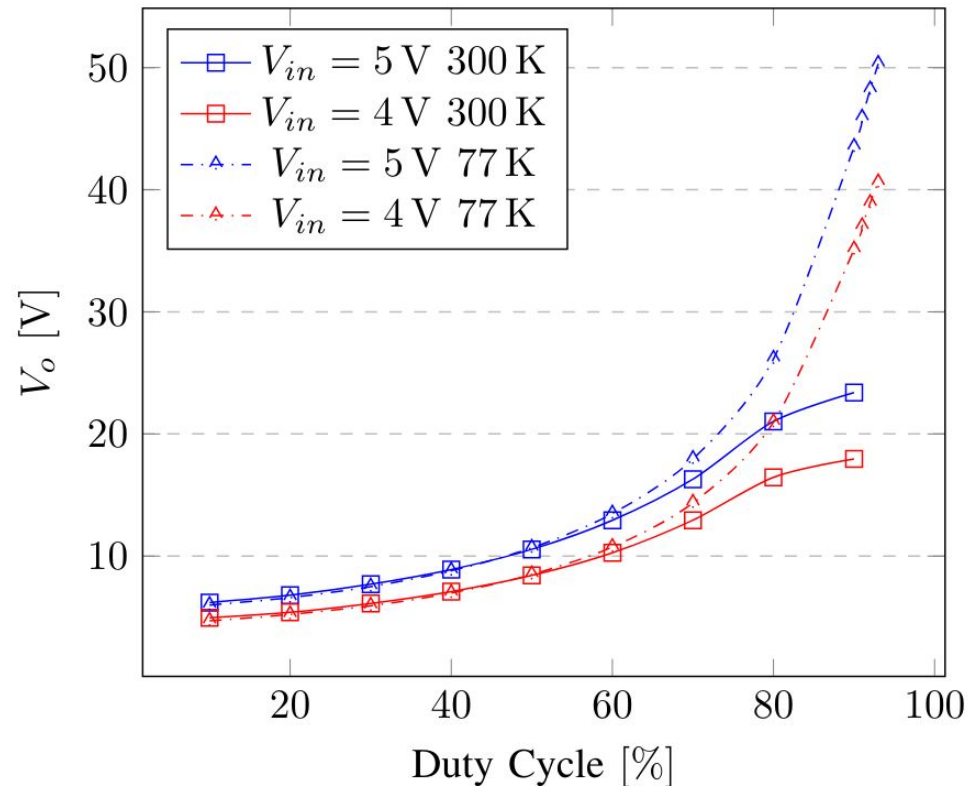


L1	10 mH
D1	BAV16W
C1	C0G 100 nF
R1	10 k Ω
Q1	NTF3055L108T1G

The system is tested at room and LN2 temperature, with different inputs (4V, 5V) and different duty cycle [0.1, 0.93].

- Output readout with a Lecroy HDO6104A oscilloscope.

DC-DC prototype results



- On the left the output voltage, at RoomT it is limited by the inductor series resistor. At LN2 T, it is possible to reach 50 V at 93% of Duty Cycle.
- The efficiency at LN2 is always greater than 60%

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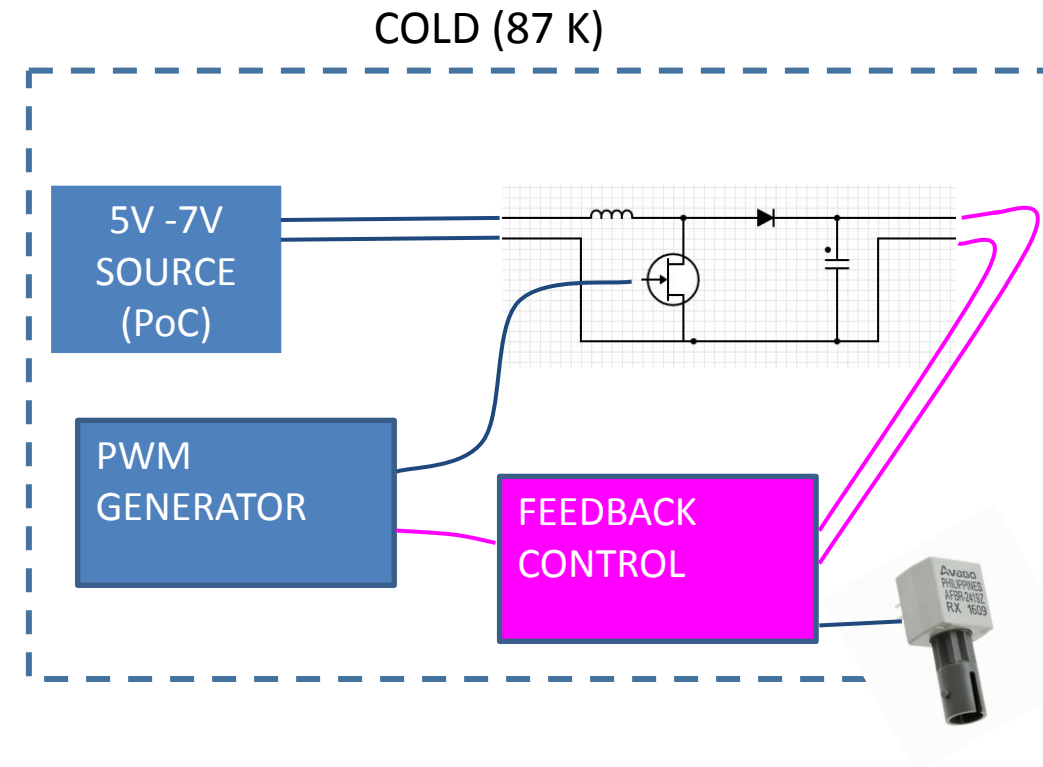
Control design

The idea is to have an **internal feedback** that keeps the output of the DC-DC boost design at the desired set point through the control of the PWM generator

The **set point** can be:

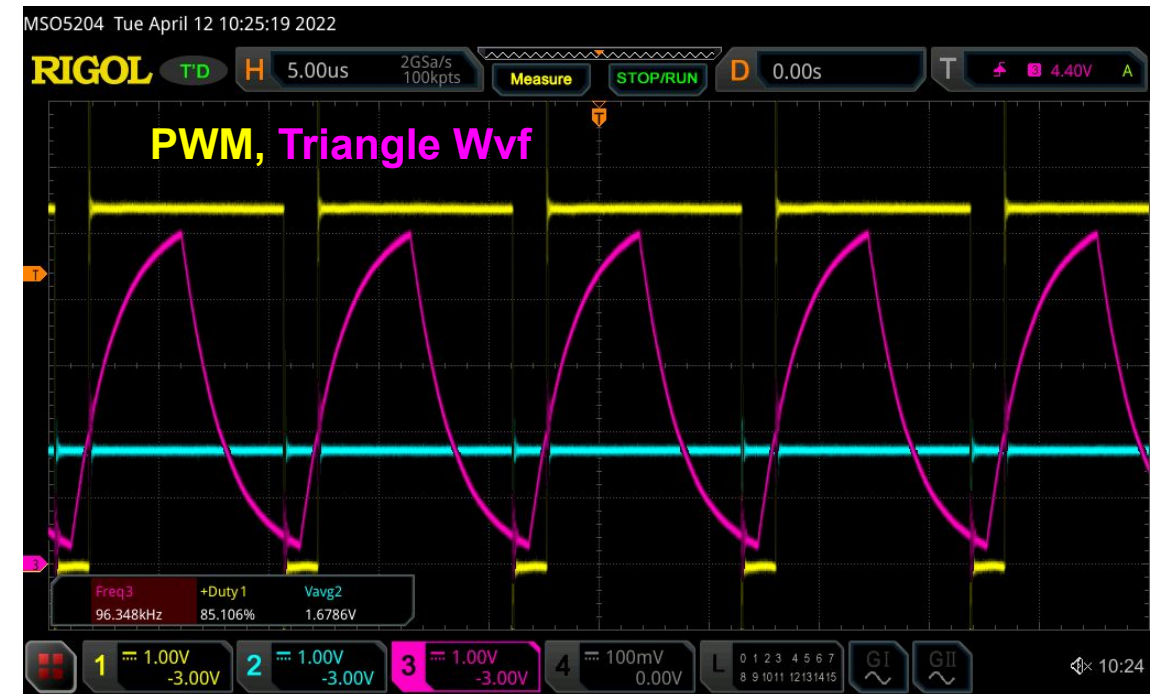
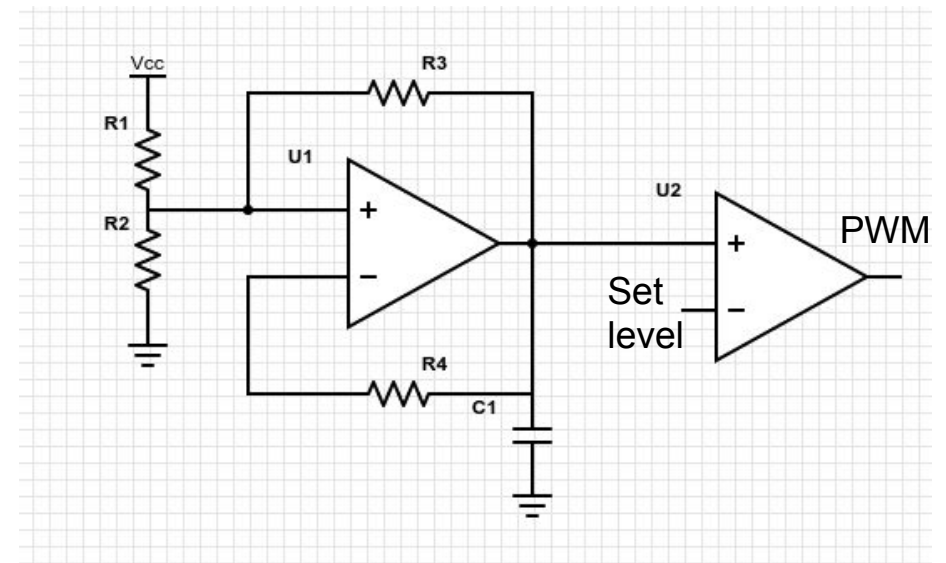
- **Nominal:** Set at the design stage and fixed throughout the entire run. It works in standalone mode
- **External:** The set point can be adjusted with an external communication.

In this way the voltage can be adjustable, but in case of failure of the communication system and/or interfaces, the DC-DC converter will continue working at the nominal set-point.



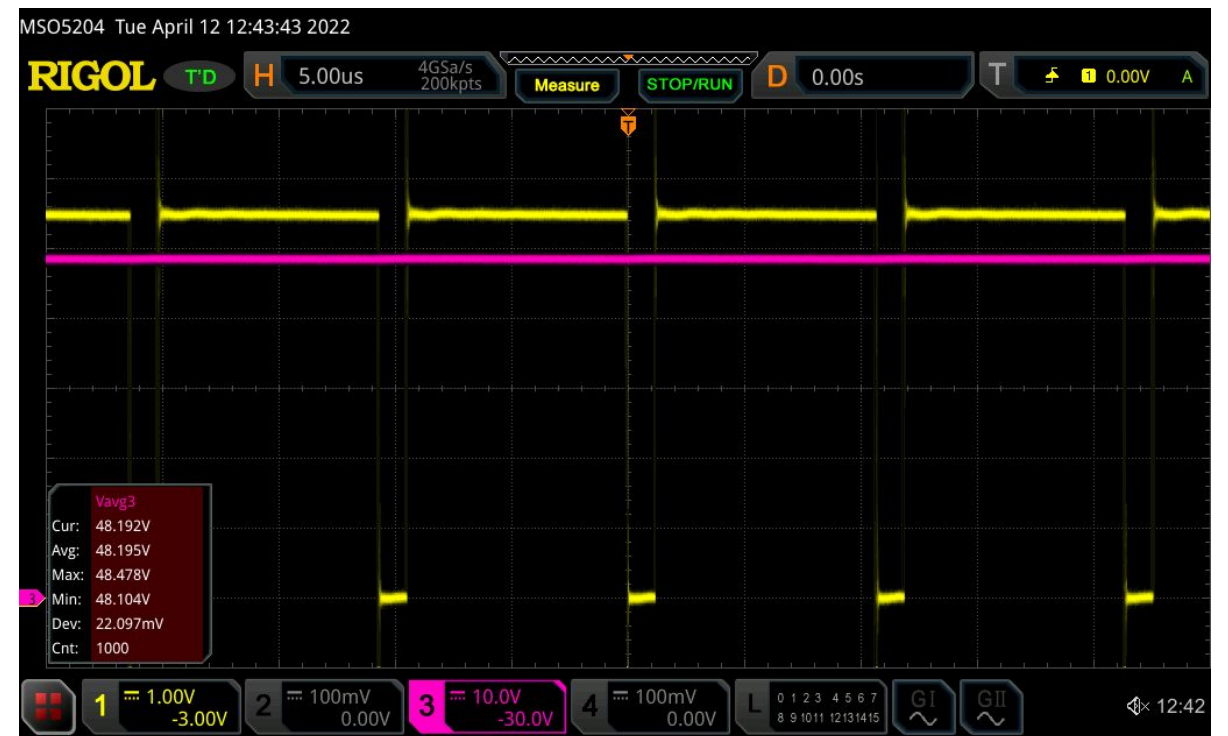
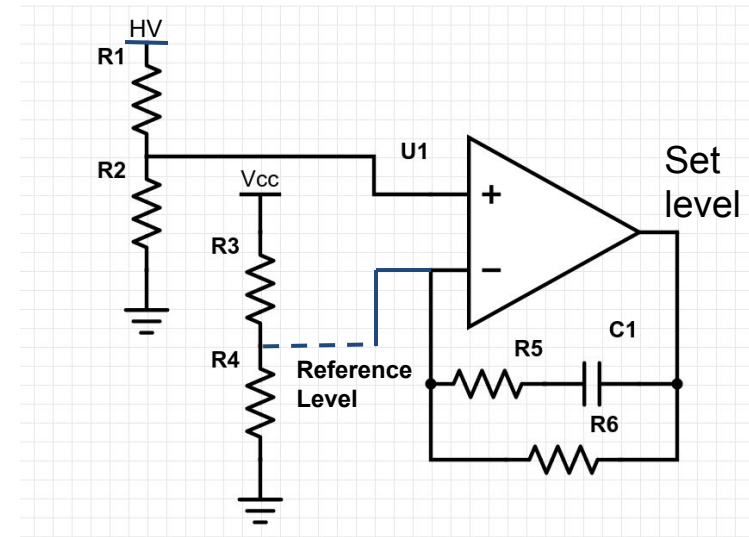
PWM generator

- To drive the MOS switch we built a **PWM (Pulse Width Modulator) generator** based on **two comparators**.
- Some comparators have been tested to work in LN_2
- The first is set in a positive feedback circuit that makes it oscillate, creating a triangle-like waveform
- The **second comparator** produces the PWM taking as input the **triangle waveform** and a **level** that is used to **select** the output **duty cycle**.
- The working frequency is ~ 100 kHz



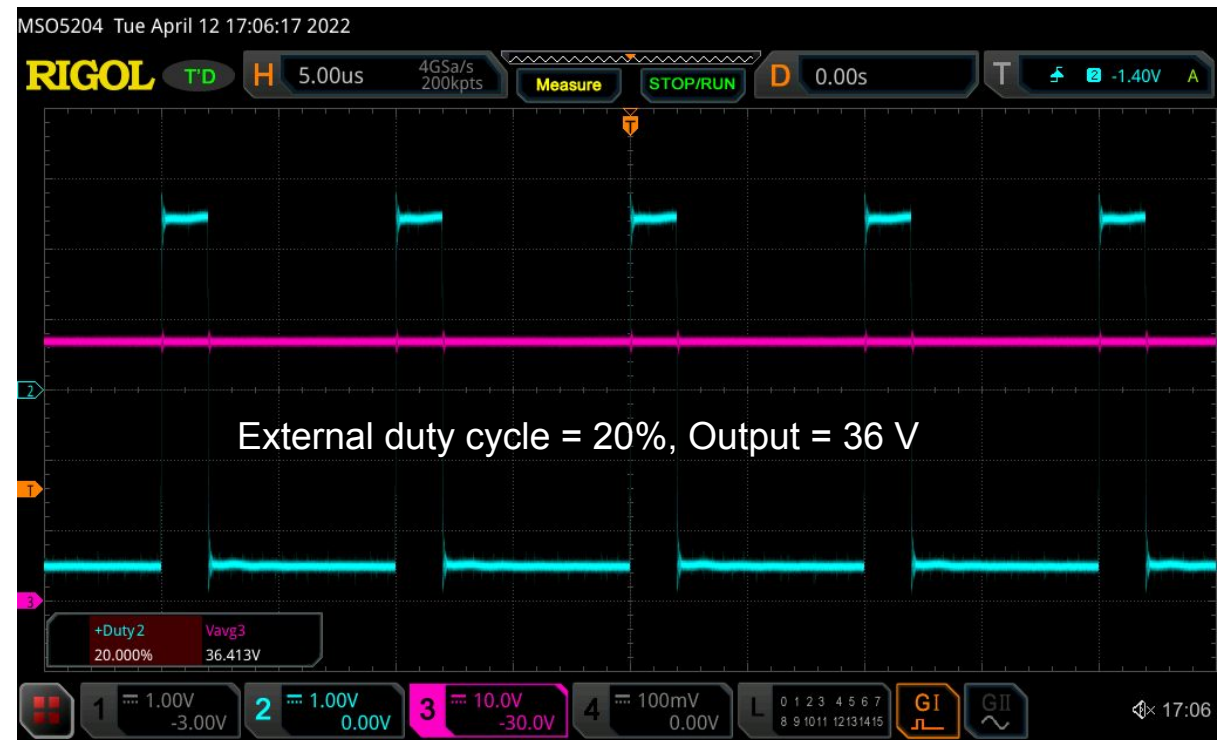
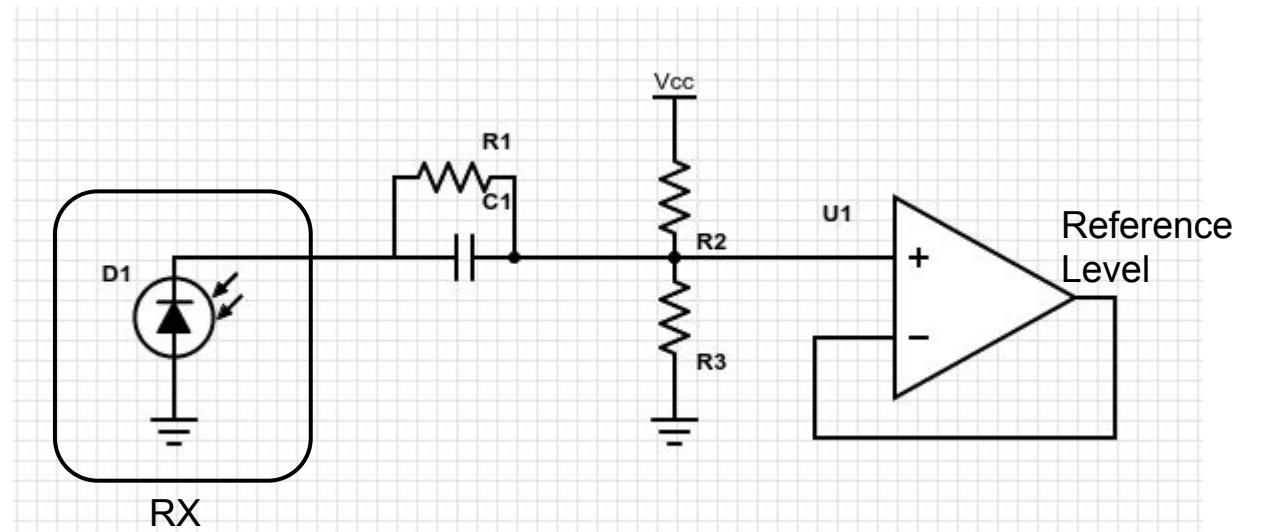
Control feedback

- The **output of the DC-DC boost converter is compared with the nominal** working point with an error amplifier
- Some operational amplifier have been tested to work in LN₂
- The error amplifier output controls the set level of the PWM generator
- **48.2 V** generated and kept constant by the internal feedback at LN₂ temperature (5.5 V input)
- 100 k Ω load \rightarrow 485 μ A
- Efficiency = $(48.5 \times 0.485)/(5.5 \times 10.3) \sim 50\%$ mainly because of control electronics.



External set point

- An external PWM signal drove through an optical fiber can provide an **external working point** to adjust the nominal one
- The output signal of the buffer stage **is summed** to the one of the **internal working point**
- The signal is AC-coupled and in case of absence of communication the R2-R3 divider will bring the DC-DC output back to the pre-set nominal point.
- With the current configuration (5.5 V input) the output can vary in **[35V, 51V]**



Integration considerations

- Next step is the refinement of components values and the design of a PCB
- The **PCB will be fully integrated** with the **DUNE VD Cold Electronics board**, we target the DC-DC test in ColdBox #3
- The communication receiver to adjust the working point will be implemented as a daughter board

Assuming a max saturation of 2000 PE per channel, that would correspond to:

$$2000 \text{ [PE/Ch]} \times 5 \cdot 10^{-6} \text{ [A]} \times 200 \text{ [Hz]} \times 20 \cdot 10^{-6} \text{ [s]} = 40 \text{ } \mu\text{A/Ch}$$

PoC / Power fibers	Communication fiber	DC-DC	X-ARAPUCA*	Tot. Fibers
1	1	1	Up to 5 Ch	2
2	1	1	Up to 10 Ch	3
2	2	2	Up to 10 Ch	4

*Assuming 50% efficiency

Conclusions

- A multi stage circuit for bias generation in cryogenic environment (LN₂) has been developed with discrete components
- The different stages have been tested independently and in-sequence, while adding functionalities to the circuit:
 - Typical design for the DC-DC boost converter, fully characterized and operational at LN₂ temperature
 - PWM generation tested in cold, and integrated successfully in the DC-DC design
 - Internal feedback control implemented and tested in LN₂ for internal pre-set working point
 - External control increase flexibility and does not add risk. The optical communication was emulated with an electrical signal and managed to adjust the output voltage [35V, 51V].

and steps forward

- Develop the optical communication at the cryogenic level
- Develop a PCB to be integrated in the DUNE VD Cold Electronics board