

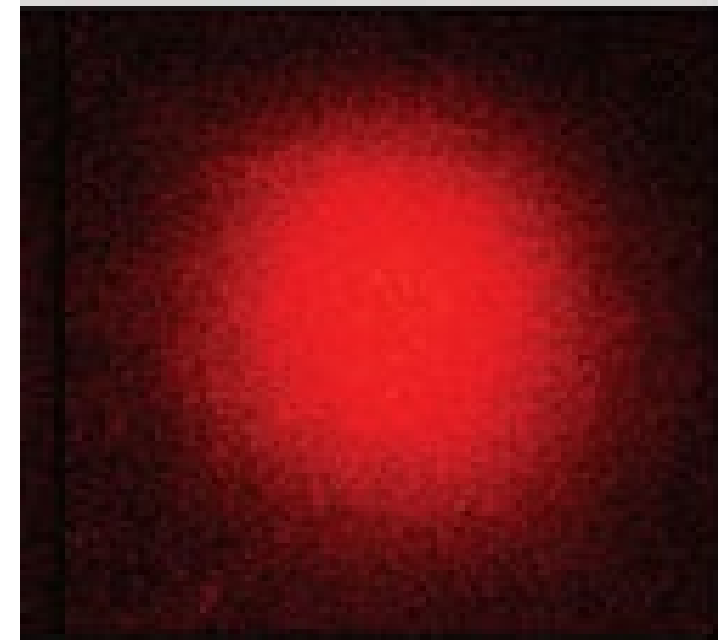
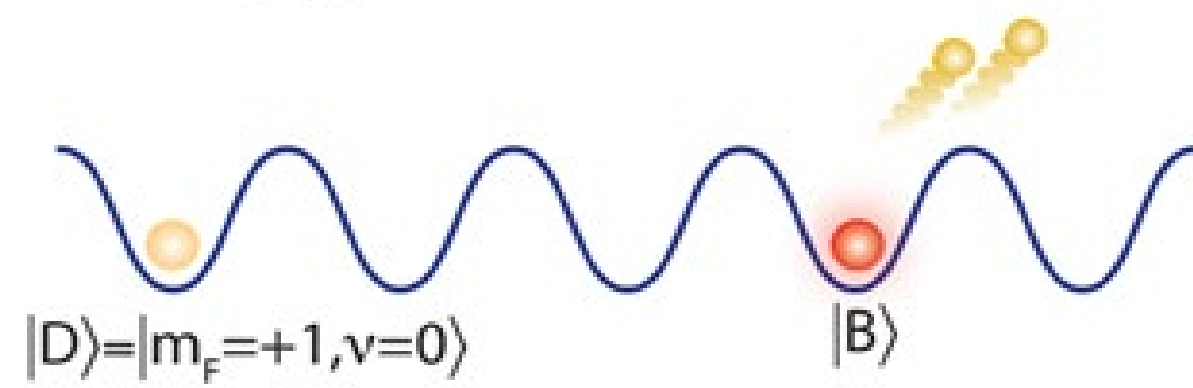
# Quantum Sensing of Low-Frequency Electromagnetic Signals

Kent Irwin

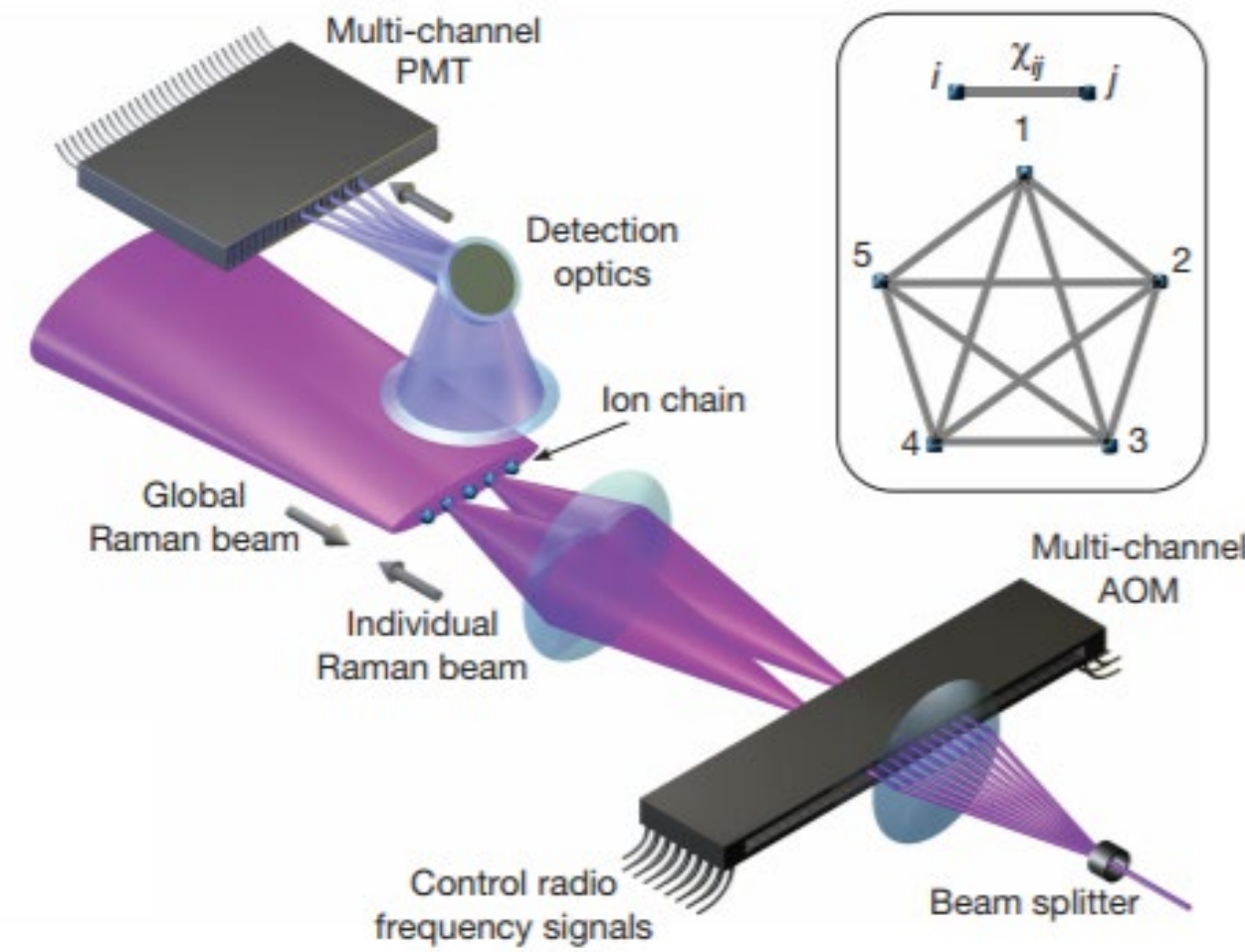
Dept. of Physics, Stanford University  
SLAC National Accelerator Laboratory

US QIS Summer School  
August 14, 2023

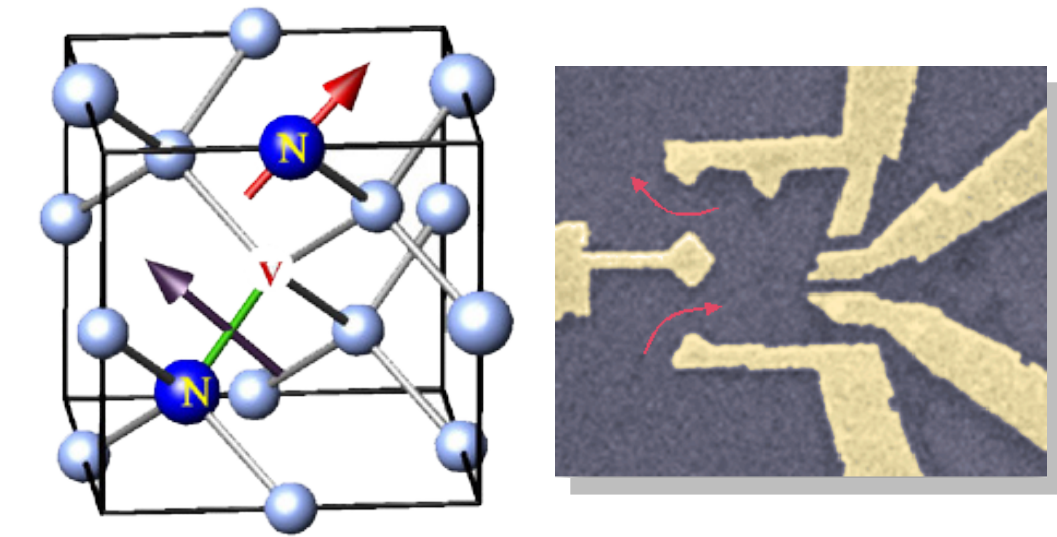
# Quantum sensing modalities



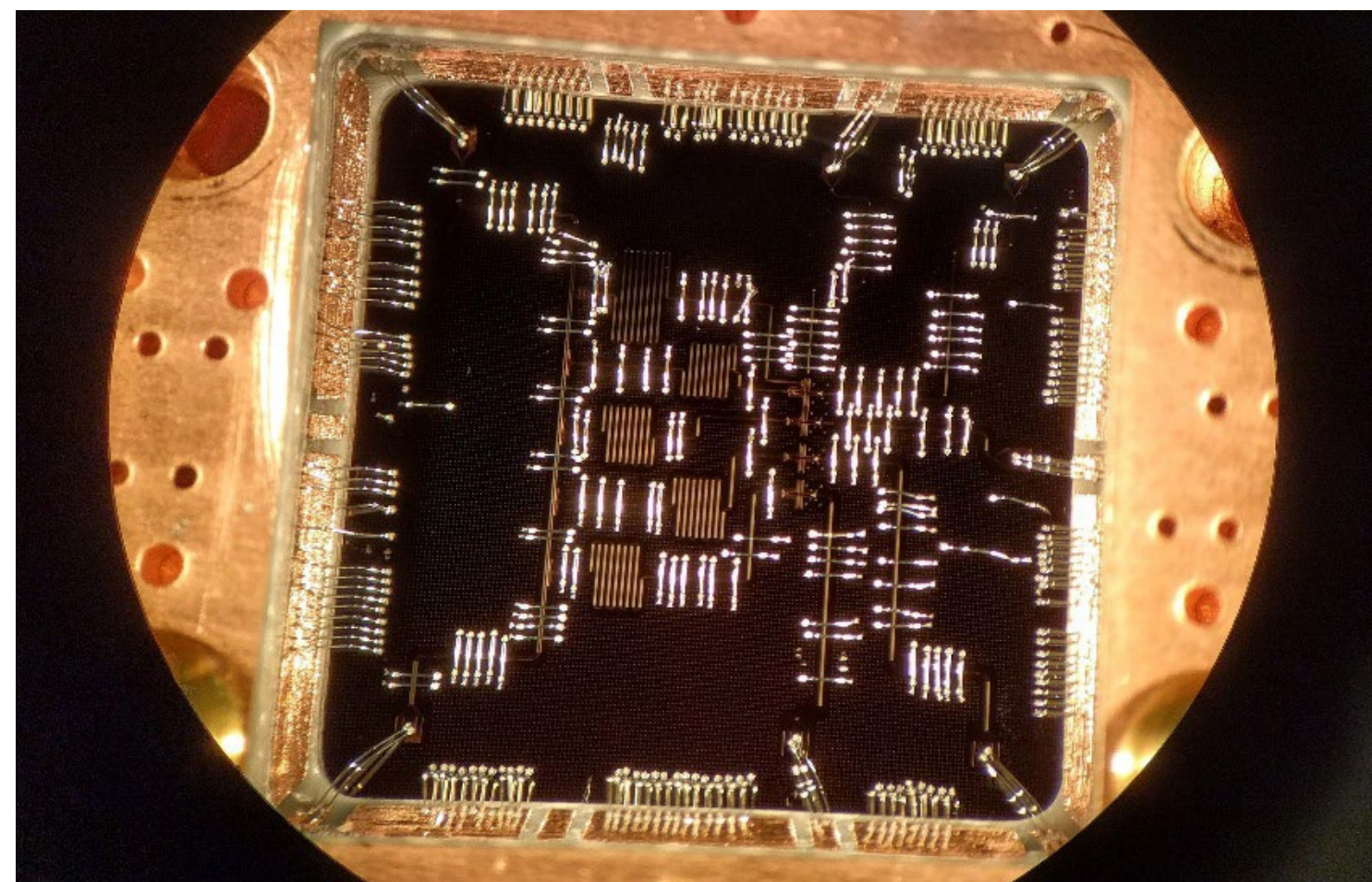
Cold atoms



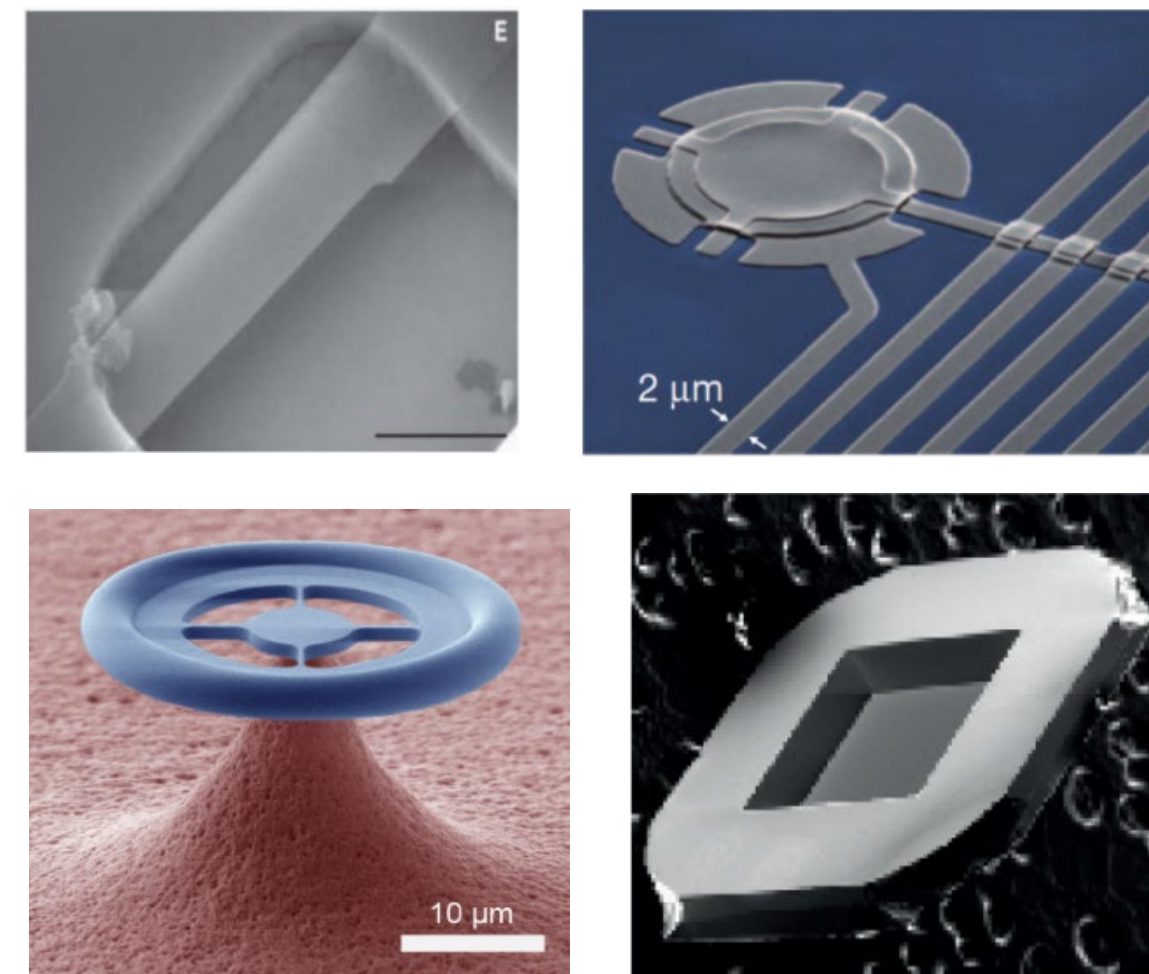
Trapped ions



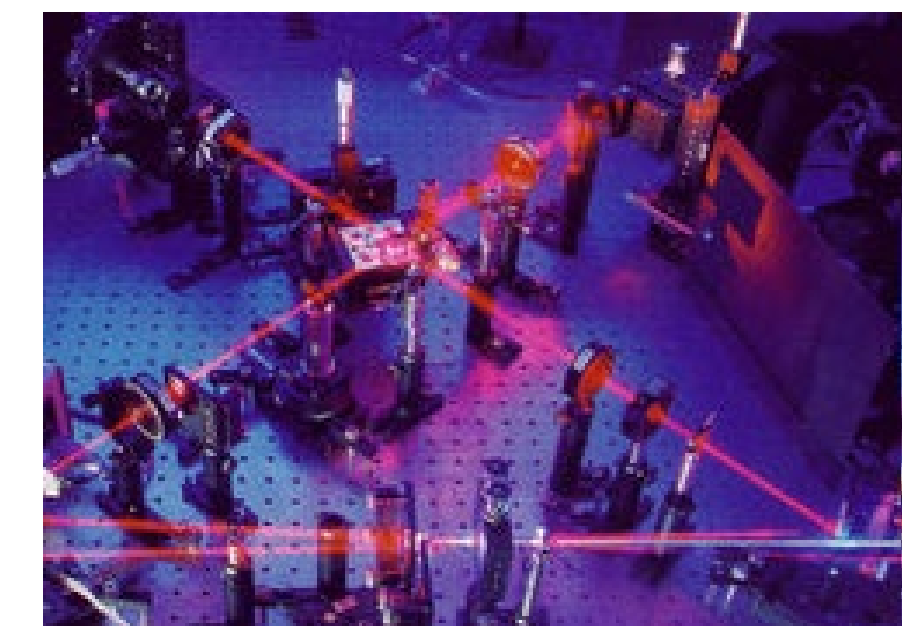
Spins (electron / nuclear) ...



Superconducting



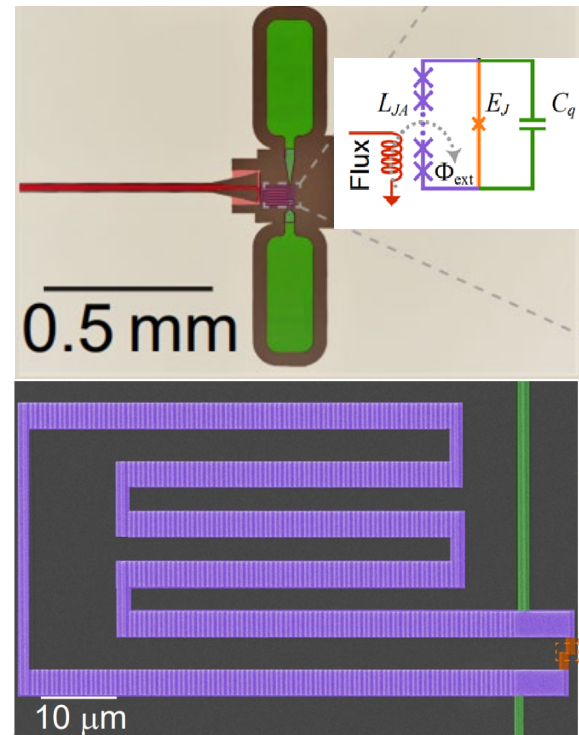
Optomechanical



Optical

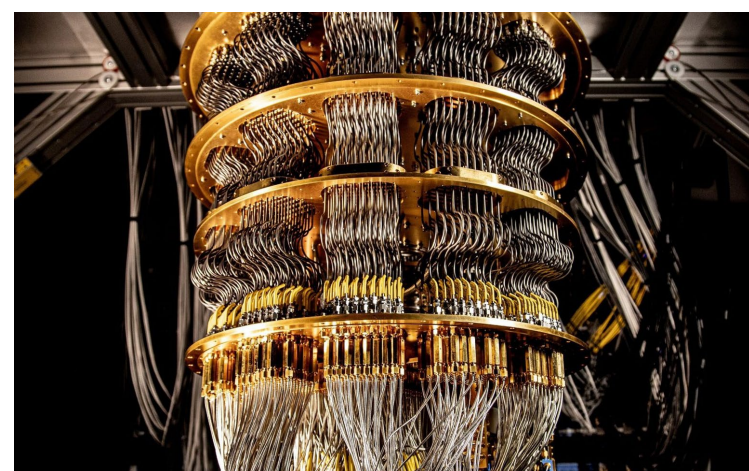
# One Motivation: A Golden Age for QIS and Dark Matter

## The Quantum Information Revolution



Qubit (Schuster)

- **Advances in quantum control**  
⇒ Major governmental, industrial, and academic investment in new quantum technologies
- Near-term opportunities for leveraging quantum advantage:
  - Sensing beyond the standard quantum limit
  - Quantum simulations of physical phenomena intractable to classical computers

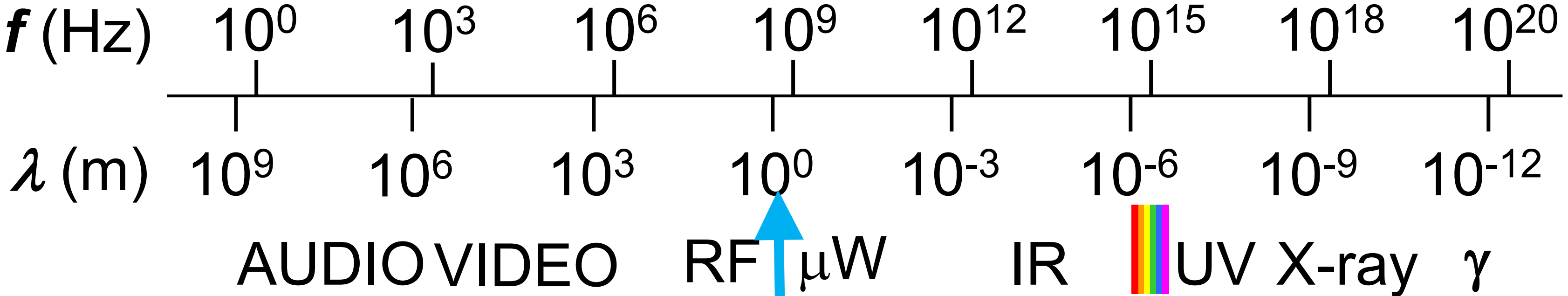


Quantum computer (Google)

## The Dark Matter Revolution

- Progress in theoretical understanding of:
  - Diverse range of dark-matter candidates
  - QCD axion: uniquely motivated for resolving the Strong CP problem
- Searching for diverse range of dark-matter candidates requires diverse new quantum sensor technologies (photon detectors, atomic clocks, spins, superconducting qubits, ...)
- Searching for QCD axion requires quantum sensitivity enhancement: reduce time to fully measure QCD axion band from millennia to years

# Focus on electromagnetic signal wavelength $\geq$ human scale

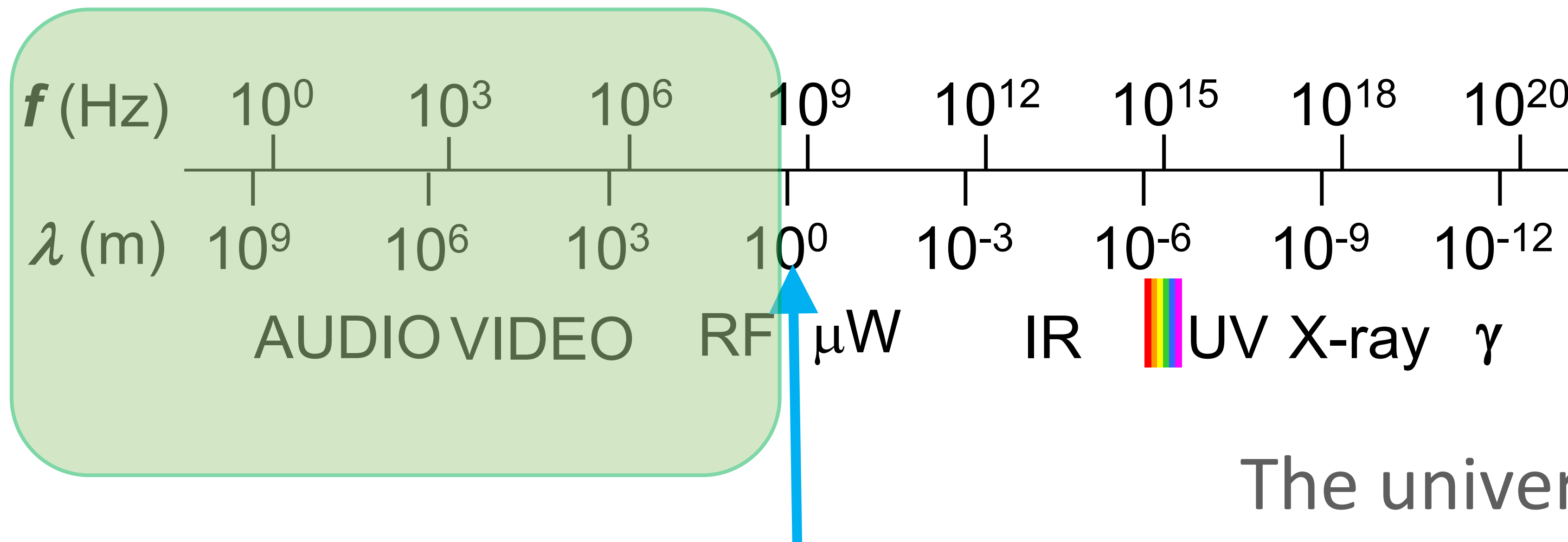


The universe  $\sim 2.7$  K

300 MHz  $\sim$  0.015 Kelvin  $\sim$  1 m

300 MHz  $\sim$  human scale  $\sim$  dilution refrigerator temperature

This lecture will address techniques to measure frequencies  $< 300$  MHz  
(including upconverting them to higher frequencies)

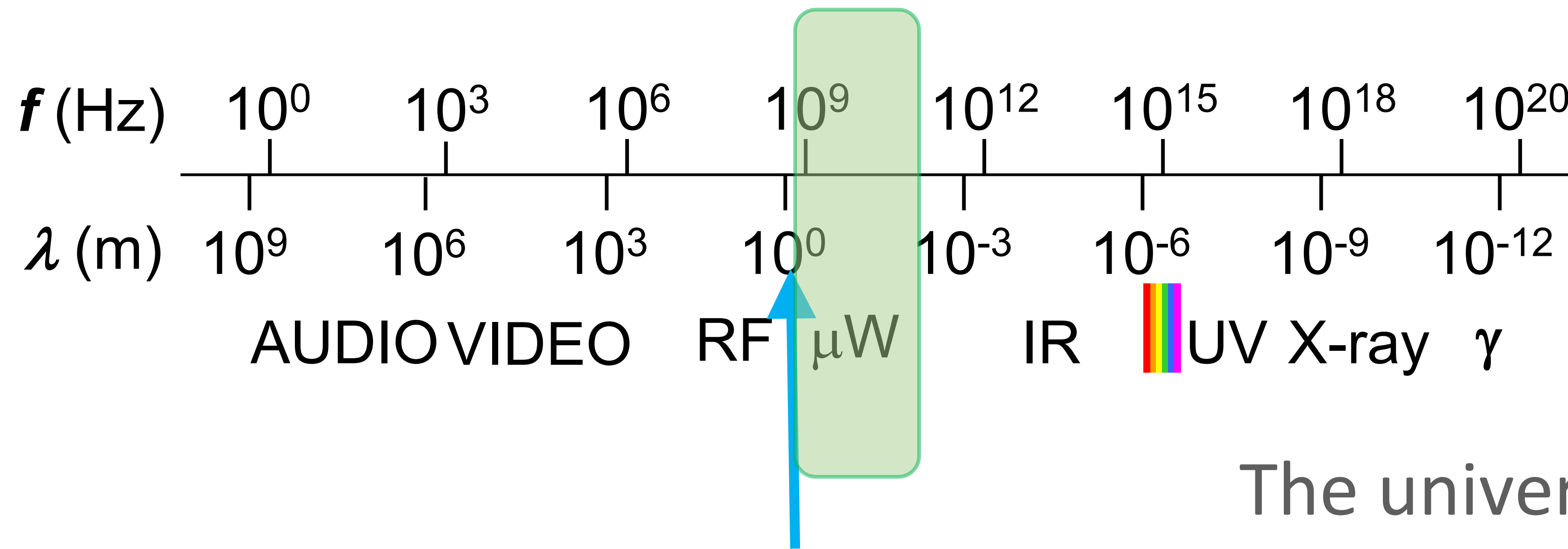


The universe  $\sim 2.7$  K

300 MHz  $\sim 0.015$  Kelvin  $\sim 1$  m

300 MHz  $\sim$  human scale  $\sim$  dilution refrigerator  
temperature

Aaron Chou's lecture (next) will address microwave-frequency techniques with qubits



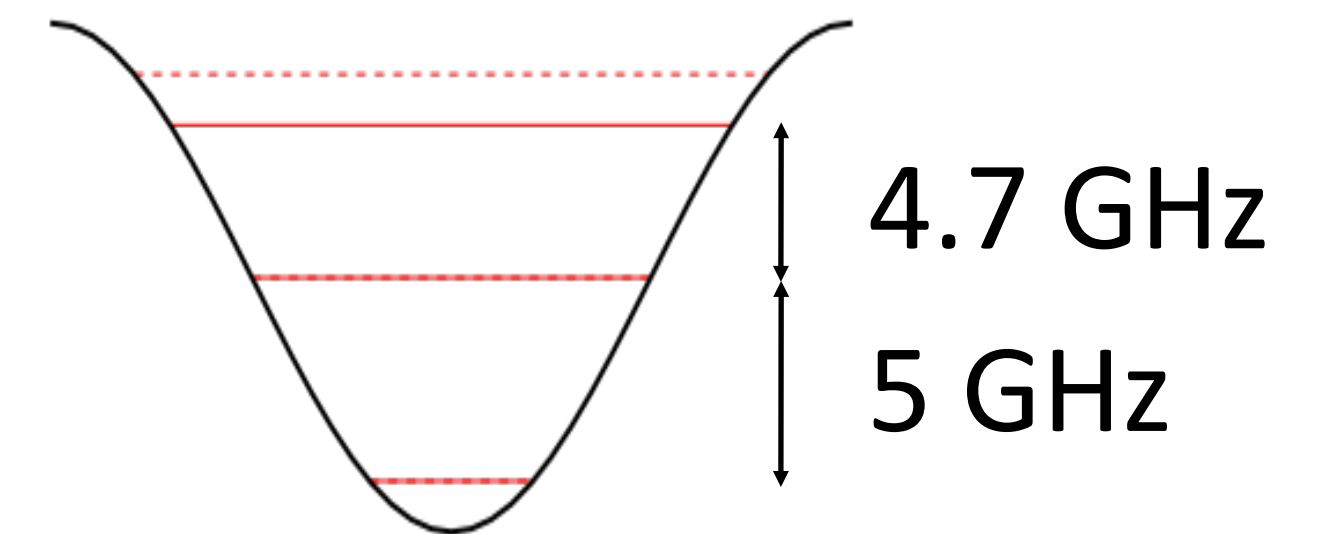
The universe  $\sim 2.7$  K

300 MHz  $\sim$  0.015 Kelvin  $\sim$  1 m

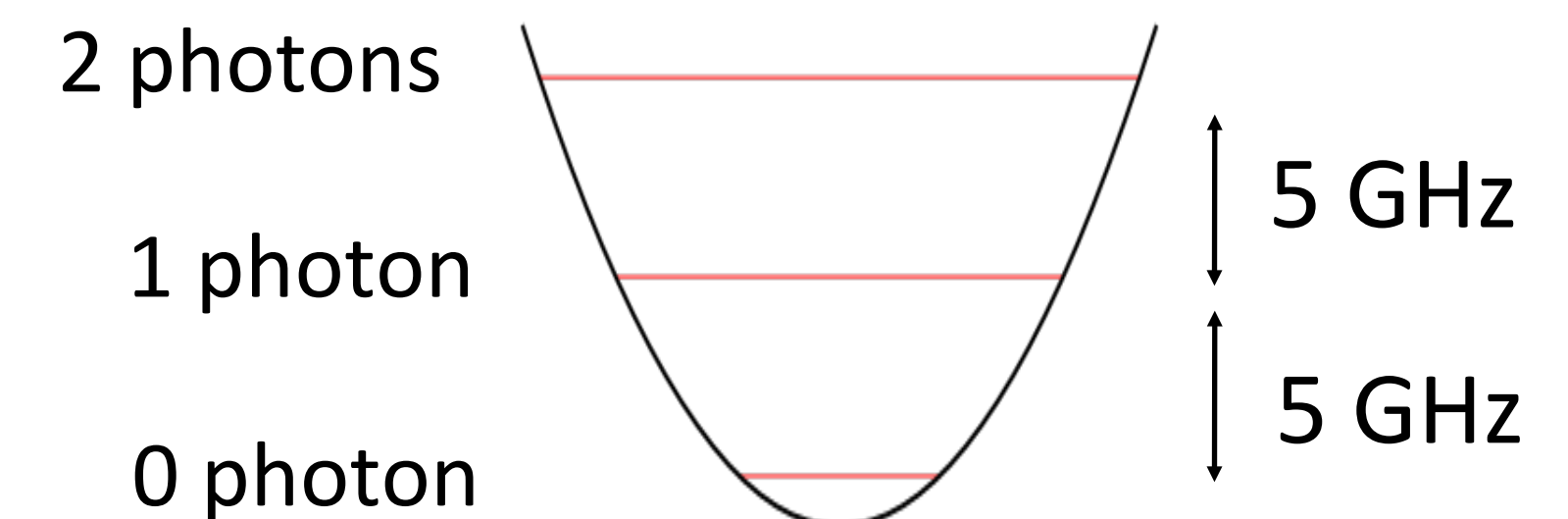
300 MHz  $\sim$  human scale  $\sim$  dilution refrigerator temperature

# Continuous variables quantum information

- A qubit, or two-level system, is in some superposition of two states,  $|0\rangle$  and  $|1\rangle$ . It is “digital”-like.
- Physical observables (e.g. the strength of an electromagnetic field) have continuous intervals. They are “analog”-like.
- Often, continuous variables signals are sensed in a weak, continuous measurement, rather than a single projection. (the realized quantum limits are similar in either case).
- *The quantum optics approach to continuous variables:* model each mode of an electromagnetic field as a harmonic oscillator with annihilation & creation operators
- The state of the harmonic oscillator can be expressed as a phasor diagram, with  $\hat{X}$  and  $\hat{Y}$  quadrature components



## Qubit with nonlinear level spacing

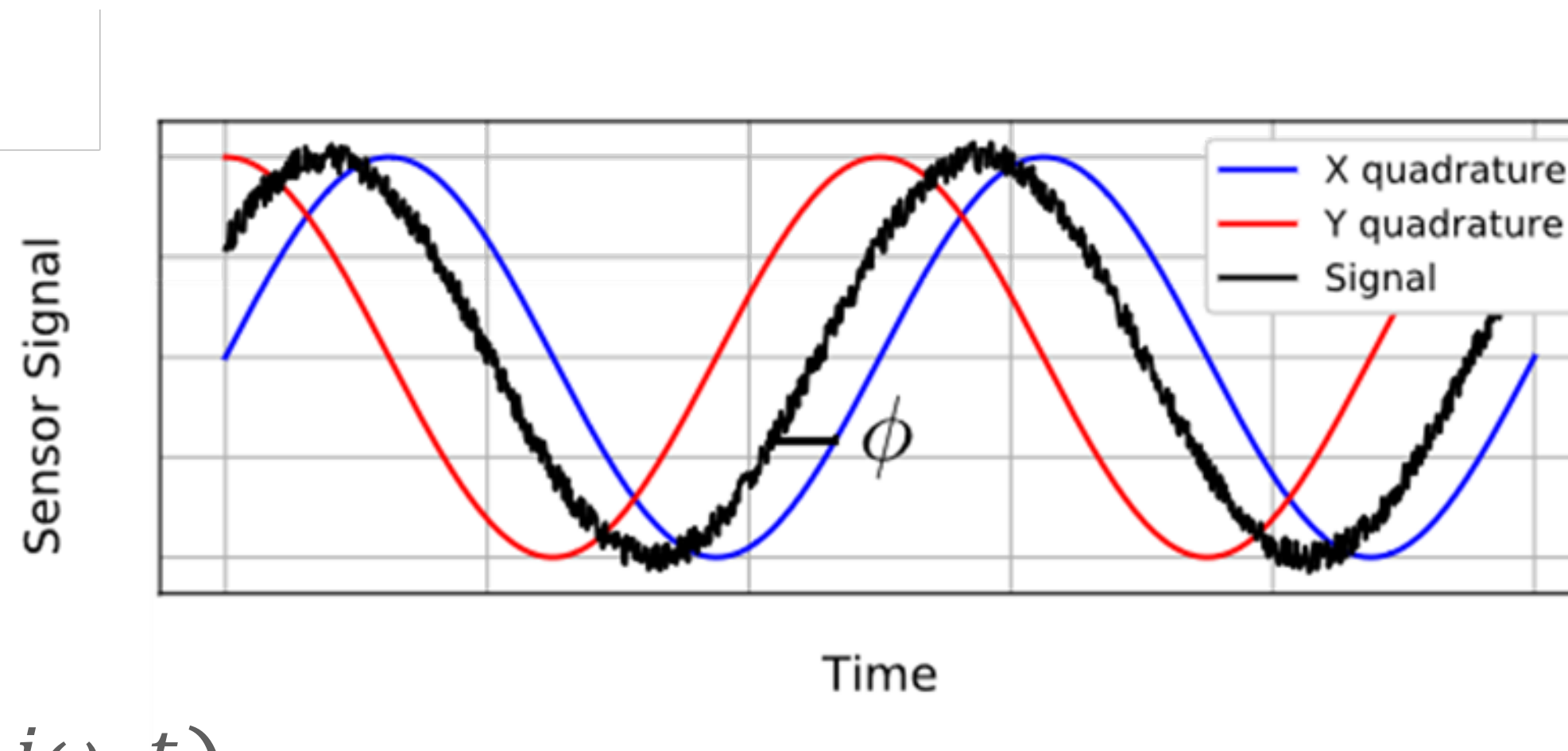


## SHO with linear level spacing

$$\hat{H} = \hbar\omega(a^\dagger a + 1/2)$$

# Low Frequency LC Circuit Quadratures of a single mode

Low-frequency signal (black) has components in the X-quadrature (blue) and in the Y-quadrature (red)



$$\hat{X} = \frac{1}{\sqrt{2}} (\hat{a} e^{i\omega_a t} + \hat{a}^\dagger e^{-i\omega_a t})$$

$$\hat{Y} = \frac{1}{\sqrt{2}} (\hat{a} e^{i\omega_a t} - \hat{a}^\dagger e^{-i\omega_a t})$$

$$[\hat{X}, \hat{Y}] = i$$

Consider a magnetic flux signal  $\hat{\Phi}(t)$

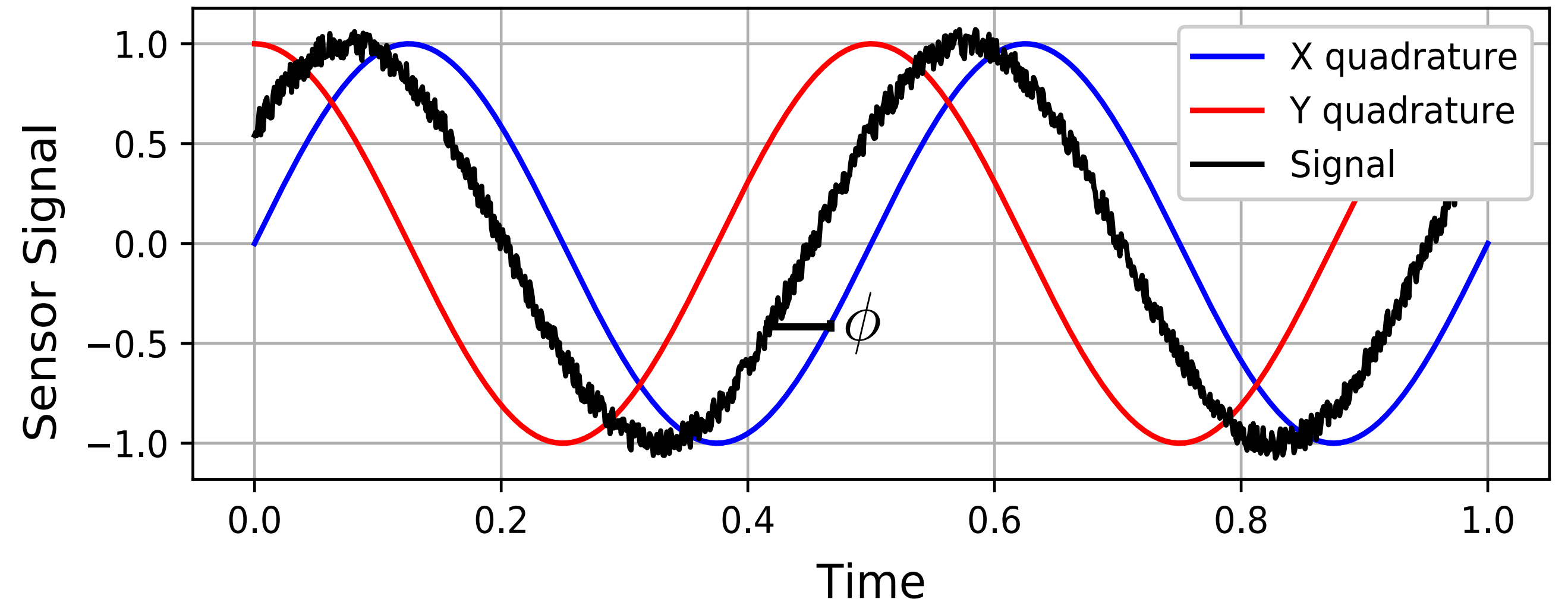
$$\hat{\Phi}(t) = \sqrt{2}\Phi_{zpt} (\hat{X}(t) \cos \omega_a t + \hat{Y}(t) \sin \omega_a t)$$

See review: Clerk et al, RMP **82**, 1155 (2010)



# The Standard Quantum Limit for Electromagnetism

- A generic E&M signal has ‘sine’ and ‘cosine’ components, which do not commute.
- The Heisenberg uncertainty principle says we cannot measure both perfectly, so an amplifier must add noise.
- The “standard quantum limit” is achieved when both components are measured as well as possible, saturating the Heisenberg Uncertainty Relation
- Equivalent to one photon of noise



$$[\hat{X}, \hat{Y}] = i$$
$$\Delta X \Delta Y \geq \frac{1}{4}$$

Commutator  
Heisenberg  
Uncertainty  
Relation

# The Standard Quantum Limit (SQL) in a harmonic oscillator

The Hamiltonian of a harmonic oscillator is

$$\hat{H} = \hbar\omega(a^\dagger a + 1/2)$$

The Hamiltonian can be written in the cosine component ( $\hat{X}$ ) and the sine component ( $\hat{Y}$ )

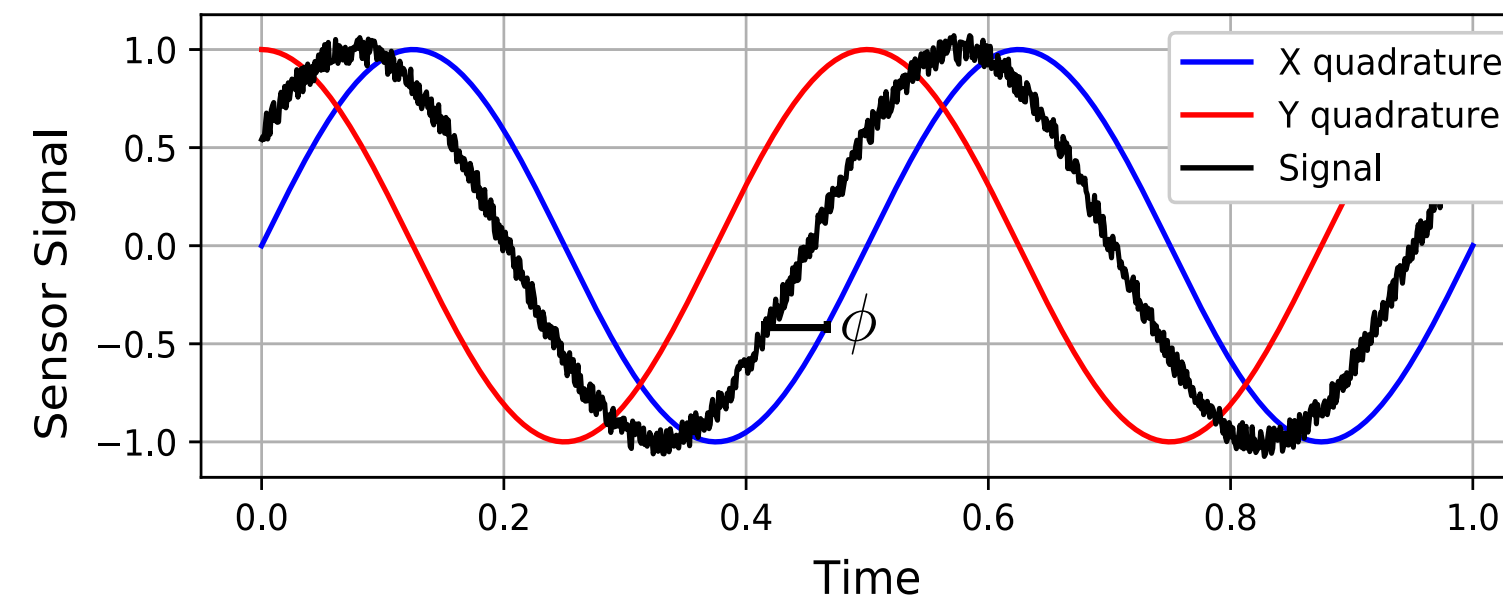
$$\hat{H} = \frac{\hbar\omega}{2}(\hat{X}^2 + \hat{Y}^2)$$

$$[\hat{X}, \hat{Y}] = i$$

vacuum noise

$$\Delta\hat{X}\Delta\hat{Y} \geq \frac{1}{2}$$

Time representation



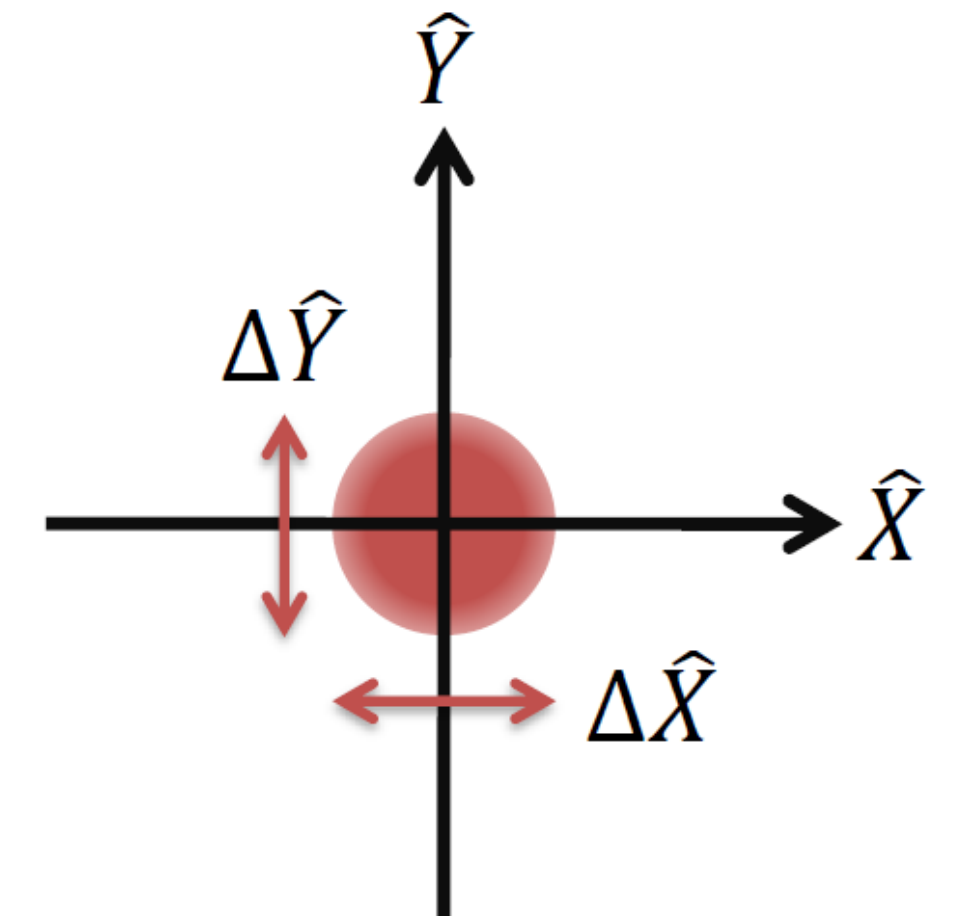
When amplified, add one more 1/2 quantum

$$N_{add} \geq \frac{1}{2}$$

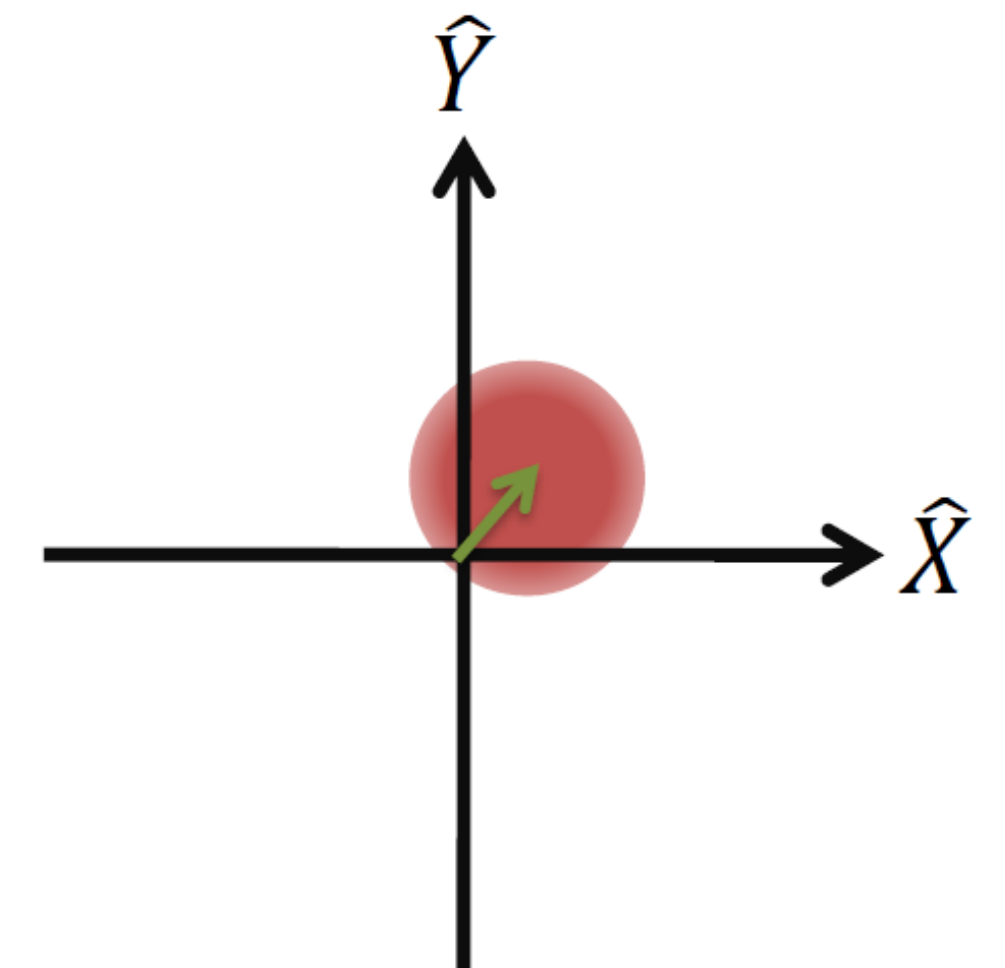
“SQL”: 1 photon of noise

- Per measurement
- Measure ~ once per resonator coherence time

Phasor representation



With nonzero expectation value

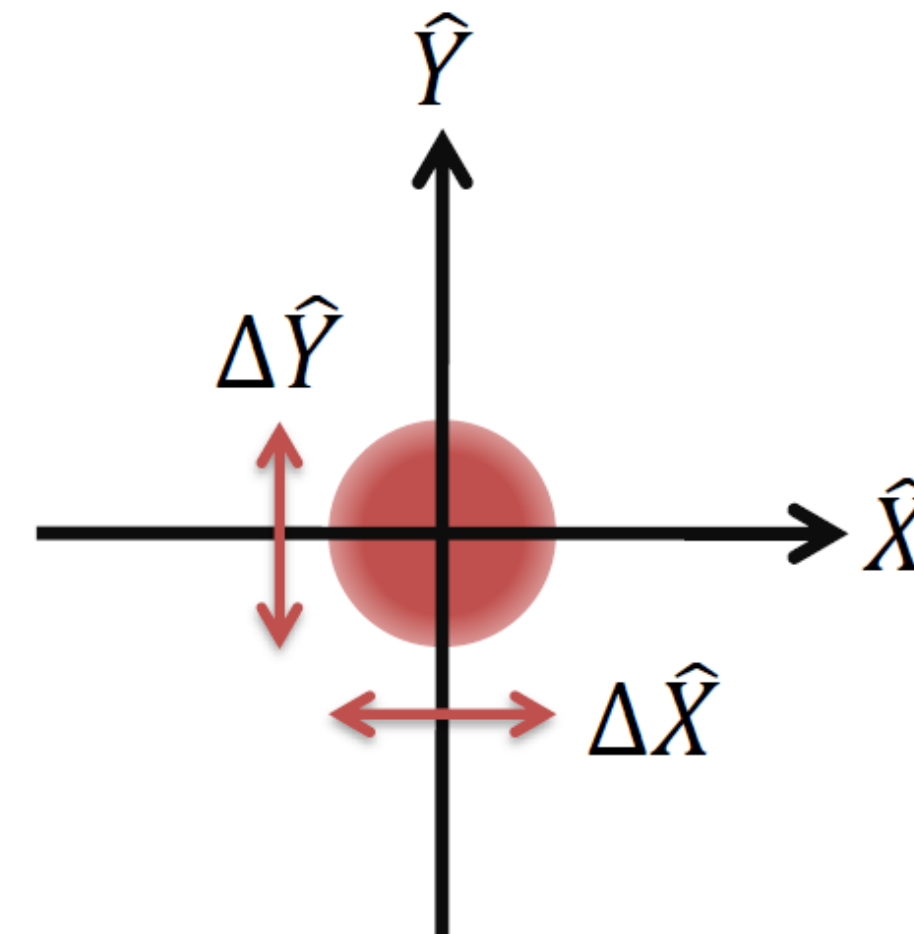
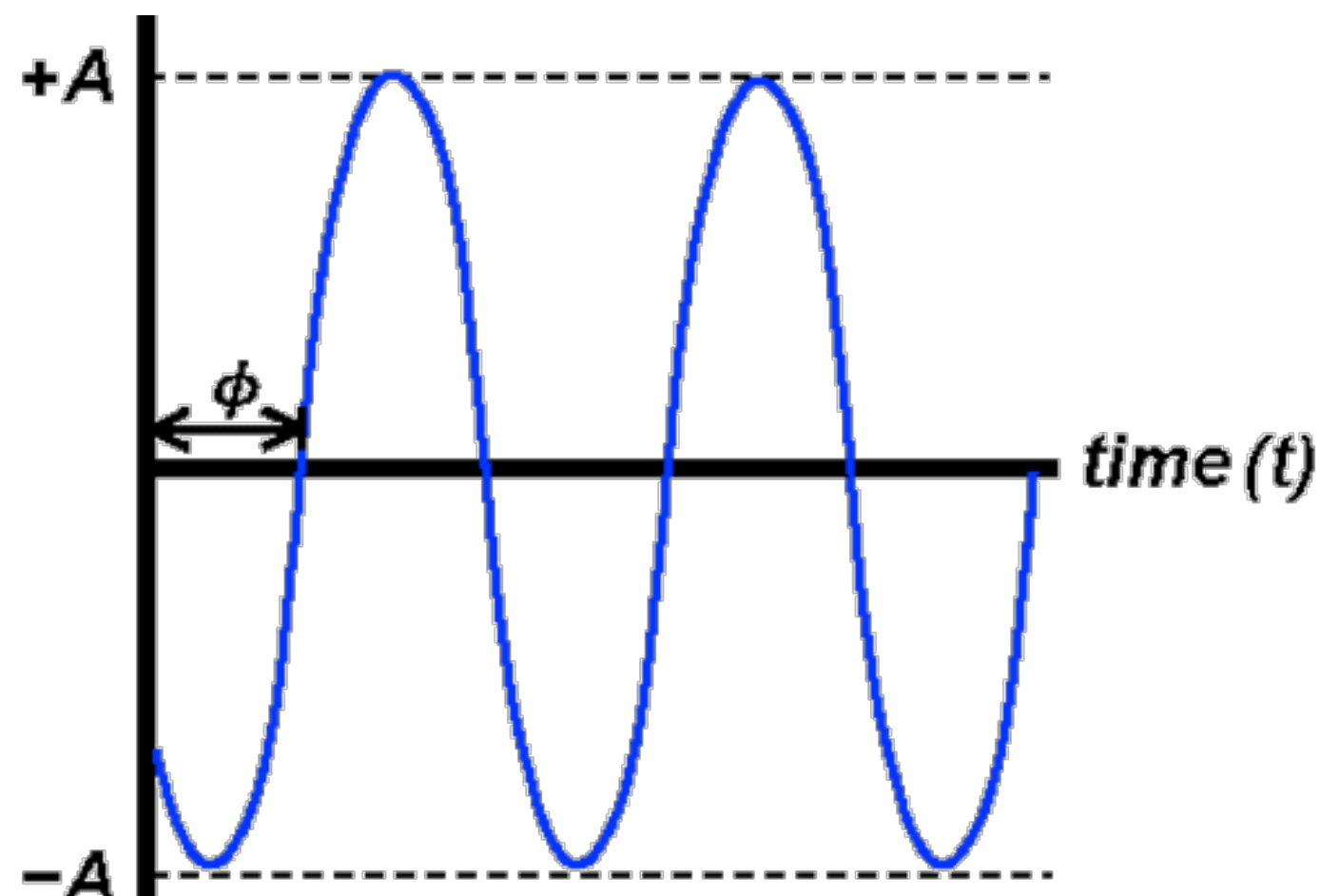


# You can't know both amplitude and phase perfectly

Heisenberg tells us that you can't know the position and momentum of a particle perfectly at the same time:

$$\Delta x \Delta p \geq \frac{\hbar}{2}$$

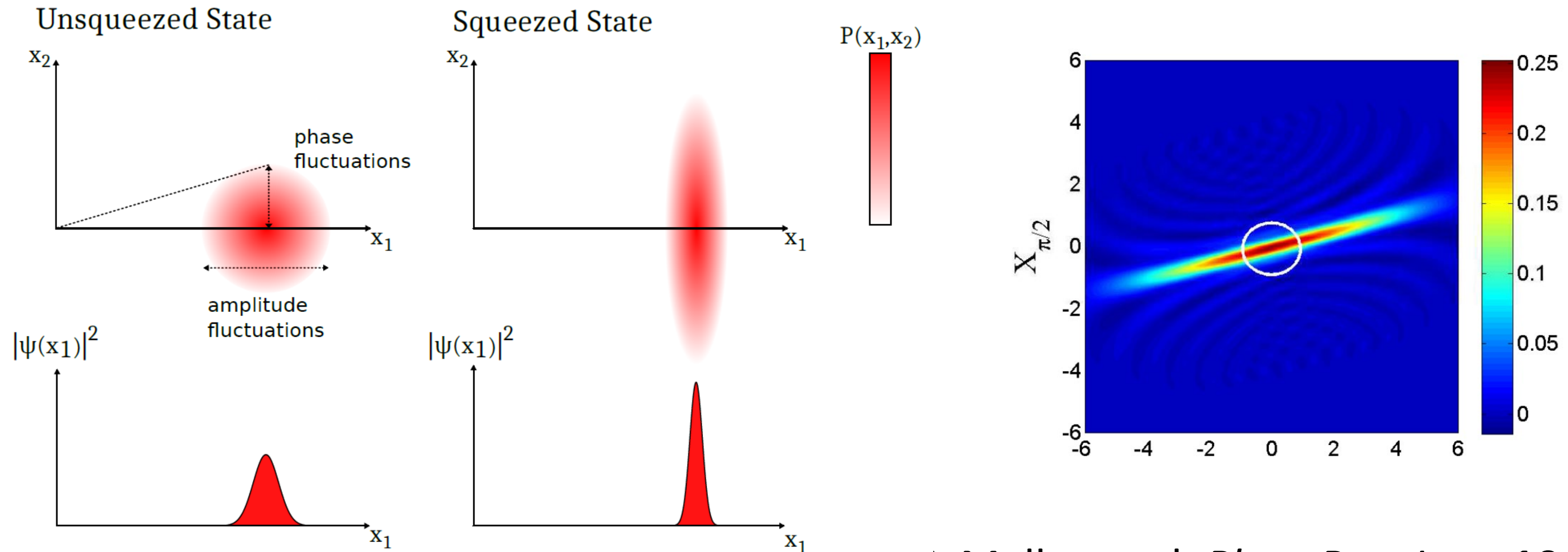
$$A \cos(\omega t + \phi) = X \cos(\omega t) + Y \sin(\omega t)$$



A "classical" sensor measures both amplitude and phase with equal sensitivity, limited by the Standard Quantum Limit of  $\hbar\omega$

What if I don't care about phase???

# One way to evade the SQL - squeezing



F. Mallet et al. *Phys. Rev. Lett.* **106**, 220502 (2011).

- Don't measure both amplitude and phase. (Equivalently, don't measure both sine and cosine quadratures)
- There are several ways to achieve this outcome. They are deeply inter-related (and involve entanglement), and all obey the Uncertainty Principle.

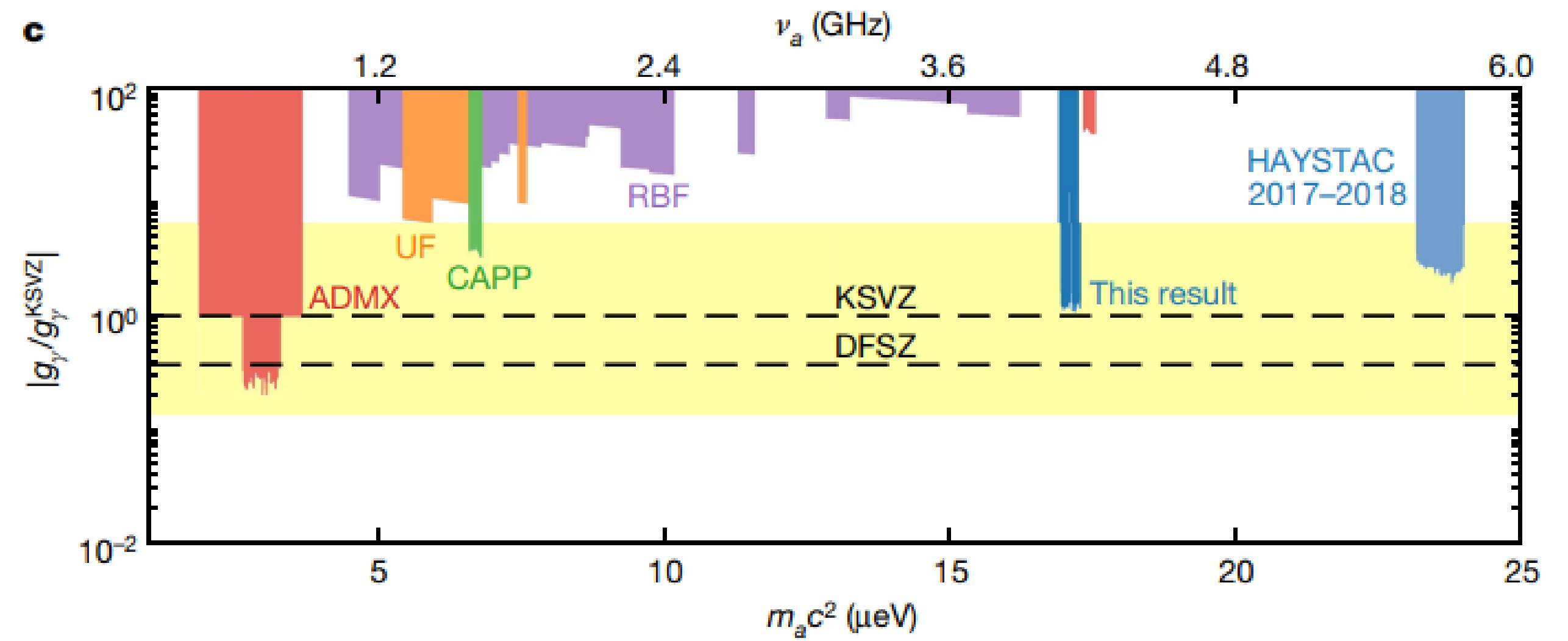
# Better science thru evading the Standard Quantum Limit

LIGO Livingston



- LIGO performs quantum measurements at better than the SQL to measure gravitational waves more sensitively
- LIGO uses squeezing to improve interferometry, but it could also utilize backaction evasion (BAE).

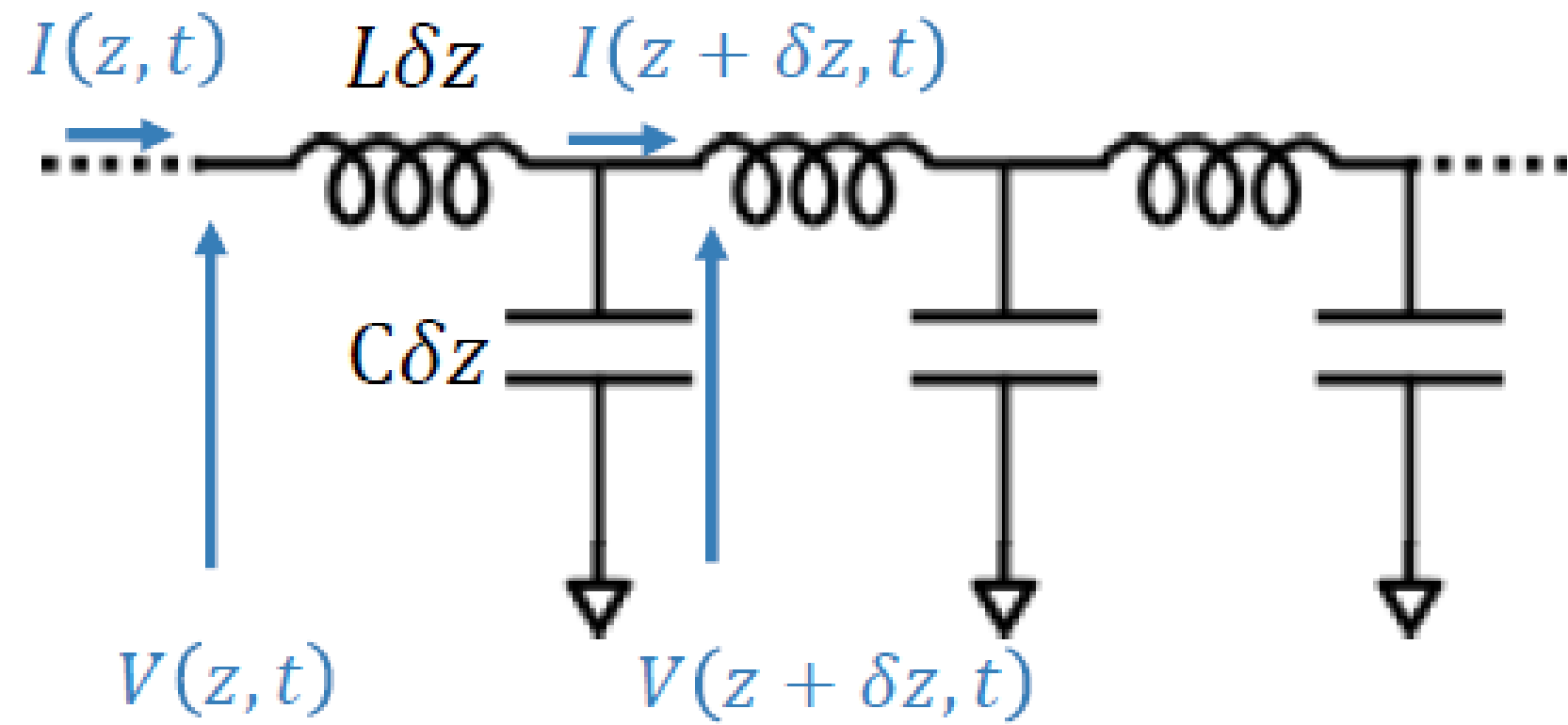
# HAYSTAC: Faster science through squeezing



HAYSTAC quantum accelerated science reach

Backes, Kelly M., et al. "A quantum enhanced search for dark matter axions."

*Nature* 590.7845 (2021): 238-242.



From Florent's lecture:

Consider the single propagating mode limit

$$\begin{cases} -\frac{\partial V(z,t)}{\partial z} = L \frac{\partial I(z,t)}{\partial t} \\ -\frac{\partial I(z,t)}{\partial z} = C \frac{\partial V(z,t)}{\partial t} \end{cases}$$

$$\begin{cases} \frac{\partial^2 V(z,t)}{\partial z^2} - LC \frac{\partial^2 V(z,t)}{\partial t^2} = 0 \\ \frac{\partial^2 I(z,t)}{\partial z^2} - LC \frac{\partial^2 I(z,t)}{\partial t^2} = 0 \end{cases}$$

$$\begin{cases} V(z,t) = V^+ e^{i[\omega t - kz]} + V^- e^{-i[\omega t - kz]} \\ I(z,t) = \frac{V^+}{Z_0} e^{i[\omega t - kz]} - \frac{V^-}{Z_0} e^{-i[\omega t - kz]} \end{cases}$$

$$Z_0 = \sqrt{\frac{L}{C}}$$

$$v_p = \frac{\omega}{k} = \frac{1}{\sqrt{LC}}$$

## Single-mode, waves vs. IV

- In a single-moded signal, at any point in the transmission line, **four numbers** are required to describe the electromagnetic signal.

- Propagating wave:

Re{V+}, Im{V+} for the right-propagating wave

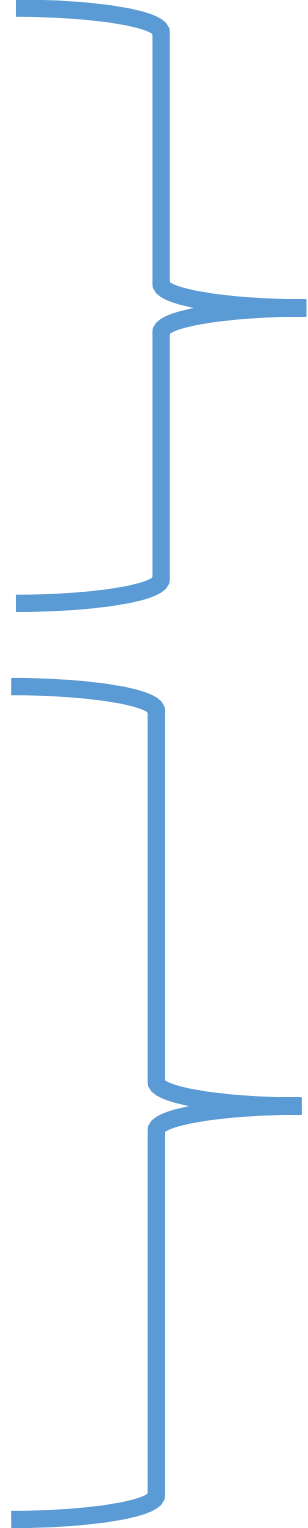
Re{V-}, Im{V-} for the left-propagating wave

- Equivalently, you can use the voltage and current at each point in space for the sum of the left- and right- propagating waves:

Re{V}, Im{V}

Re{I}, Im{I}

$$V(z, t) = V^+ e^{i[\omega t - kz]} + V^- e^{-i[\omega t - kz]}$$
$$I(z, t) = \frac{V^+}{Z_0} e^{i[\omega t - kz]} - \frac{V^-}{Z_0} e^{-i[\omega t - kz]}$$

- 
- Good for scattering calculations
  - Microwave engineering
  - Scattering-mode amplifiers
  
  - Useful for Ohm's Law /
  - Circuit diagrams
  - "Op-amp" mode amplifiers



# Scattering mode / opamp mode

## Scattering-mode amplifiers

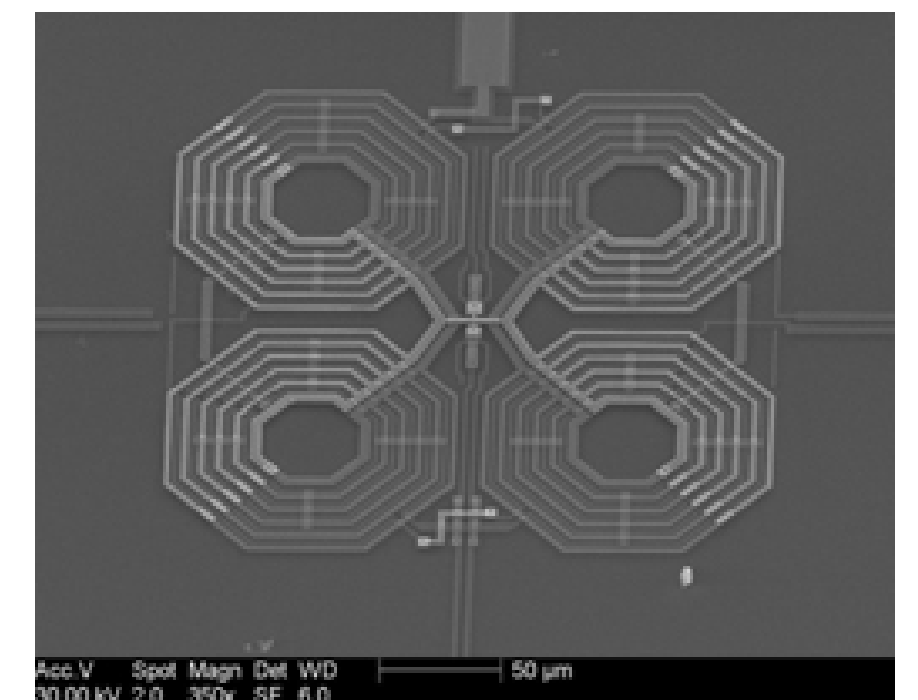
- Amplification of waveforms
- Typical when length scale  $> \lambda$
- Input described by incoming and outgoing waves
- RF amplifiers and HEMTs, Parametric amplifiers



## “Op-amp”-mode amplifiers

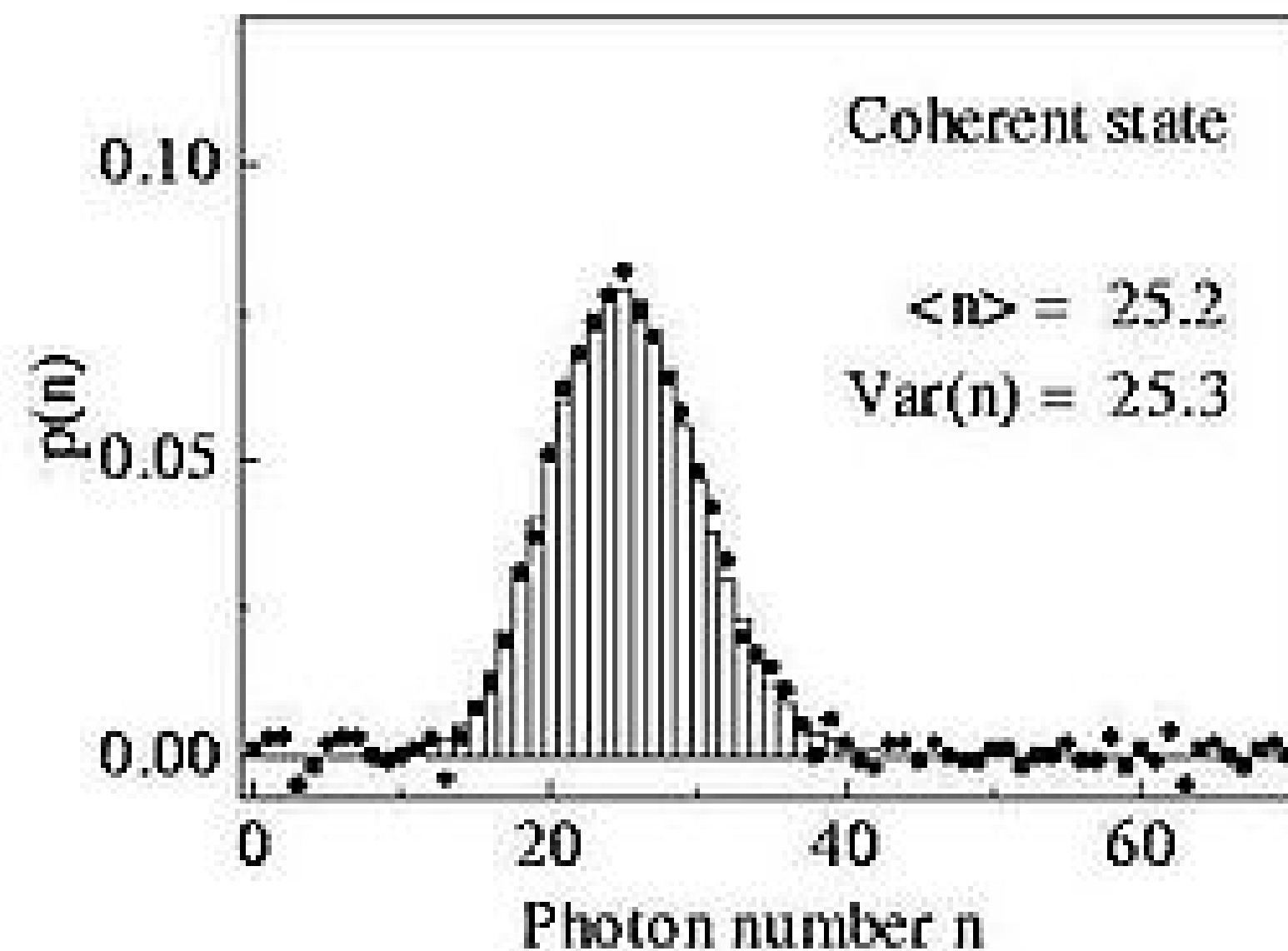
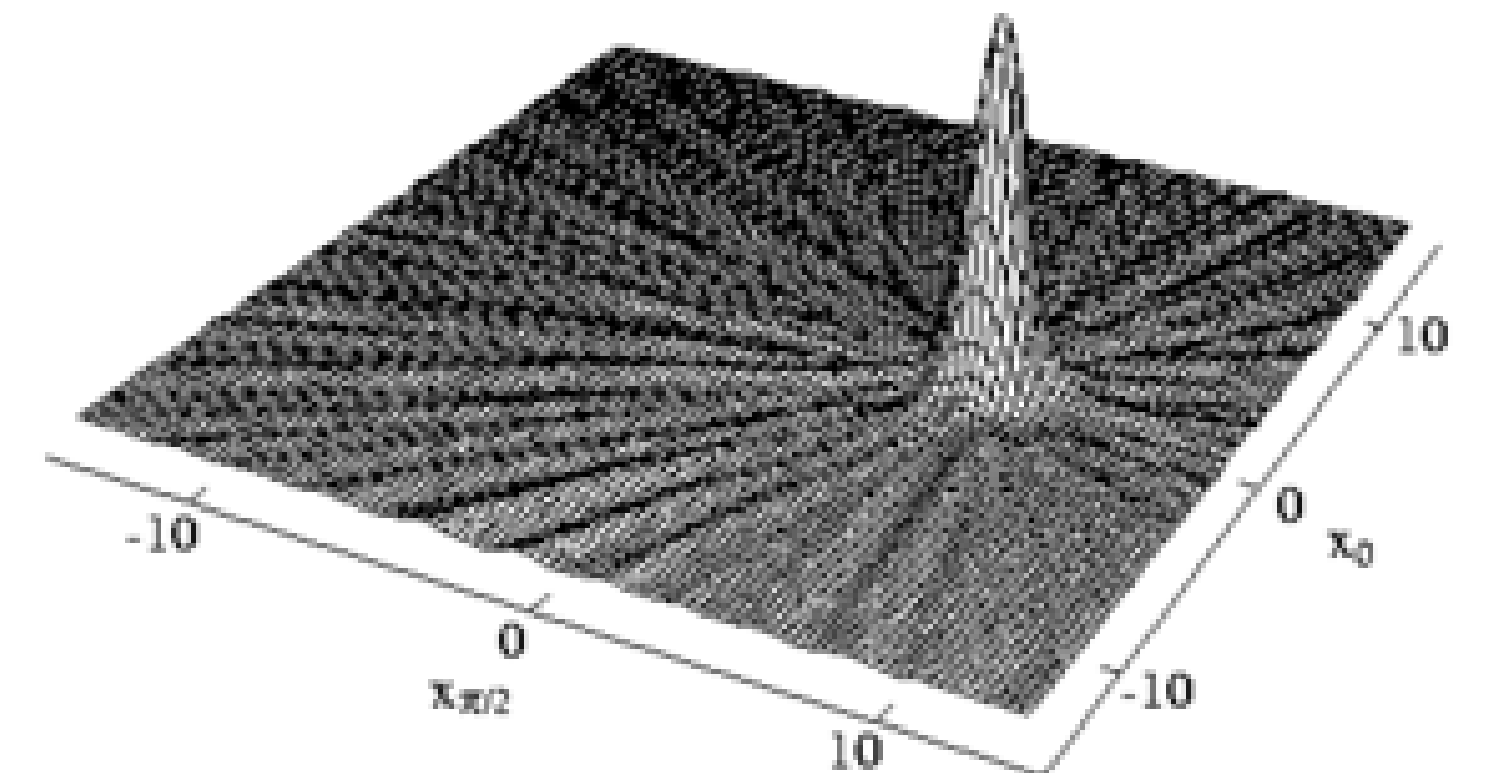
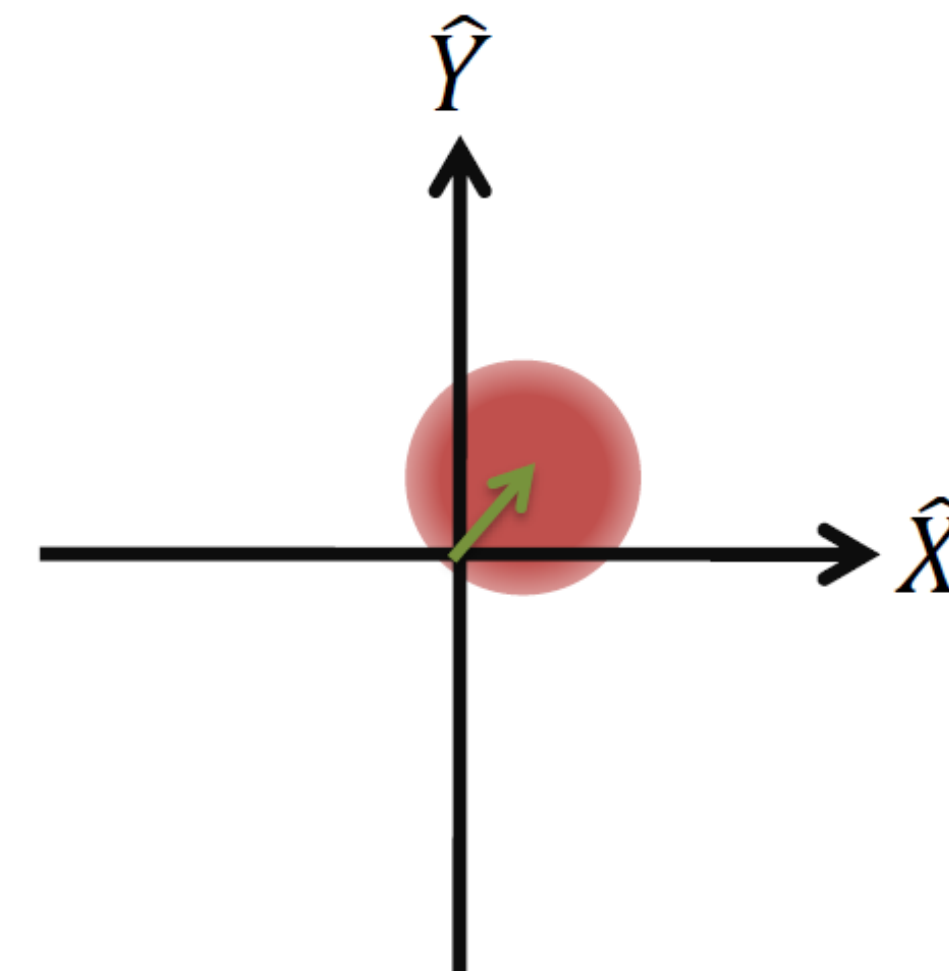
- Amplification of state variable (voltage, current, flux)
- Typical when length scale  $< \lambda$
- Input described by state variable and backaction on input, and tuning coupling
- FETs, bipolar transistors
- dc SQUIDs

Quantum limits are analogous,  
but take different treatment



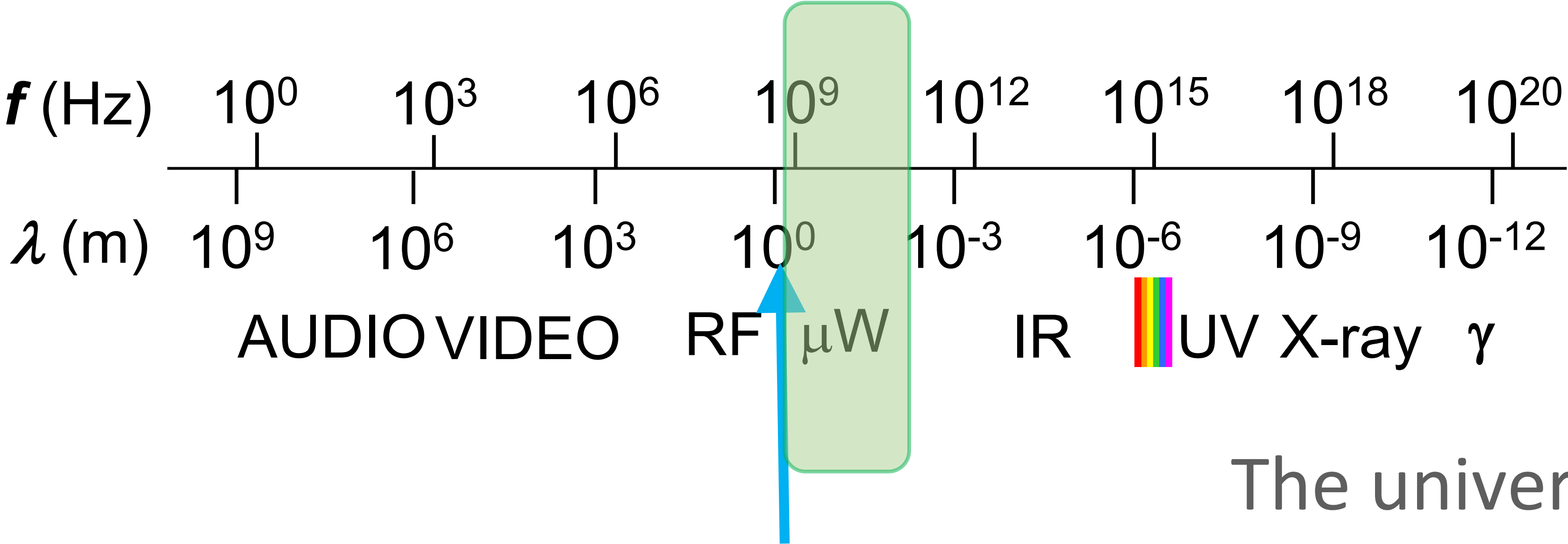
# Coherent states

- A “best approximation” of a sinewave.
- Eigenstate of the annihilation operator
- Uncertainty balanced between  $\hat{X}$  and  $\hat{Y}$



- Coherent state is a superposition of many different photon-number states.
- Eigenstate of photon number is “Fock” state
- Projection noise of coherent onto Fock state -> “shot” noise (another SQL)

Aaron Chou's lecture will address microwave-frequency techniques with qubits, including Fock state preparation



300 MHz  $\sim$  0.015 Kelvin  $\sim$  1 m

300 MHz  $\sim$  human scale  $\sim$  dilution refrigerator temperature

# Sensing electromagnetic signals at low frequency

- Preparing Fock states difficult because of thermal photons
- Op-amp mode amplifiers used because length  $\gg \lambda$
- Many types of amplifiers are traditionally available: FETs, bipolar transistors, SET electrometers. Most are far from any Standard Quantum Limit
- The more traditional amplifier that is often considered “quantum limited” is the dc SQUID, an op-amp mode amplifier
- Mode: flux-to-voltage amplifier (current can be transduced to flux, so can be a current-to-voltage amplifier).

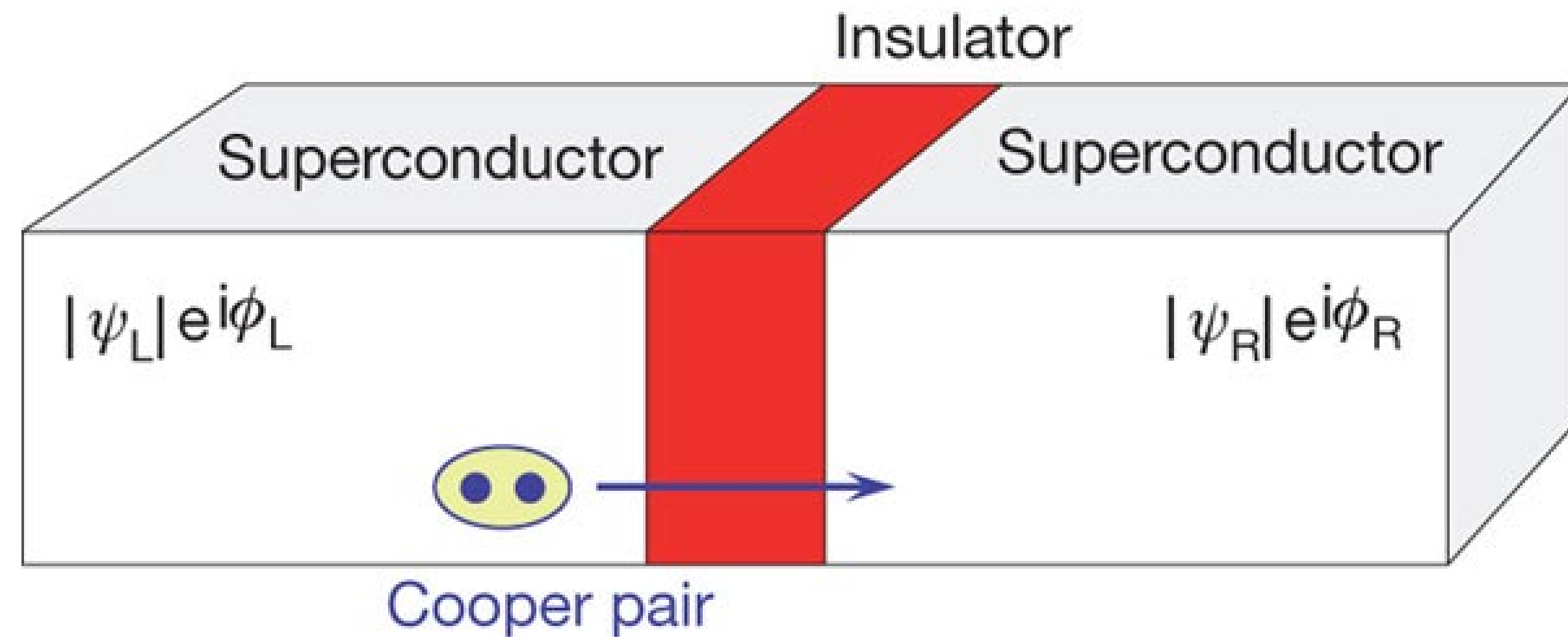
Let's work through the problem of understanding quantum limits in a quantum-limited flux amplifier. *Noise matching and backaction are more central in an op-amp-mode measurement than in a scattering-mode measurement*

*Then we consider newer alternatives to the dc SQUID*

# Josephson Junctions

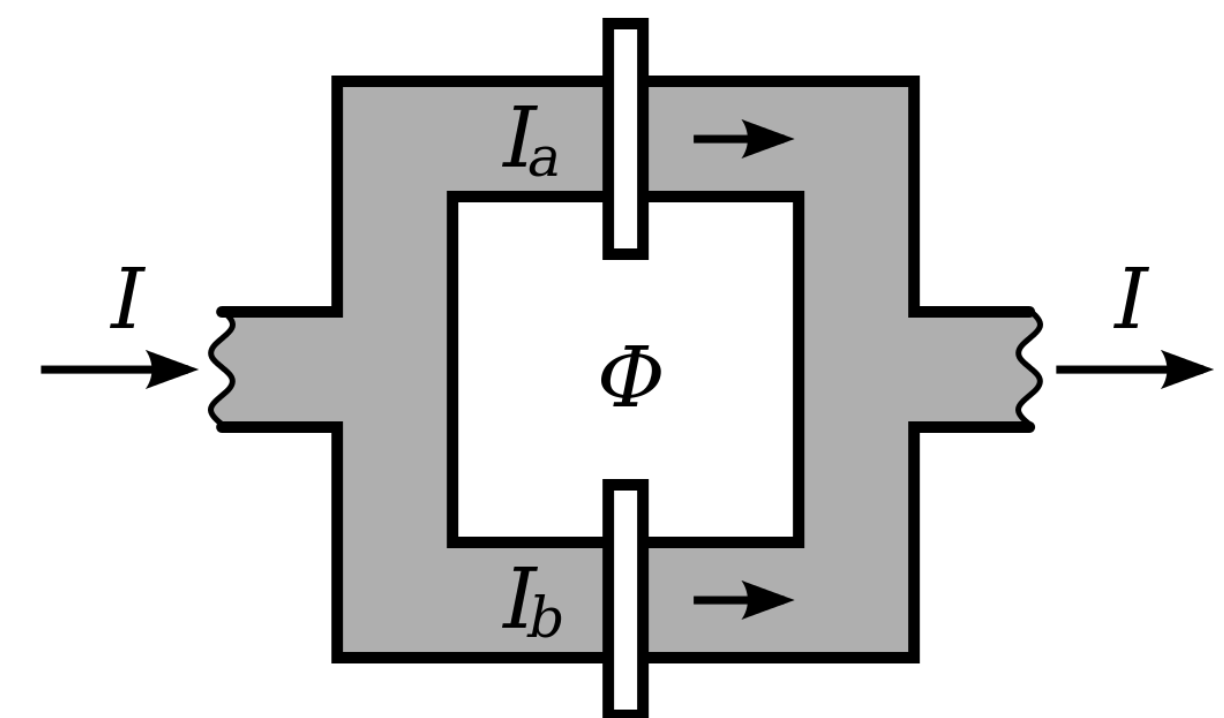


**Brian Josephson**  
**Nobel Prize, 1973**



$$I = I_c \sin(\phi_R - \phi_L)$$

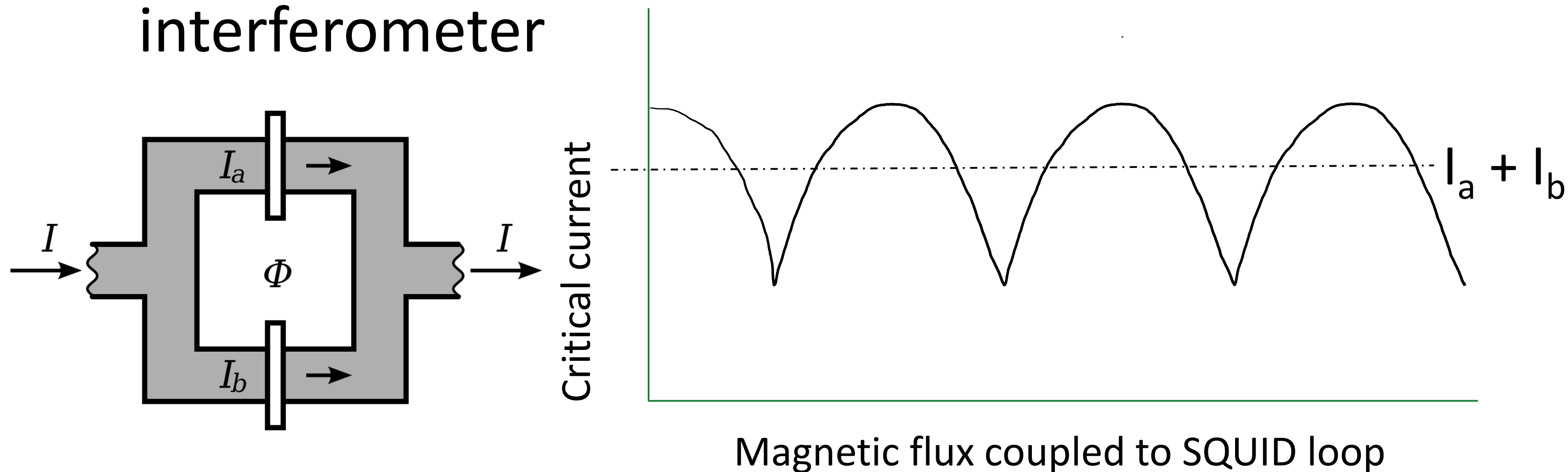
It gets more interesting when you make a loop:  
The magnetic flux modulates the superconducting phase



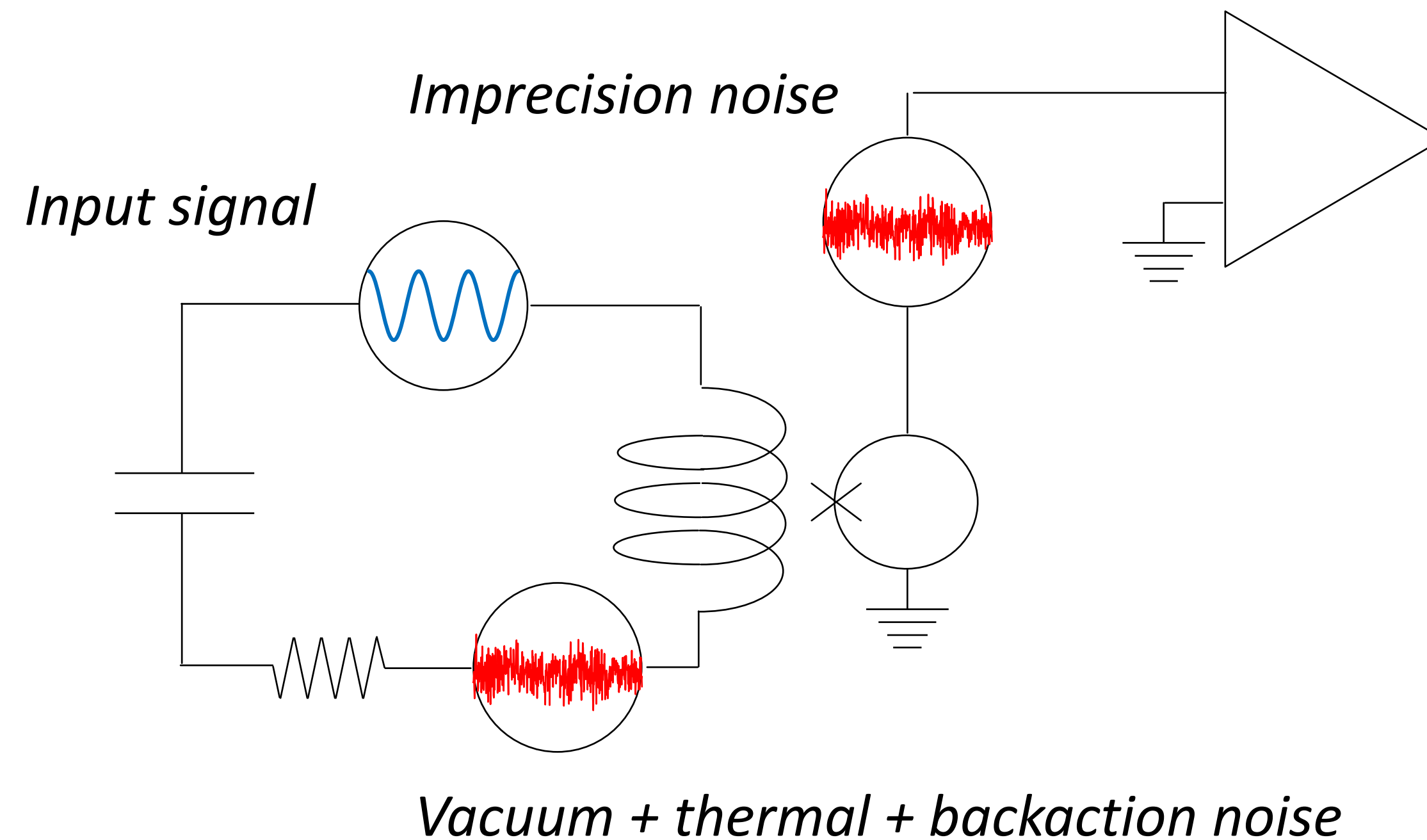


# SQUIDs

- Superconducting Quantum Interference Device (SQUID)
- Invented by Arnold Silver, Ford
- Quantum interference pattern analogous to a two-slit interferometer



# Noise sources in op-amp mode electrical measurement



- 1. Thermal Noise:** set by the resonator's thermal occupation.
- 2. Vacuum Noise:** required by quantum mechanics.
- 3. Amplifier Noise:** composed of *imprecision* and *backaction* noise, can be subject to a Standard Quantum Limit.

# How do you noise match an op-amp mode amplifier?

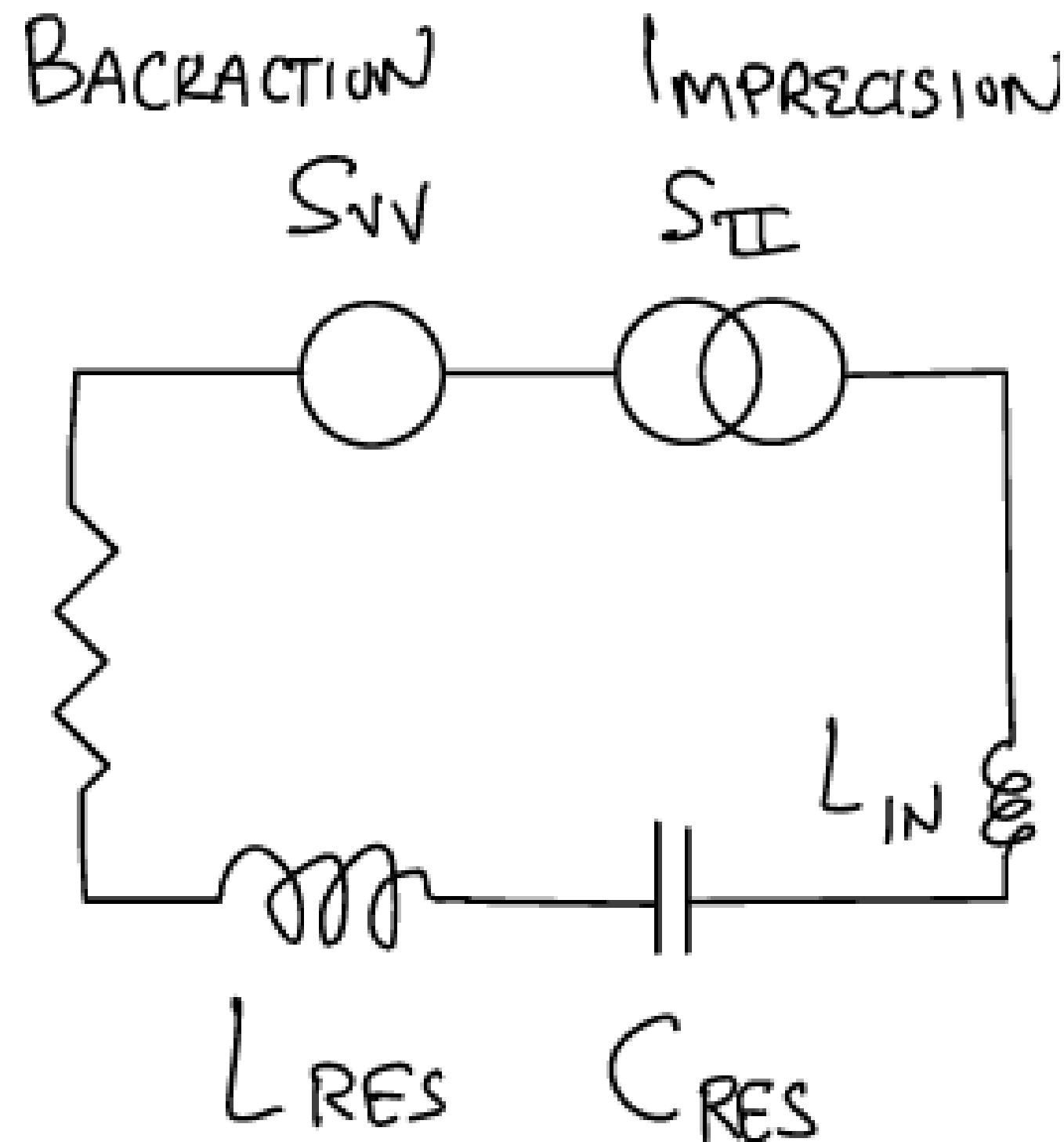
$$\omega_0 = \frac{1}{\sqrt{L_{\text{tot}} C_{\text{res}}}}$$

Our classical noise temperature definition:

$$4k_B T_n \text{Re}\{Z_{\text{res}}\} = \mathcal{S}_{VV} + |Z_{\text{res}}|^2 \mathcal{S}_{II} - 2 \text{Re}\{Z_{\text{res}}^* \mathcal{S}_{IV}\}$$

On resonance  $Z_{\text{res}}(\omega_0) = R.$

$$4k_B T_n(\omega_0) R = \mathcal{S}_{VV} + R^2 \mathcal{S}_{II} - 2R \text{Re}\{\mathcal{S}_{IV}\}.$$



There are other definitions of noise temperature! (e.g. Caves) -> including vacuum noise in zero-temperature source. This is just the amplifier added noise.



# Quantum limit on the added noise of an op-amp mode amplifier

$$\text{Re}\{\mathcal{S}_{IV}\} = 0$$

$$4k_B T_n(\omega_0)R = \mathcal{S}_{VV} + R^2 \mathcal{S}_{II}$$

$$k_B T_{\min} = \frac{1}{2} \sqrt{\mathcal{S}_{VV} \mathcal{S}_{II}}$$

$$k_B T_{\min} = \hbar\omega_0/2$$

Consider the special case of an amplifier without real correlations. This is generally true of a dc SQUID flux amplifier.

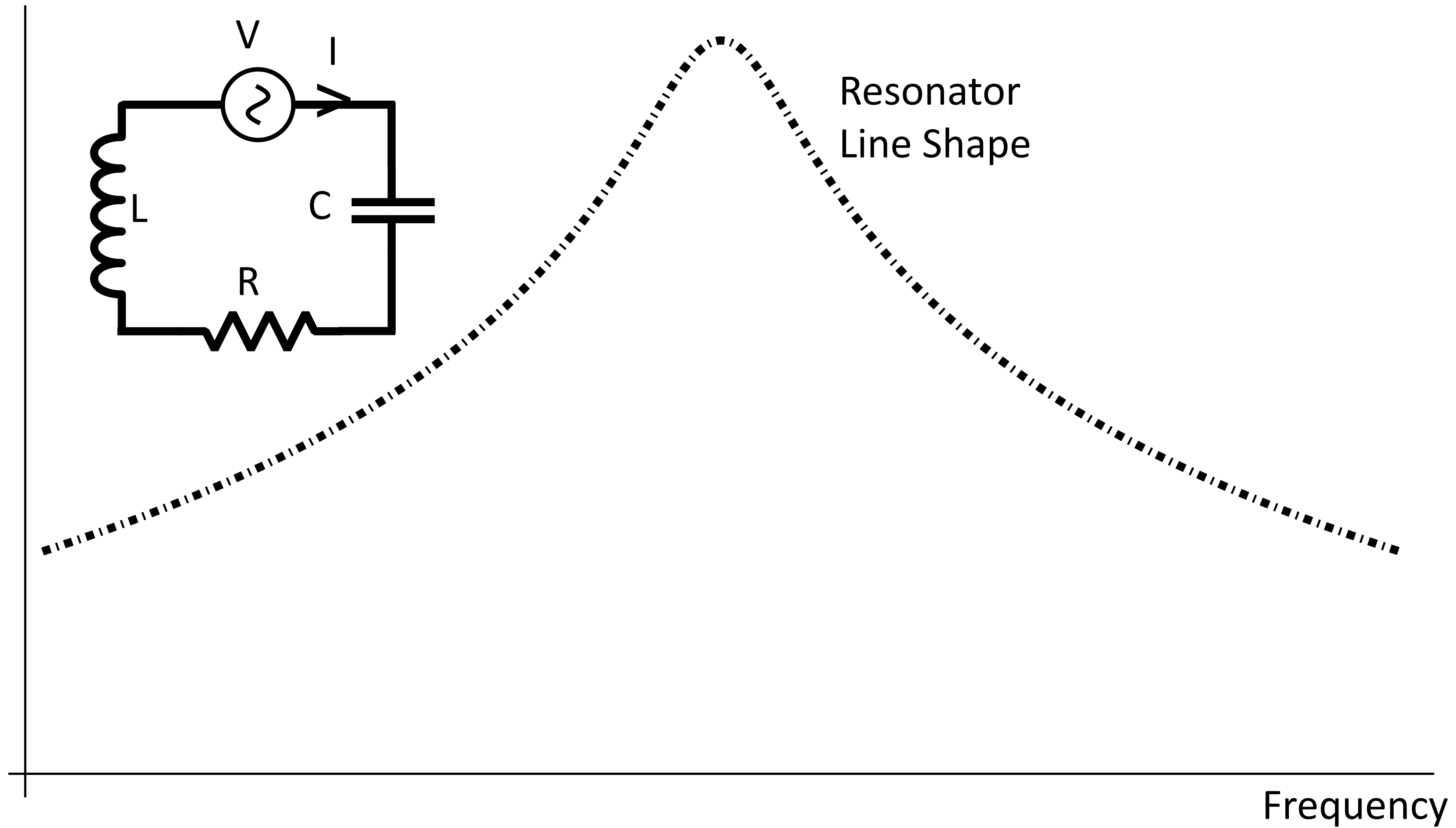
Noise match

Noise matched  $\rightarrow$  minimum noise temperature

Quantum limit on noise temperature (Devoret / Schoelkopf, not Caves' definition).

# Response of low-frequency circuit

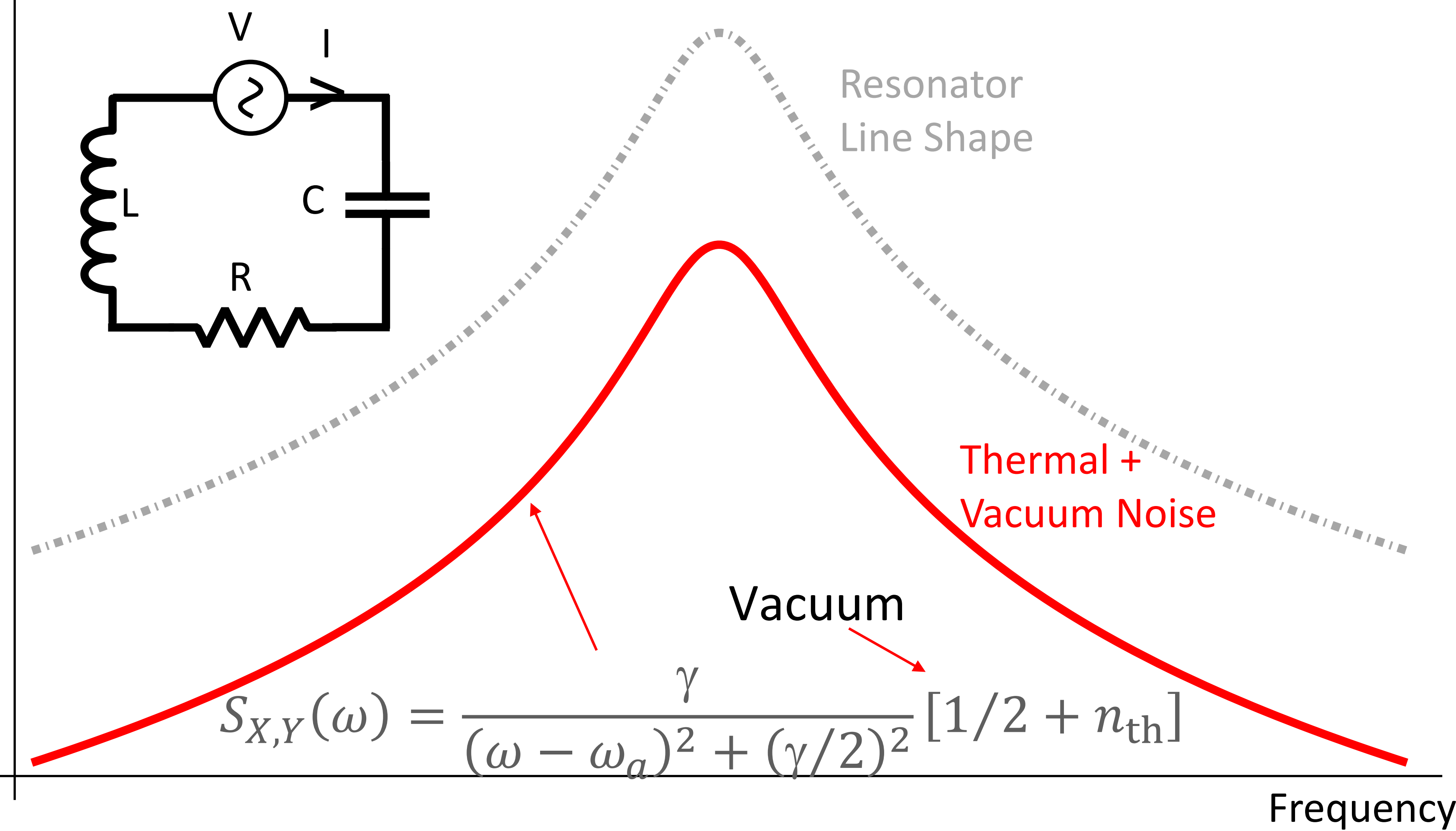
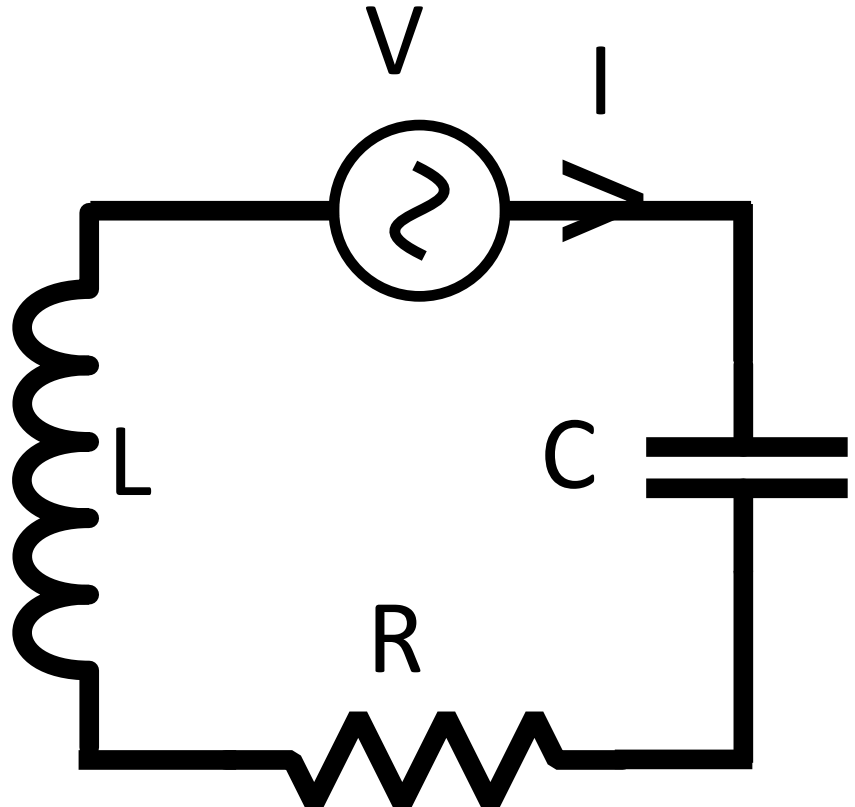
Current  
Response



There is more to noise matching than on-resonance noise!

# Noise in LF circuit

Current Response

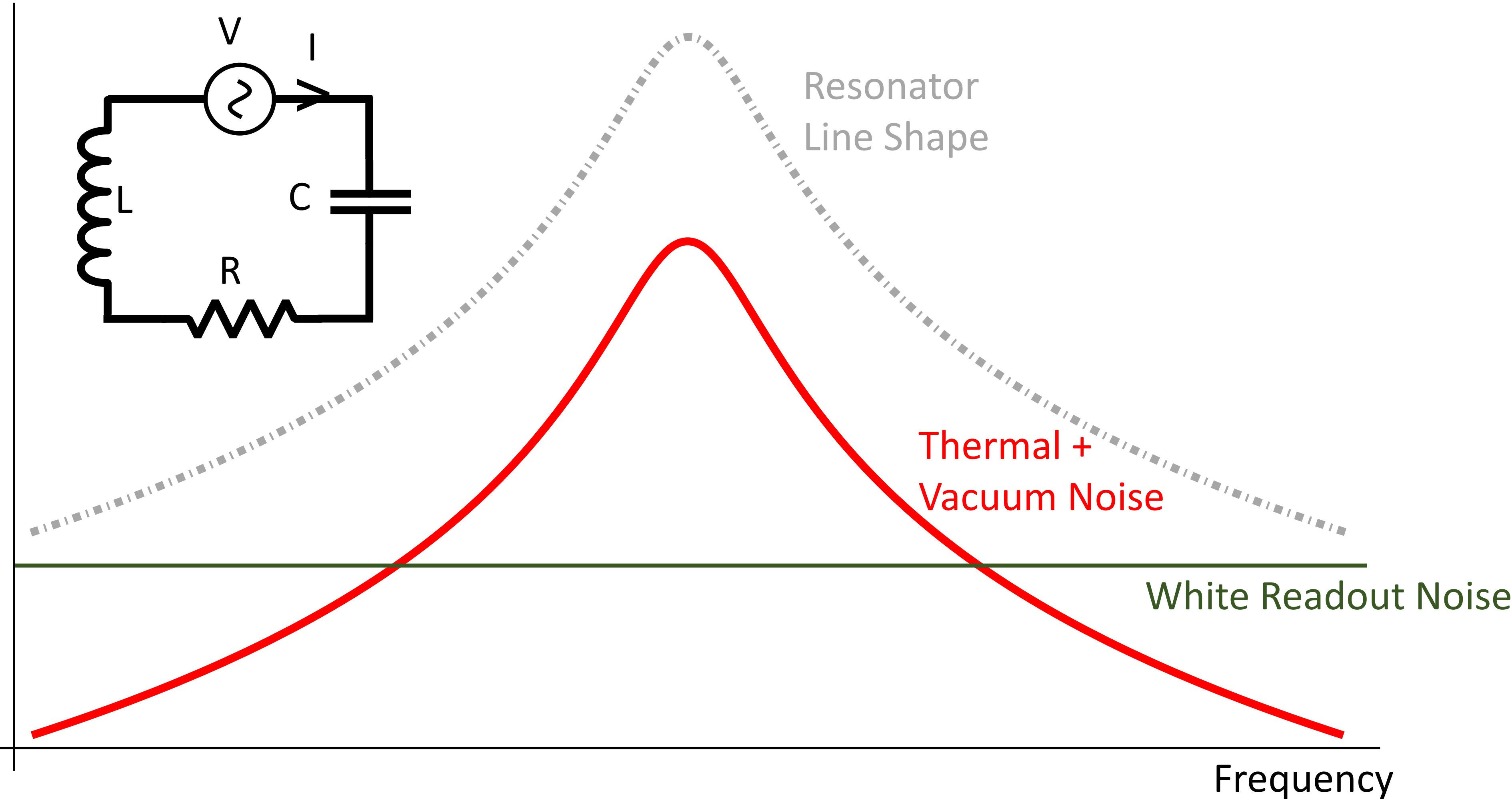


$$S_{X,Y}(\omega) = \frac{\gamma}{(\omega - \omega_a)^2 + (\gamma/2)^2} [1/2 + n_{th}]$$

Thermal occupation  $n_{th} = \left( \exp \left[ \frac{\hbar\omega_a}{k_B T} \right] - 1 \right)^{-1}$

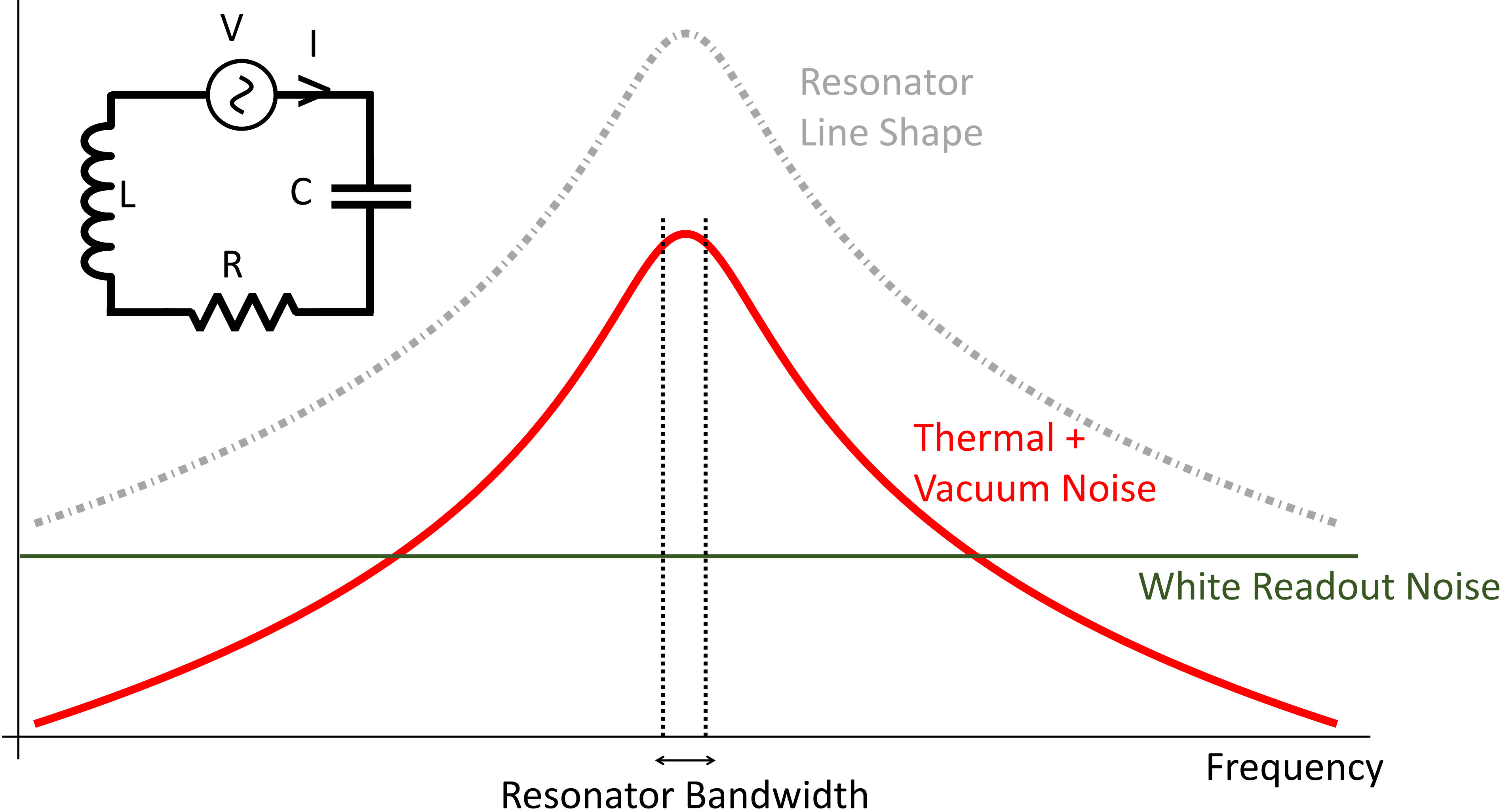
# Noise in LF circuit

Current Response



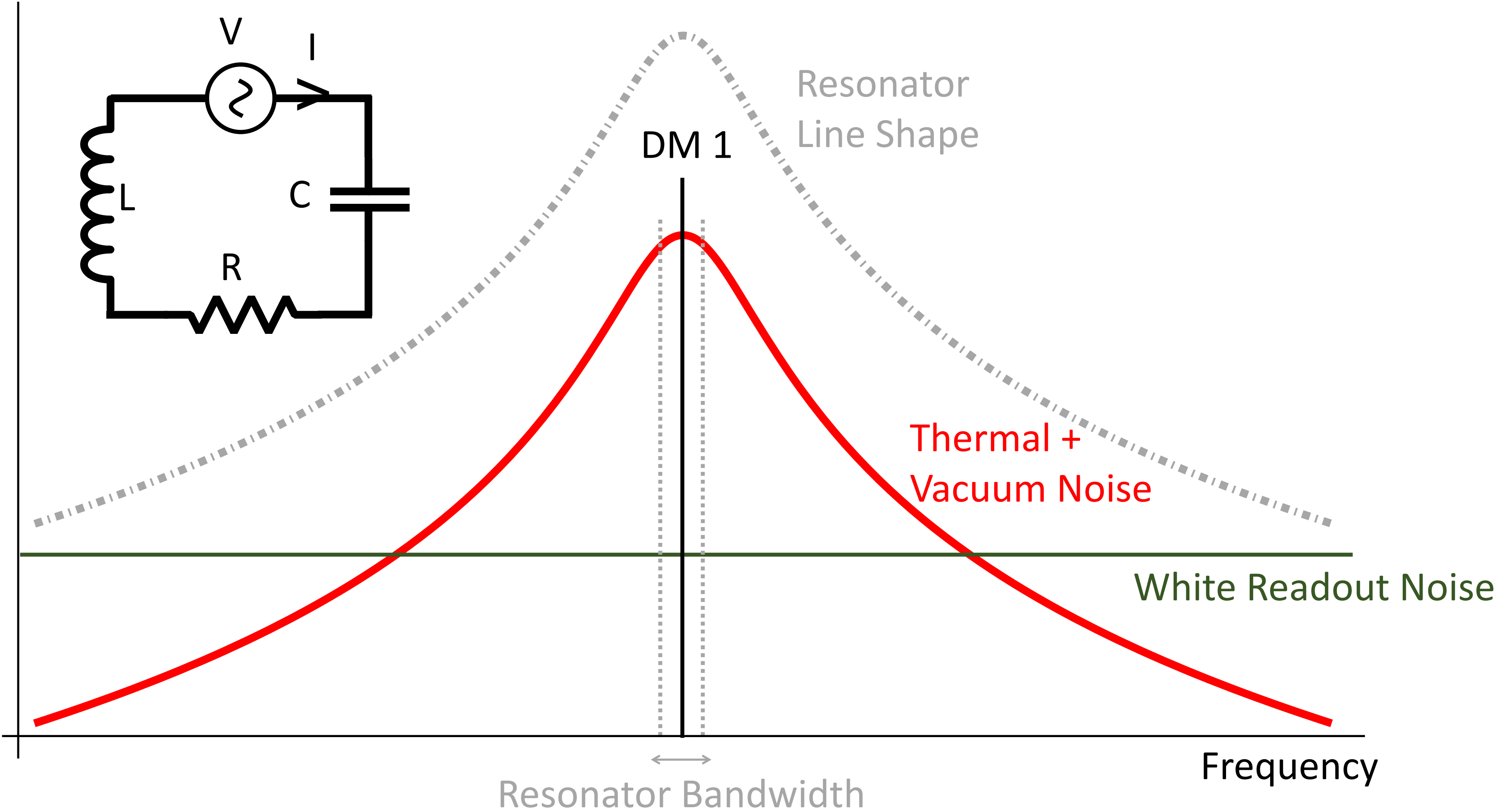
# Noise in LF circuit

Current Response

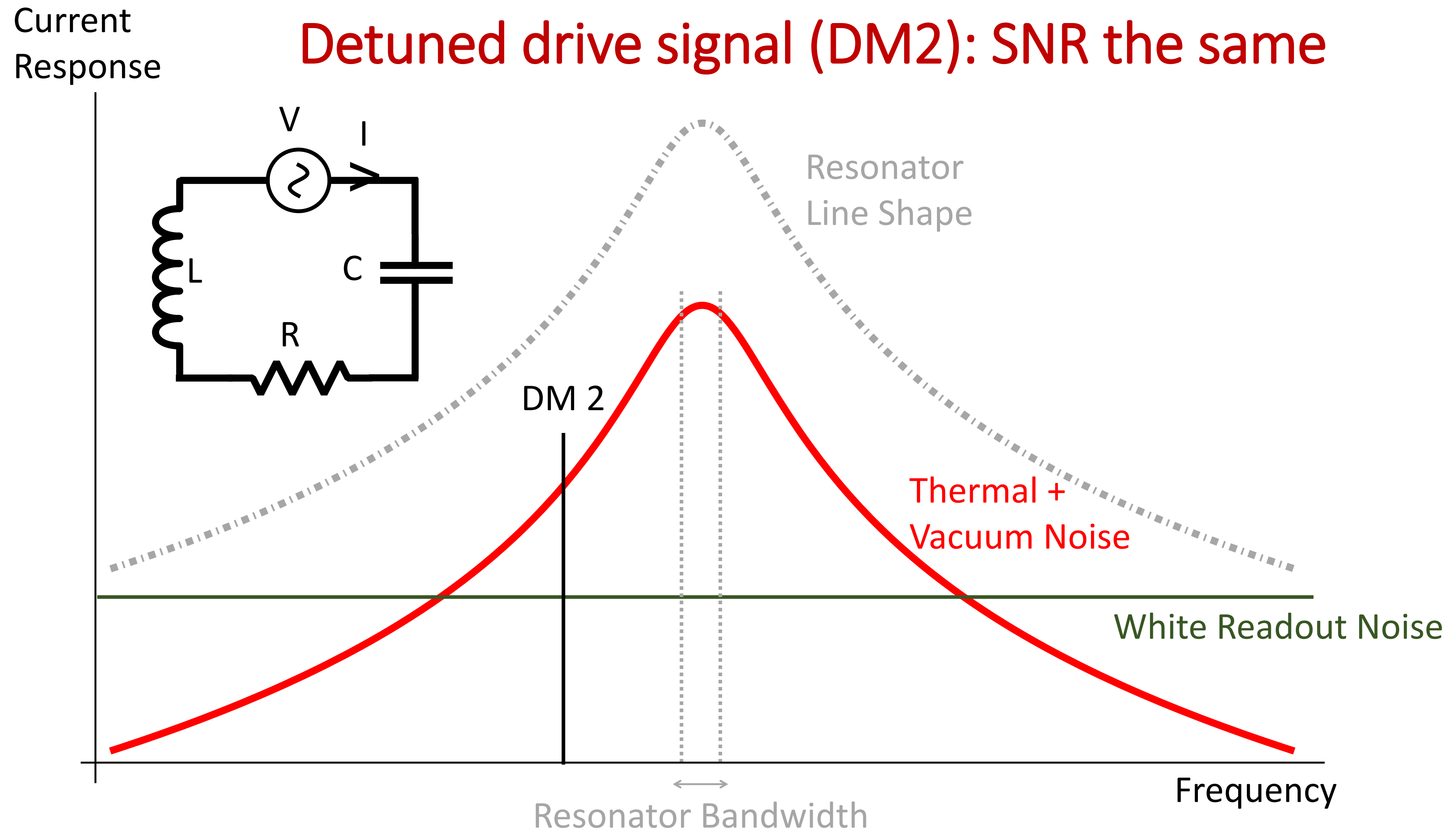


# Tuned to drive signal (DM1): max response

Current Response



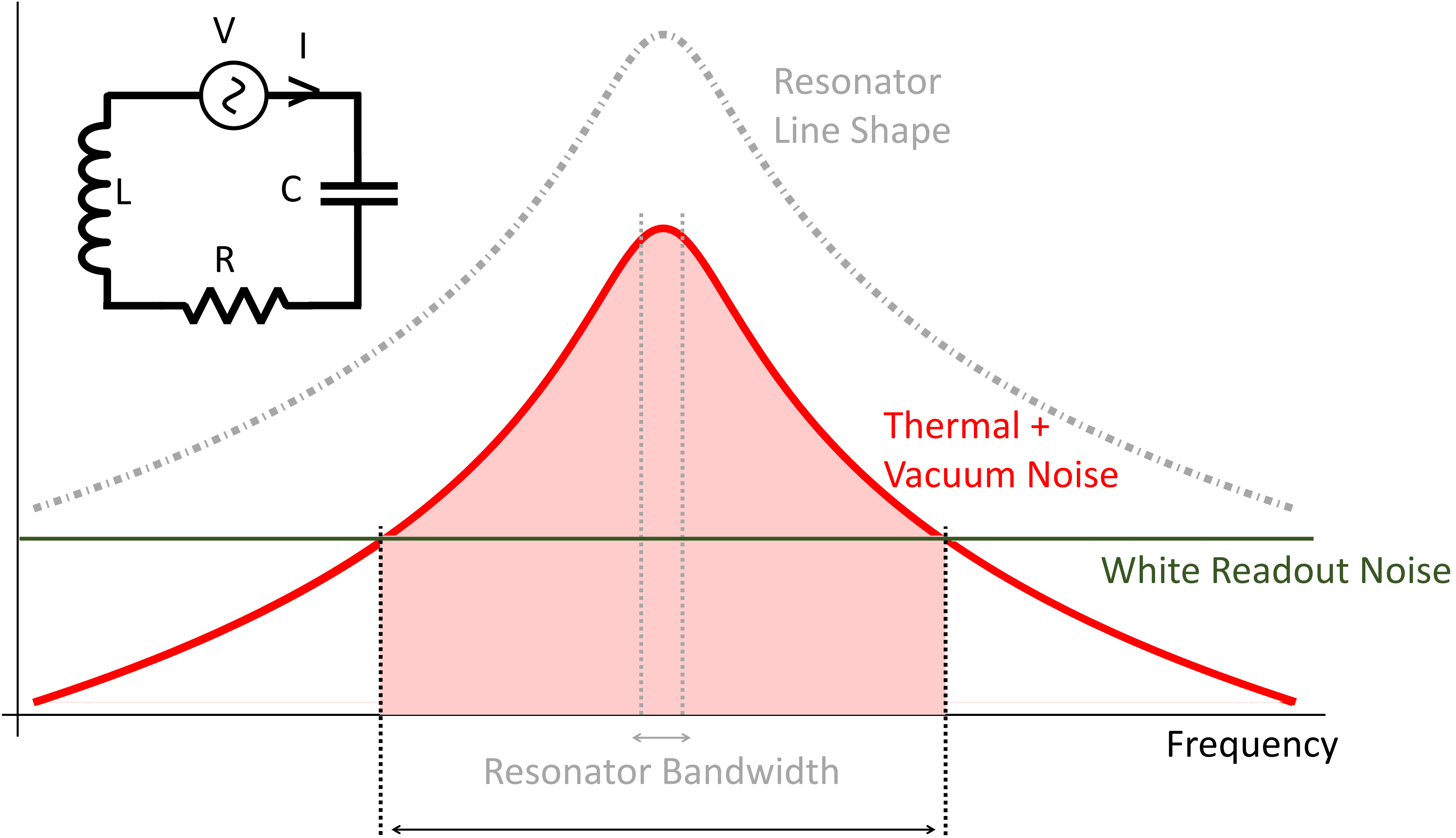
# Detuned drive signal (DM2): SNR the same



- SNR not degraded when readout subdominant to thermal noise

# SNR constant over sensitivity BW

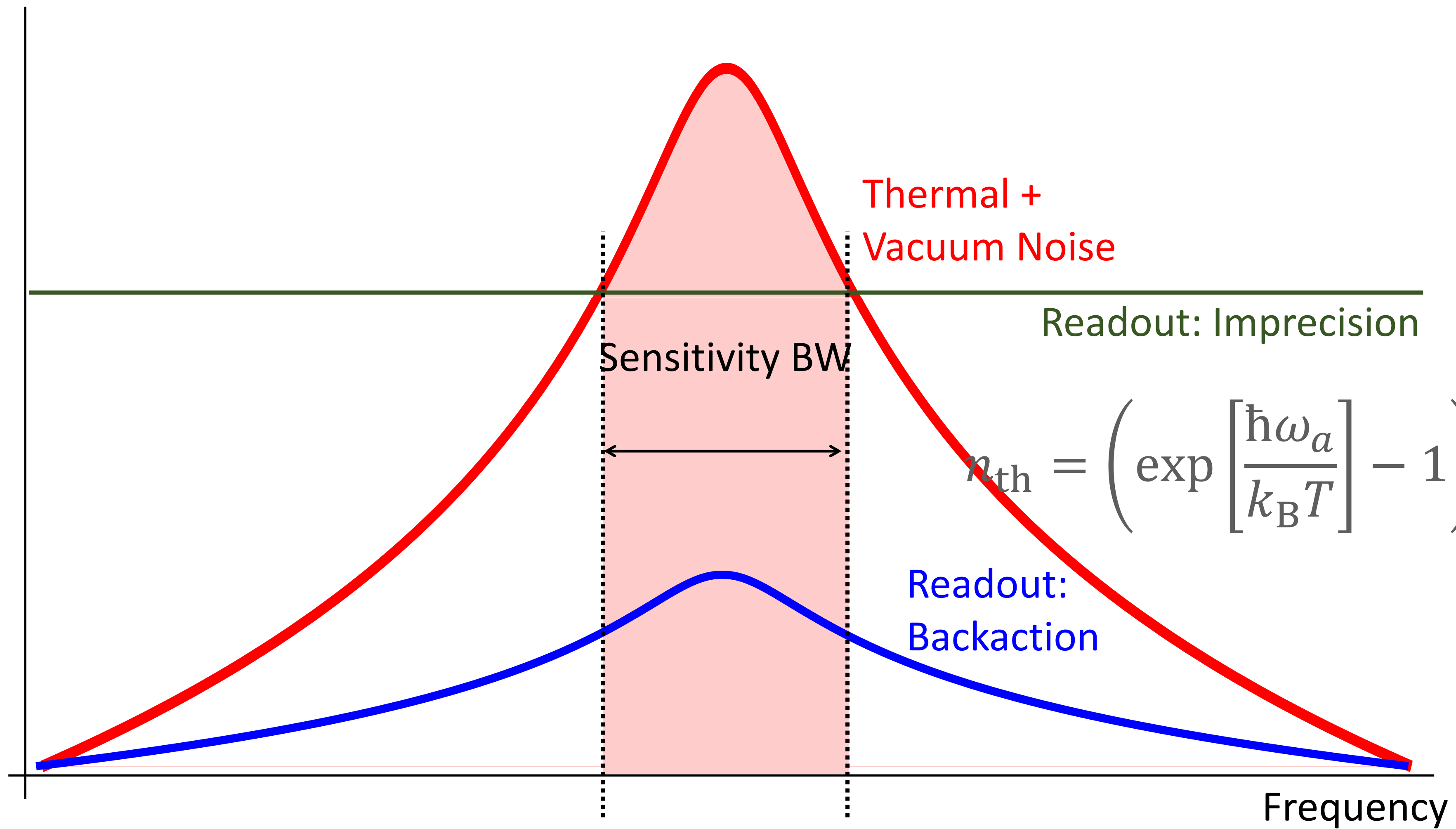
Current Response





# Two contributions to readout noise

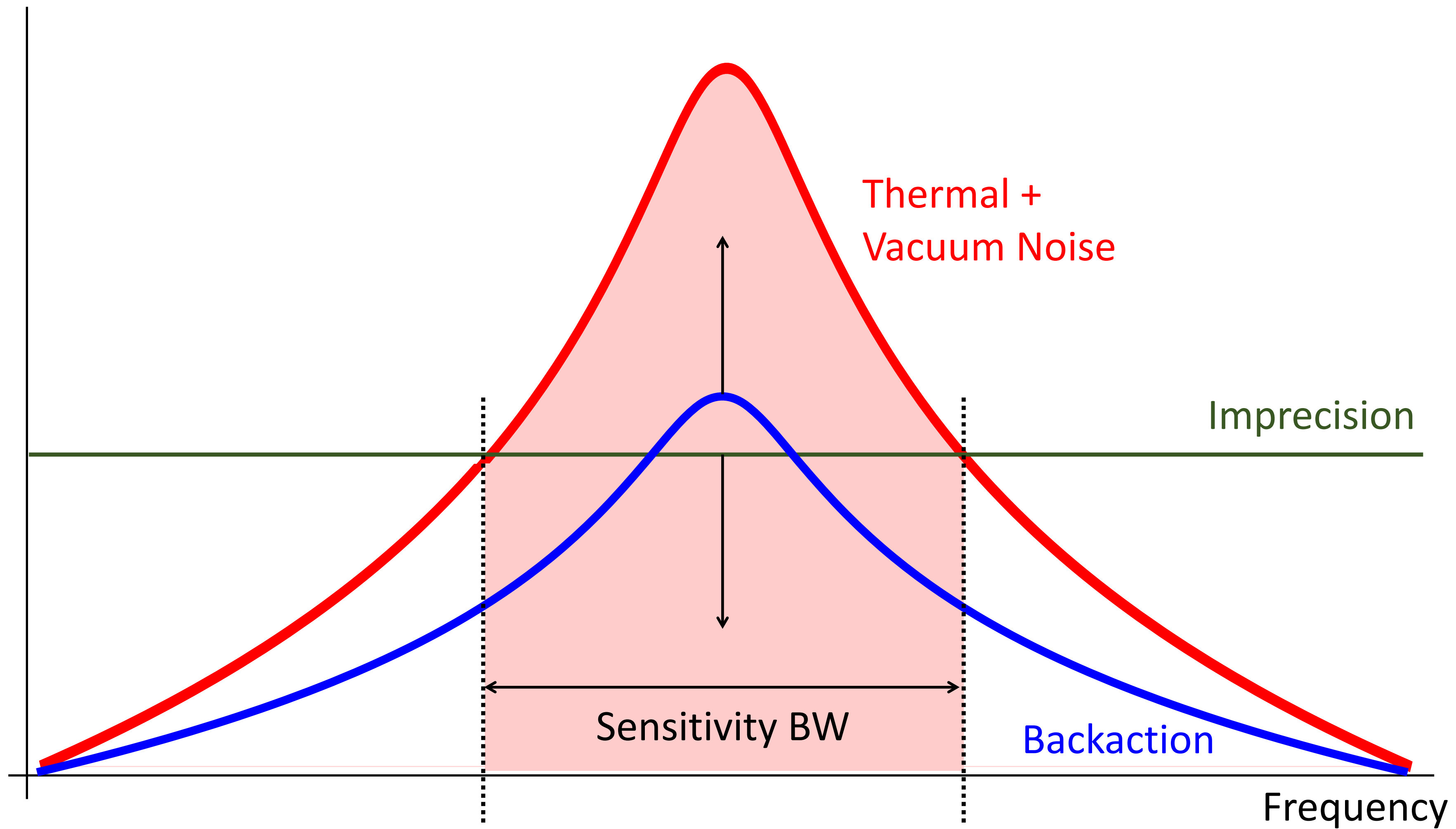
Current  
Response



$$S_{X,Y}(\omega) = \frac{\gamma}{(\omega - \omega_a)^2 + (\gamma/2)^2} [1/2 + n_{\text{th}} + n_{\text{BA}}] + S_{\text{IMP}}(\omega)$$

# Increased coupling: larger sensitivity bandwidth

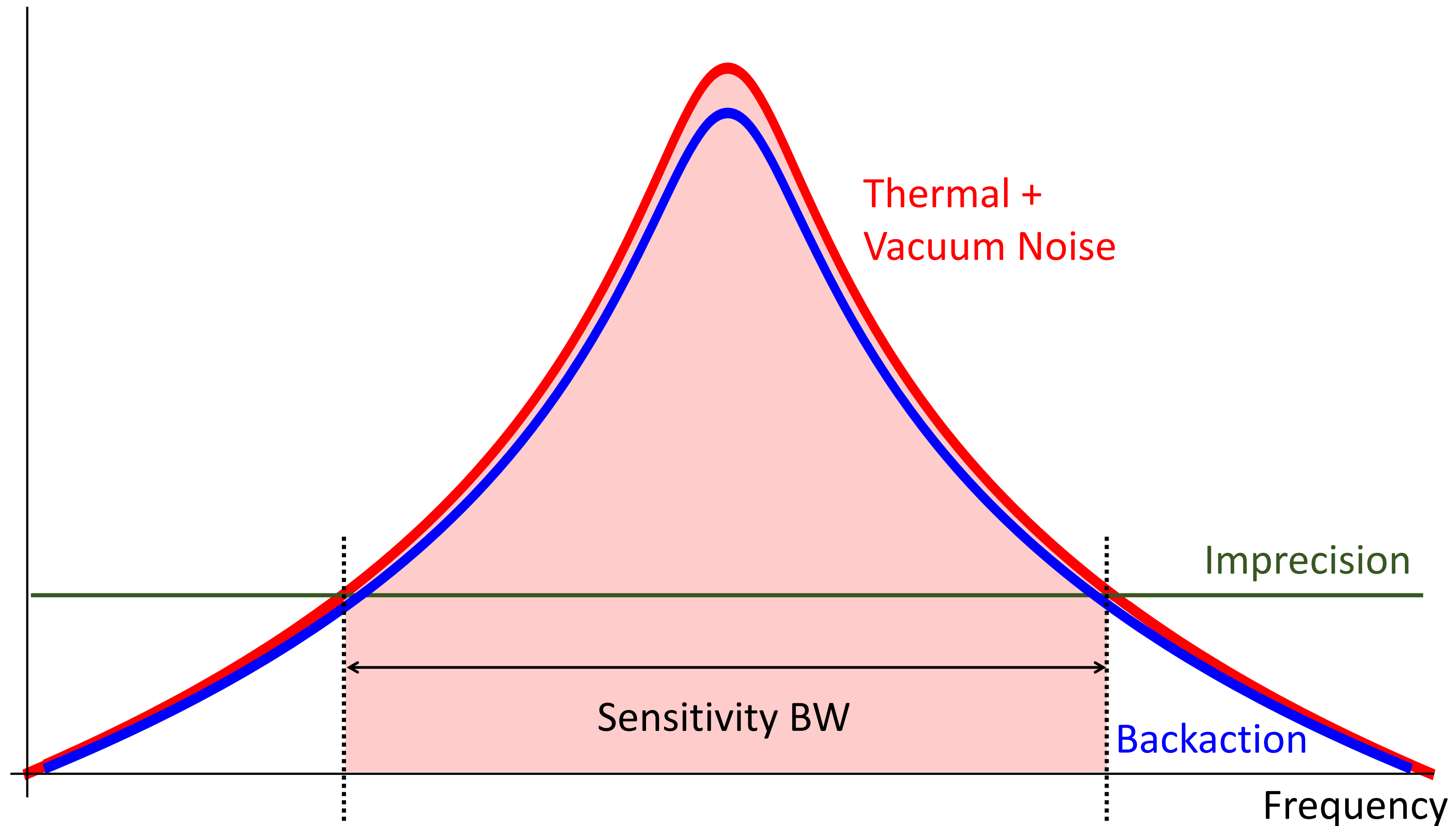
Current  
Response



$$S_{X,Y}(\omega) = \frac{\gamma}{(\omega - \omega_a)^2 + (\gamma/2)^2} [1/2 + n_{\text{th}} + n_{\text{BA}}] + S_{\text{IMP}}(\omega)$$

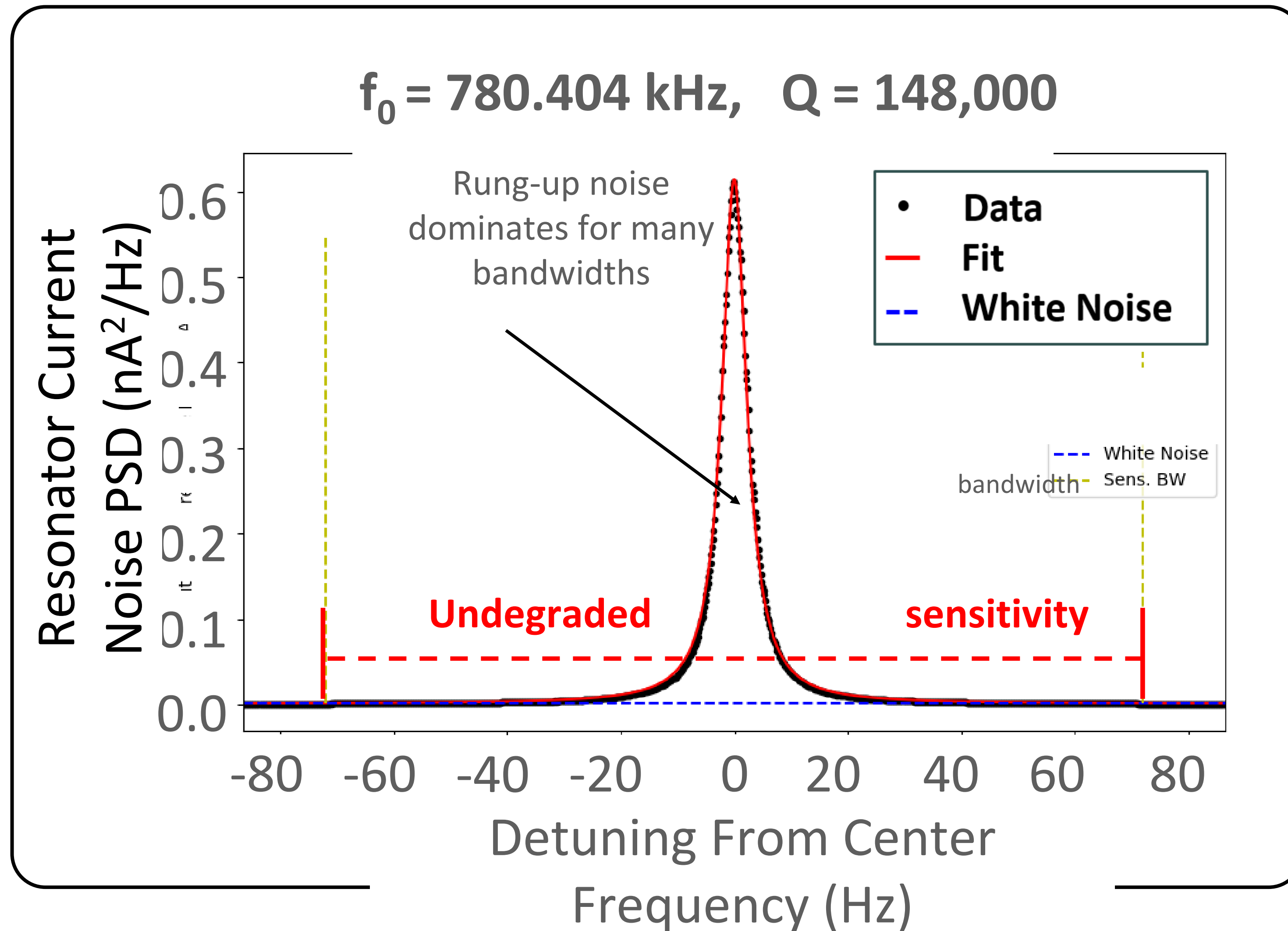
# Optimized coupling: strong backaction

Current  
Response



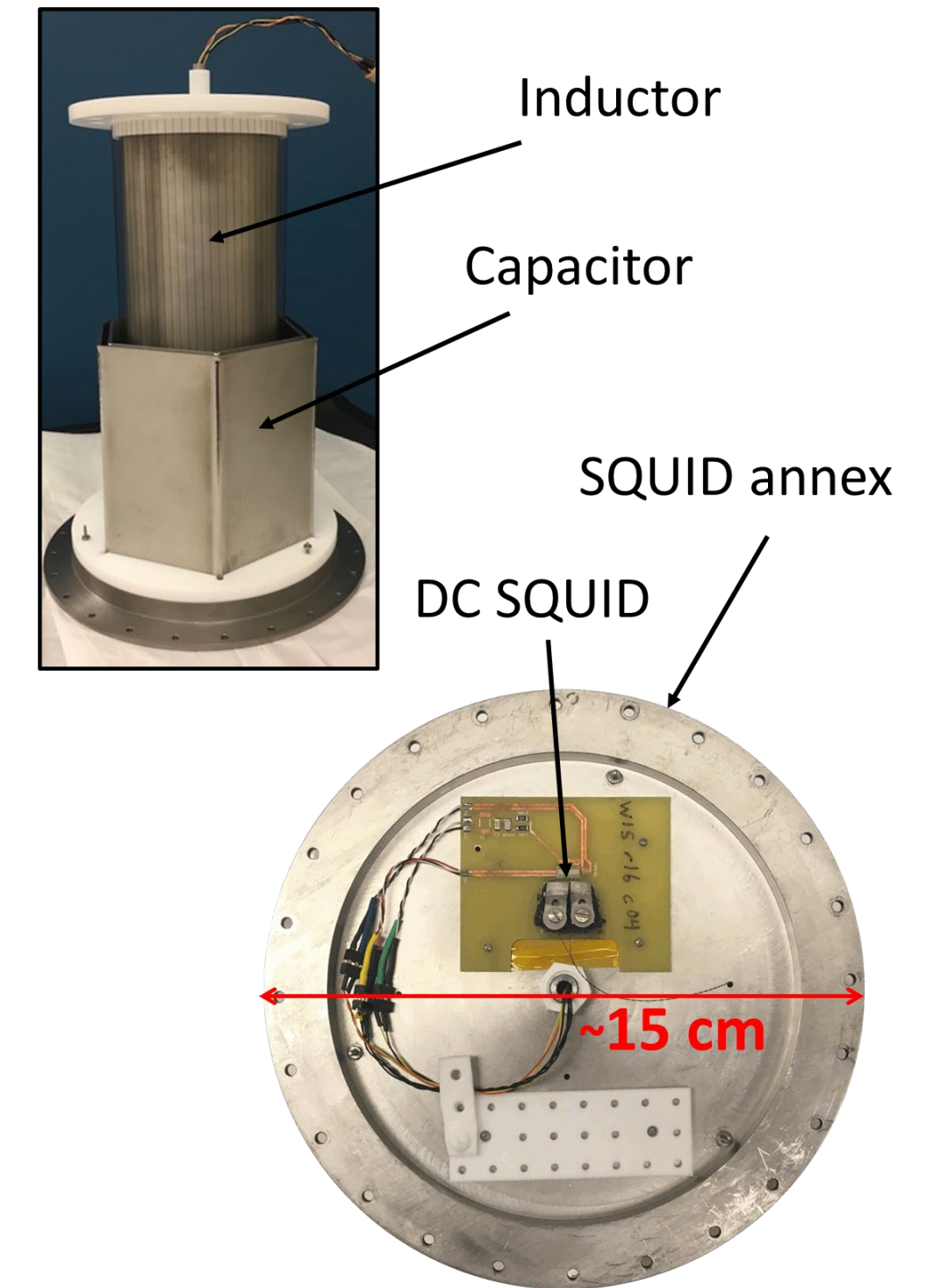
$$S_{X,Y}(\omega) = \frac{\gamma}{(\omega - \omega_a)^2 + (\gamma/2)^2} [1/2 + n_{\text{th}} + n_{\text{BA}}] + S_{\text{IMP}}(\omega)$$

# Example of optimizing noise match for bandwidth with a dc SQUID



Resonator bandwidth:  $\sim 5$  Hz  
Sensitivity bandwidth:  $\sim 150$  Hz

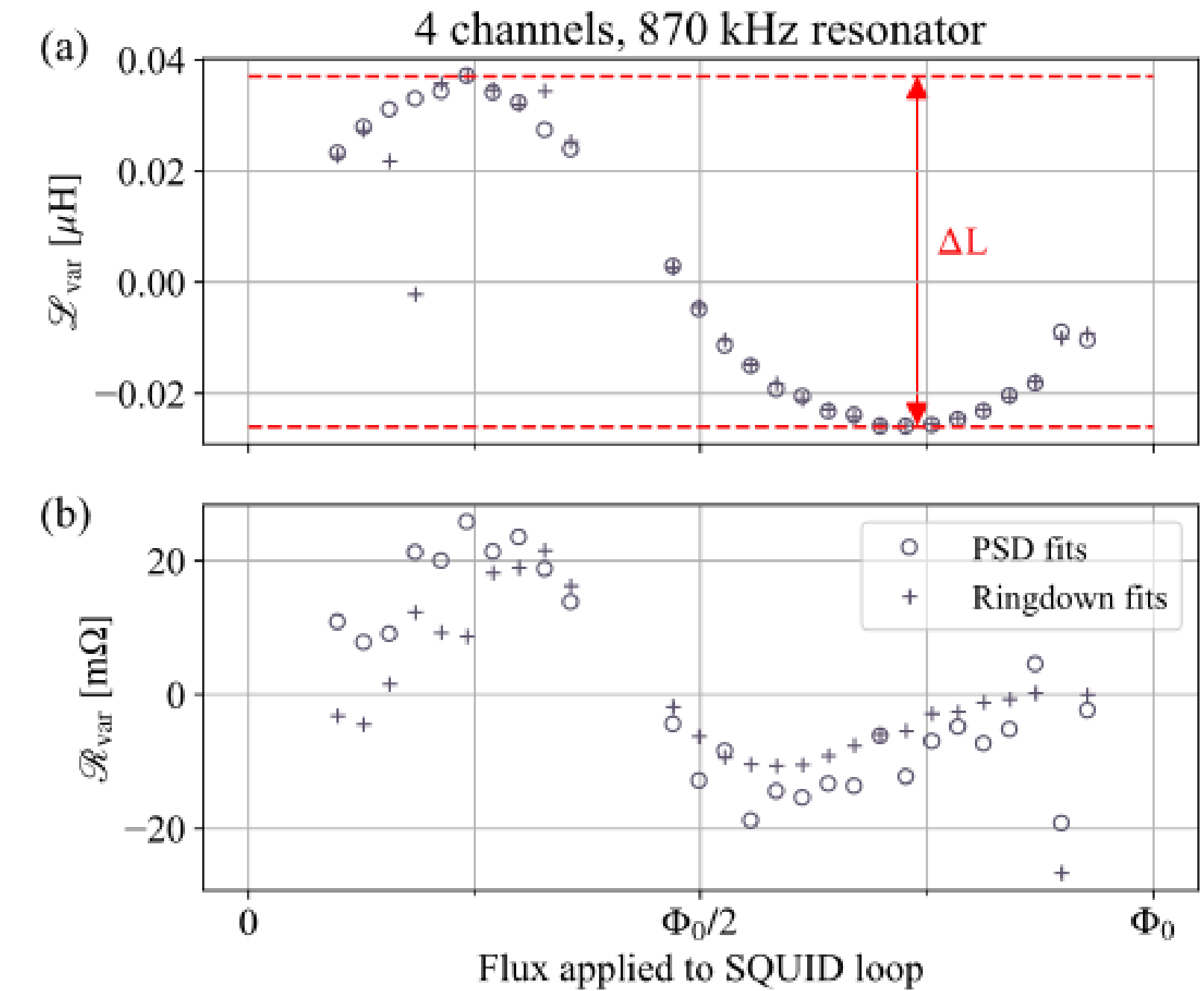
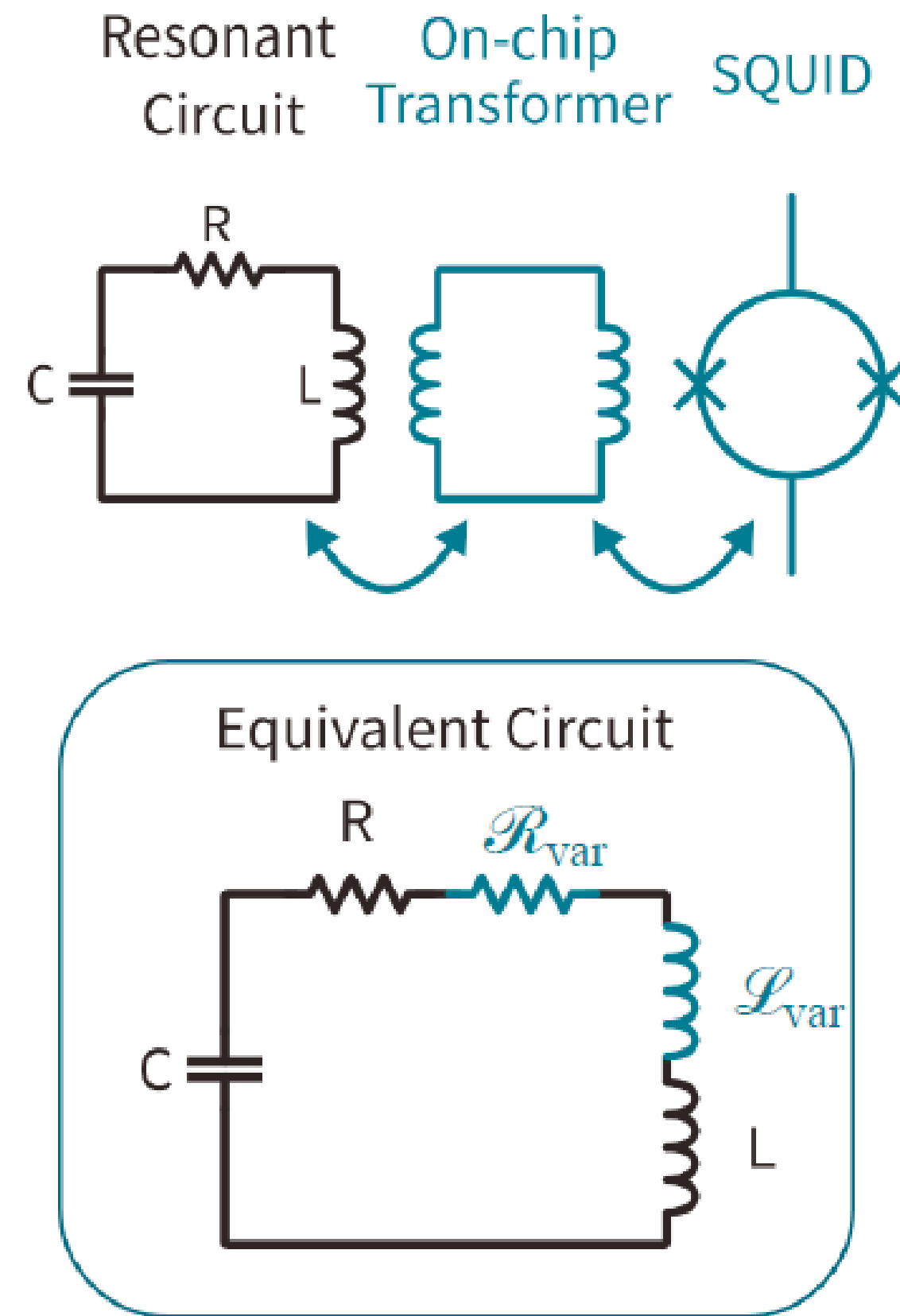
## DMRadio Pathfinder



With optimized noise match, sensitivity bandwidth 30x larger than resonator bandwidth, with negligible degradation on resonance

# Other dc SQUID issues

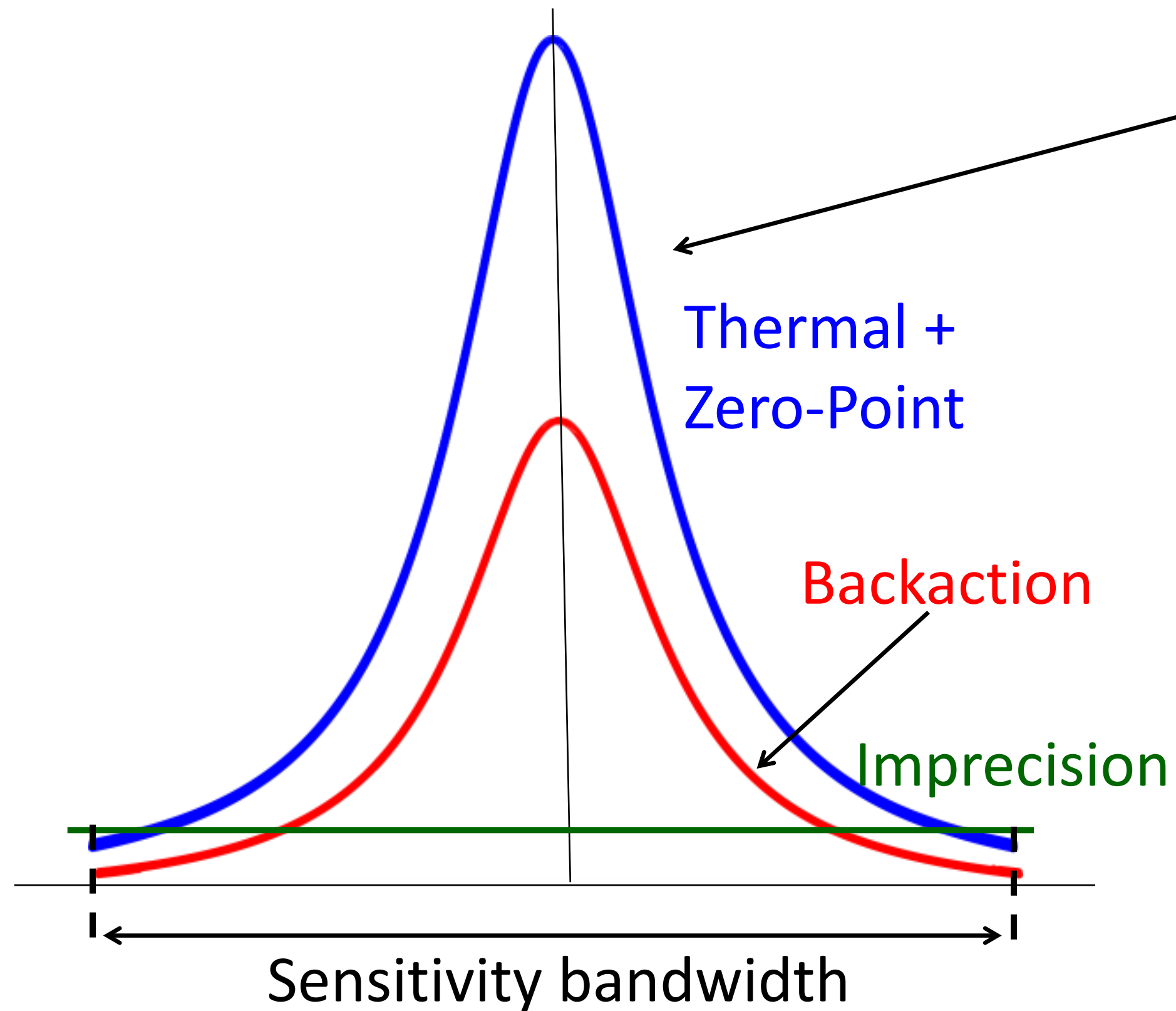
- Resistive shunts mean that significant power can be dissipated.
- Significant damping shifts frequency, degrades Q of input circuit.
- Energy sensitivity is limited by design, noise impedance is fixed.
- Challenging to truly thermalize below  $\sim 100\text{mK}$ .



	$\Delta L$	$\Delta R$
<b>870 kHz resonator</b>		
1 channel	14.4 nH	48.9 m $\Omega$
4 channels	63.0 nH	44.9 m $\Omega$
<b>630 kHz resonator</b>		
2 channels	42.9 nH	25.4 m $\Omega$
4 channels	63.1 nH	17.8 m $\Omega$

E. C. van Assendelft, *et al.* IEEE Transactions on Applied Superconductivity (2023).

# “Click” photon counting is not useful when $hf \ll k_B T$



- $\sqrt{N}$  thermal fluctuations in the number of resonator photons
  - Sensitivity not improved by photon “click” counting
  - Preparing a high- $N$  Fock state difficult
- **Need other techniques**

- ***Backaction evasion*** to reduce both imprecision and backaction noise to below the standard quantum limit
- Or squeeze thermal noise

# Better science thru evading the Standard Quantum Limit

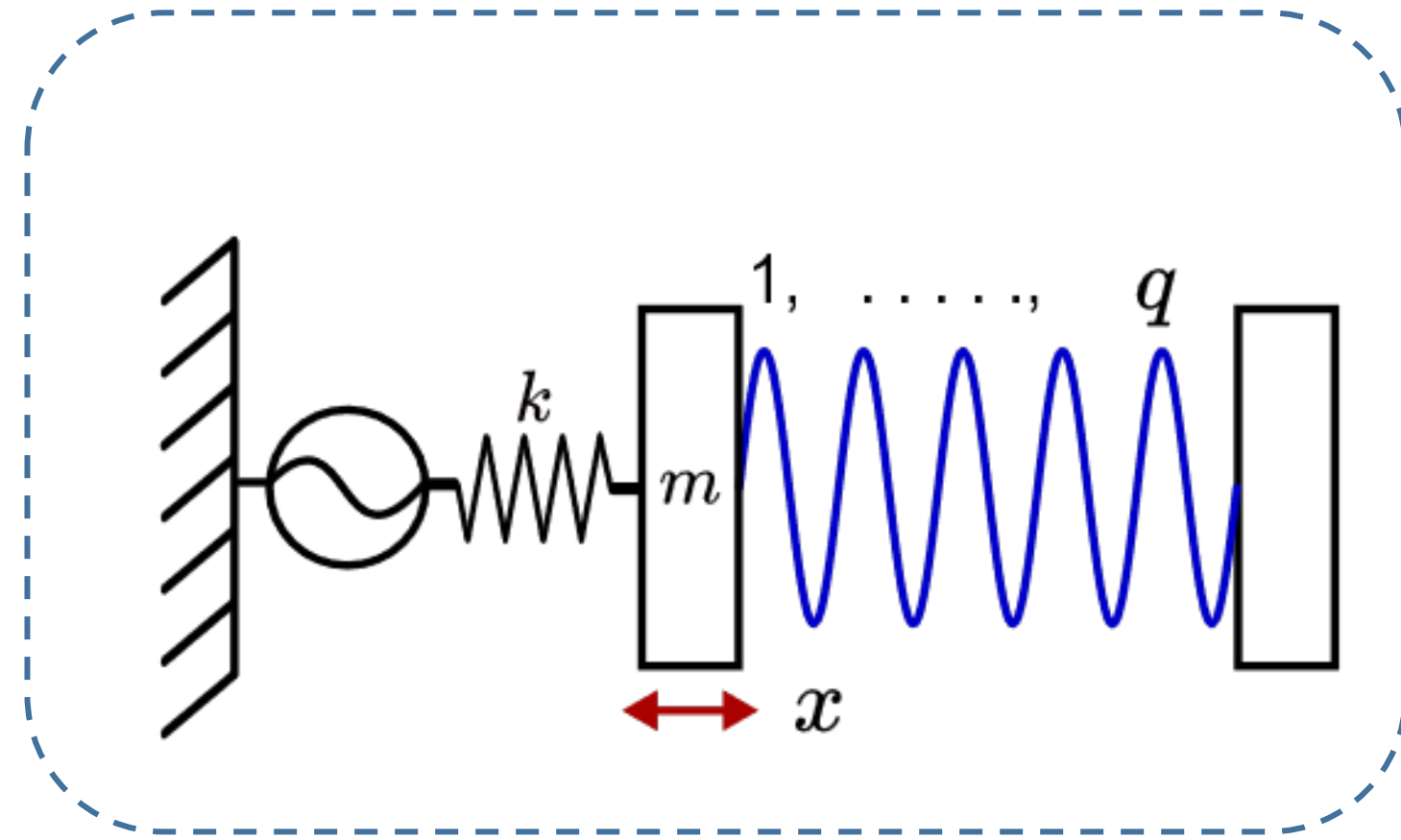
LIGO Livingston



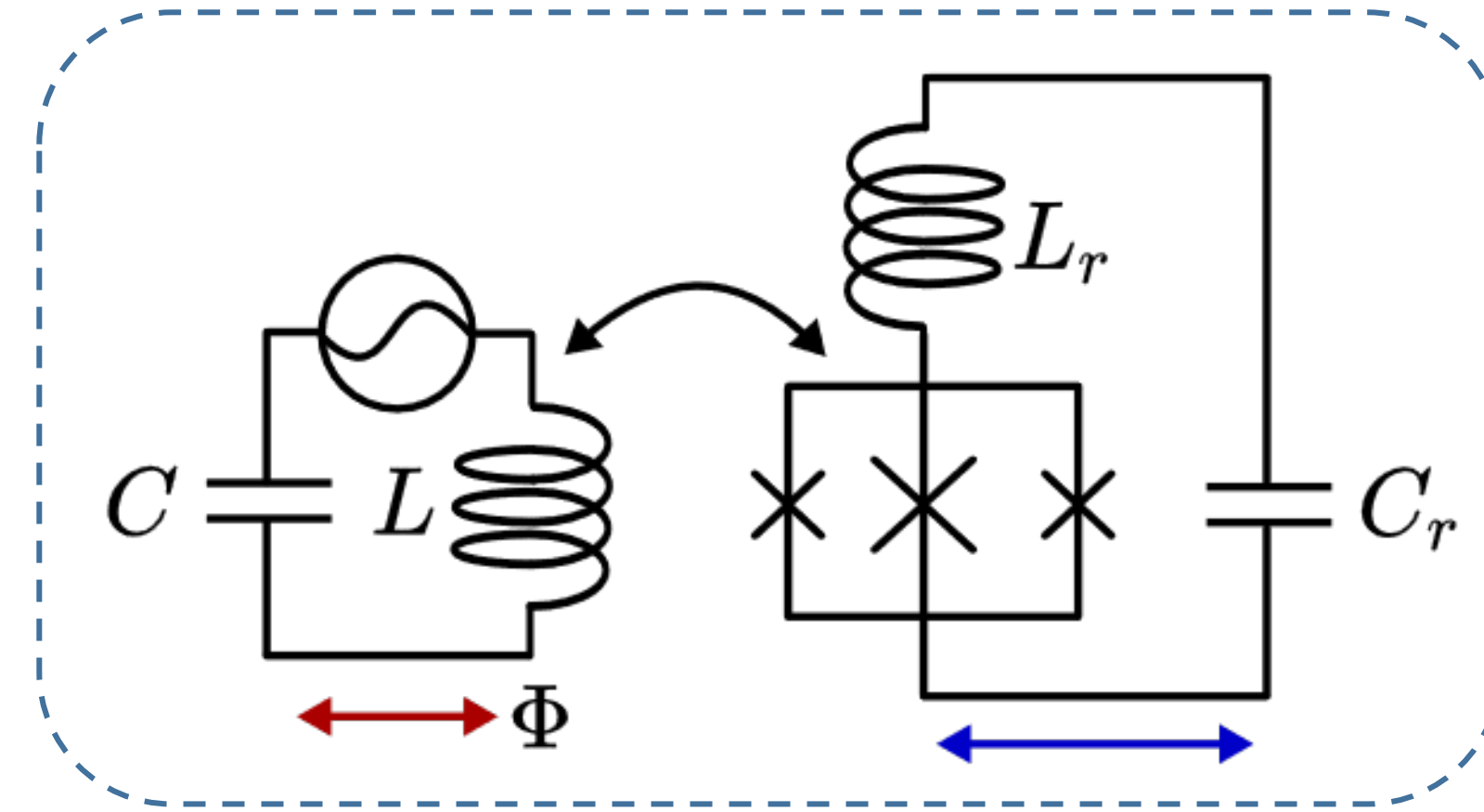
- LIGO performs quantum measurements at better than the SQL to measure gravitational waves more sensitively
- LIGO uses squeezing to improve interferometry, but it could also utilize backaction evasion (BAE).

# Radio-Frequency Quantum Upconverters: Analogous to Optomechanical Systems

**LIGO:**



**Axion detector with RQU:**



$$\omega_a = \sqrt{\frac{k}{m}} \quad \omega_b = \frac{2\pi qc}{l(x)}$$

$$\omega_a = \sqrt{\frac{1}{LC}} \quad \omega_b = \sqrt{\frac{1}{L_r(\Phi)C_r}}$$

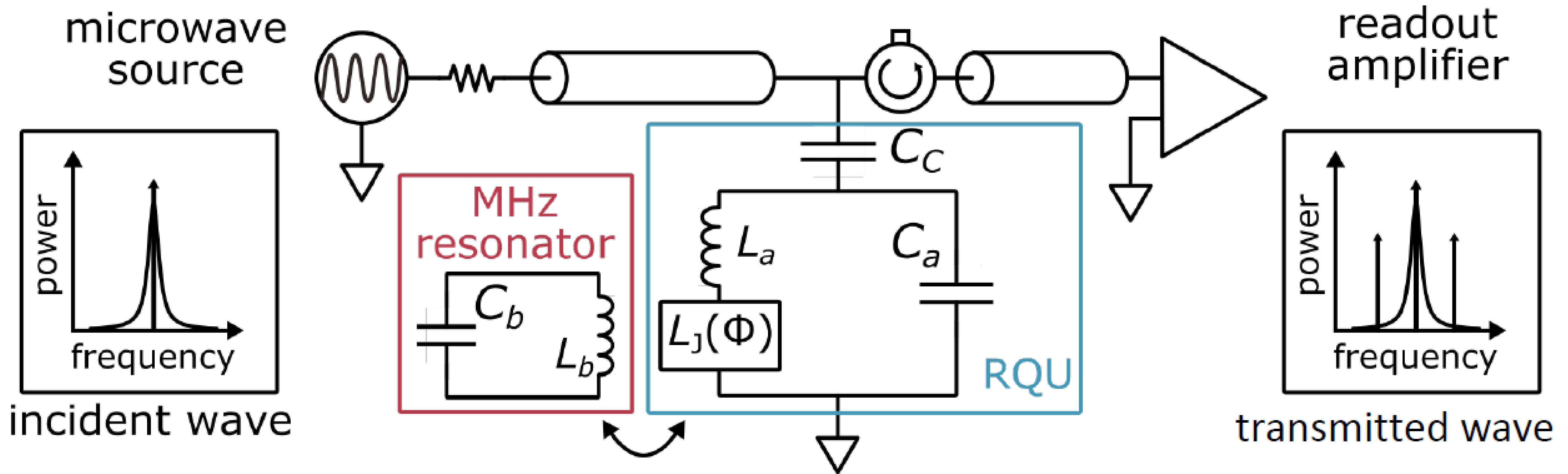
Optomechanical Hamiltonian

$$\hat{H} = \hbar\omega_a(\hat{a}^\dagger\hat{a} + 1/2) + \hbar\omega_b(\hat{b}^\dagger\hat{b} + 1/2) + \hat{H}_{\text{INT}}$$

$$\hat{H}_{\text{INT}} = -\hbar F \hat{b}^\dagger \hat{b} (\hat{a}^\dagger + \hat{a}) / \sqrt{2}$$



# RF-to-Microwave Quantum Upconversion



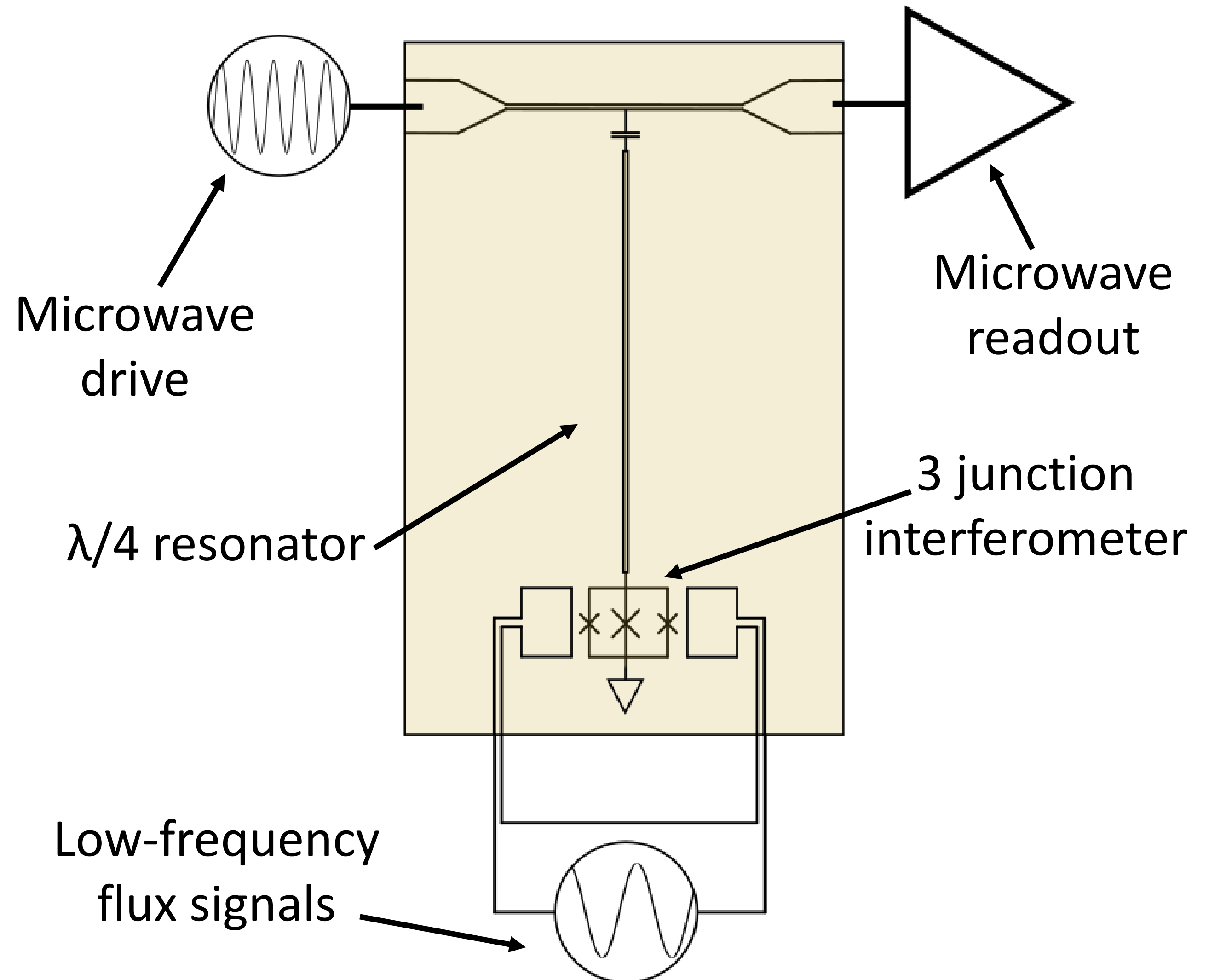
Optomechanical Hamiltonian

$$\hat{H} = \hbar\omega_a(\hat{a}^\dagger\hat{a} + 1/2) + \hbar\omega_b(\hat{b}^\dagger\hat{b} + 1/2) + \hat{H}_{\text{INT}}$$

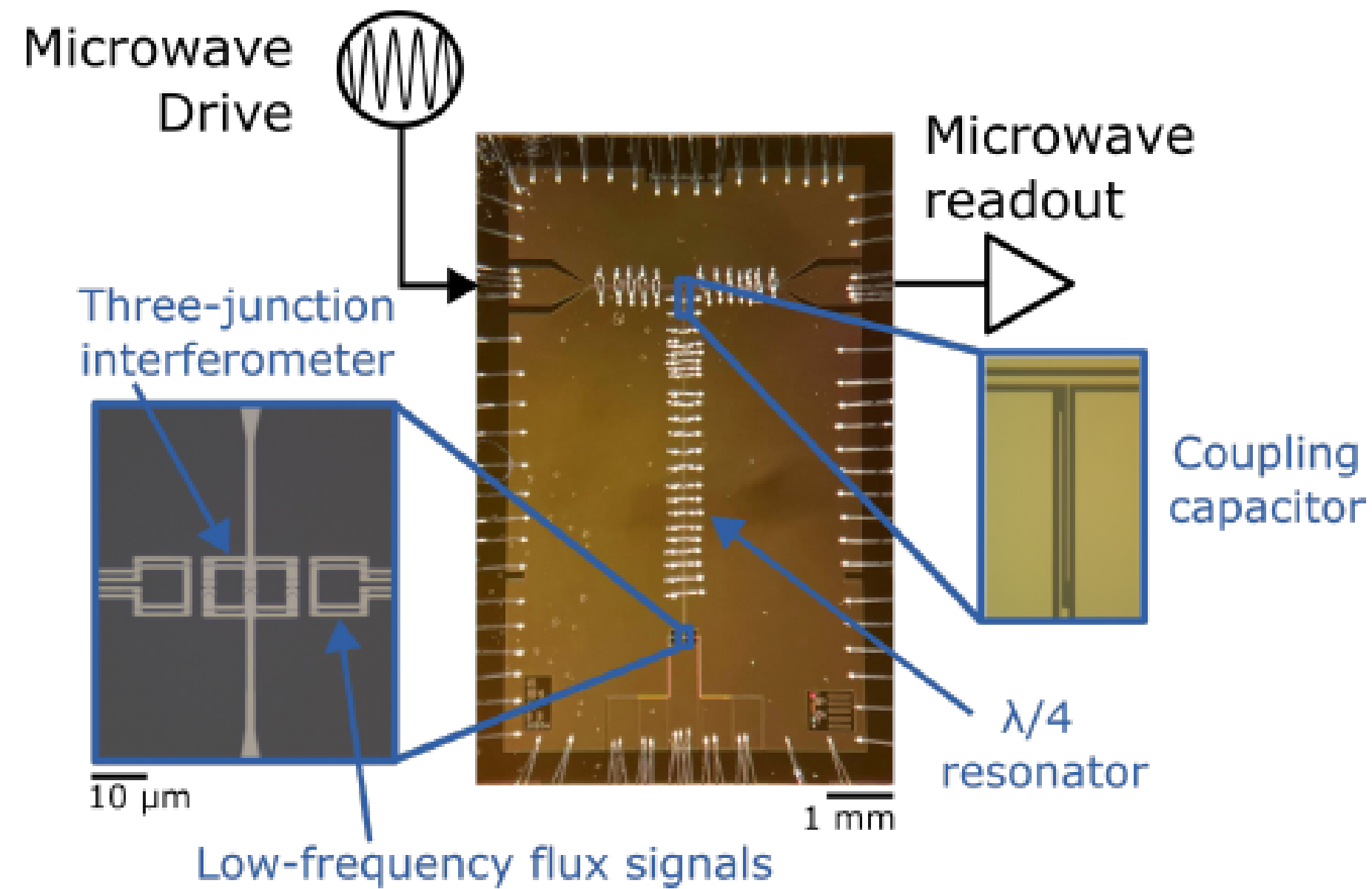
$$\hat{H}_{\text{INT}} = -\hbar\hat{F}\hat{b}^\dagger\hat{b}(\hat{a}^\dagger + \hat{a})/\sqrt{2}$$

# Implementing the RQU

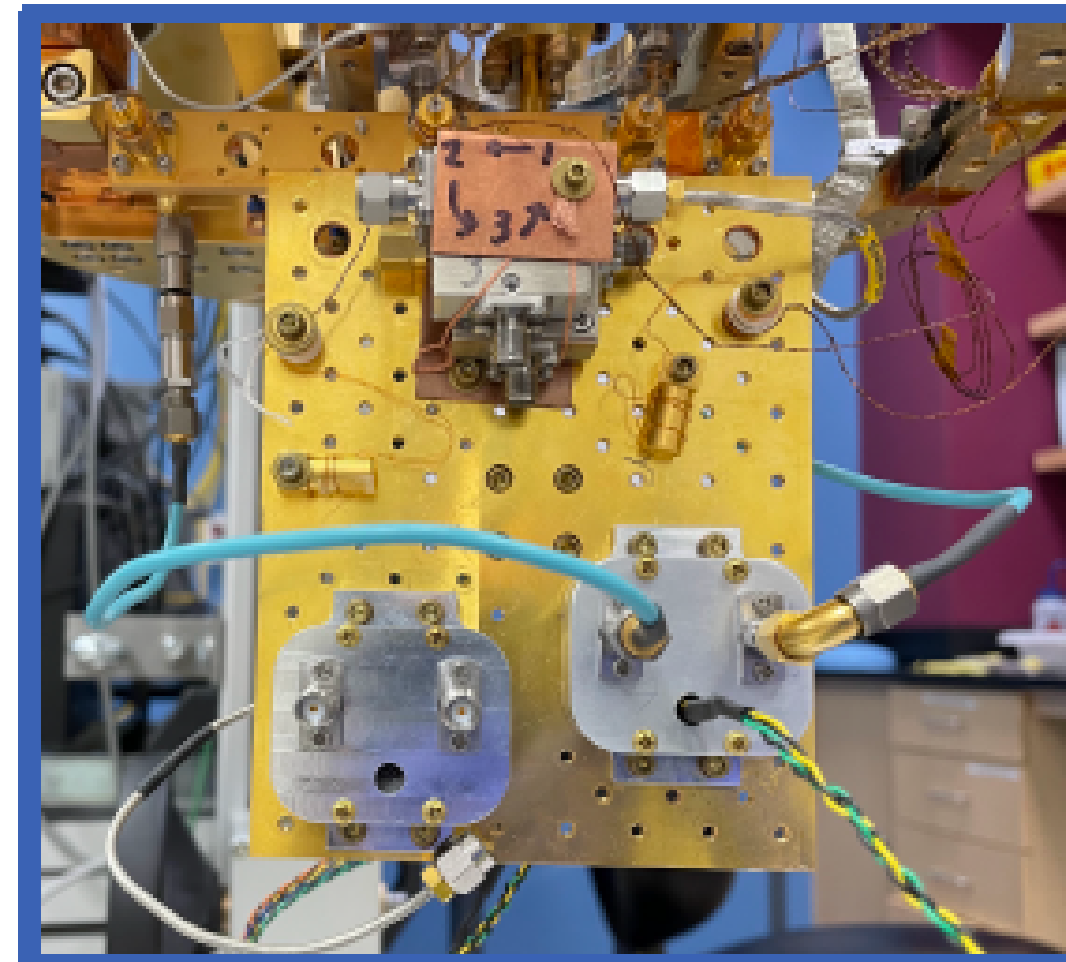
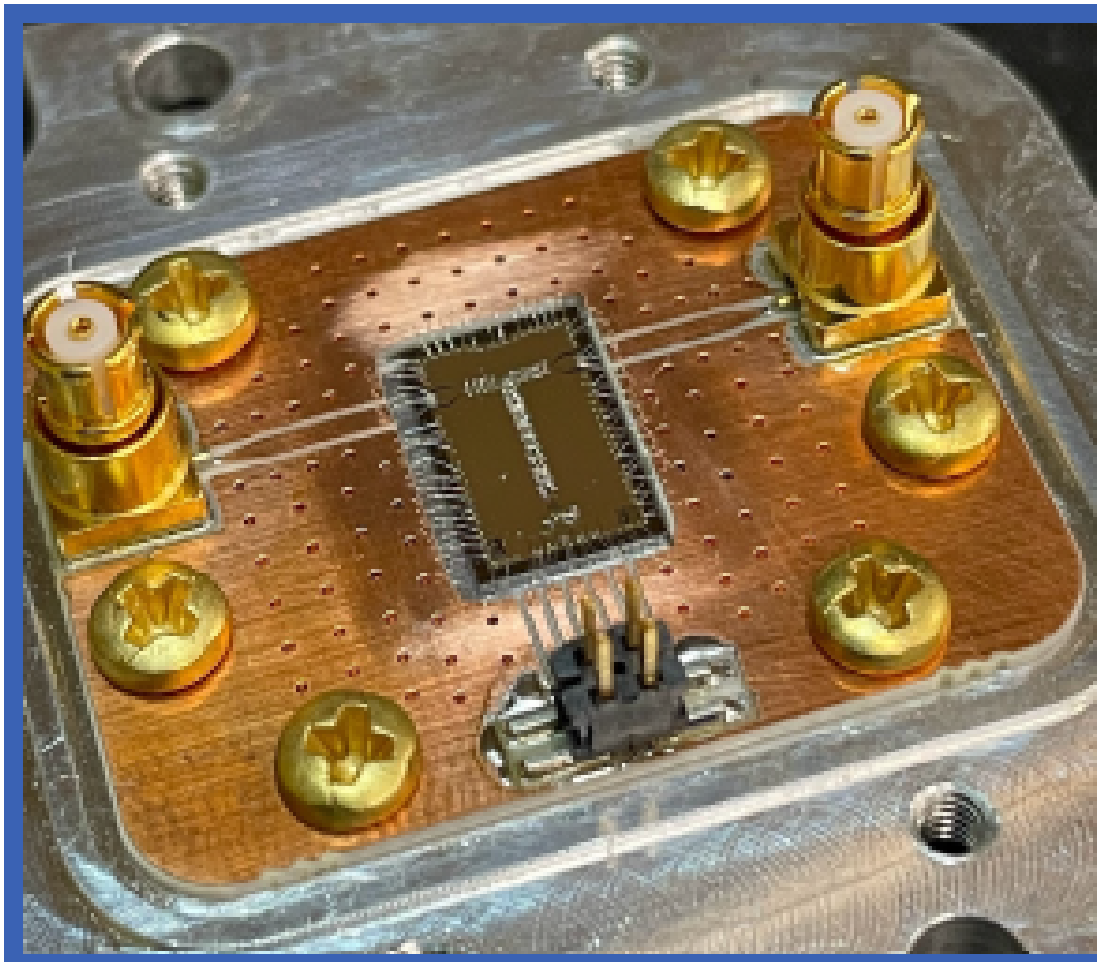
- A tunable inductor is made of an interferometer with three Josephson junctions and two loops.
- The microwave resonator is a quarter wave stub of coplanar waveguide.
- Flux inputs couple a low-frequency input signal into the interferometer.
- Low-frequency ( $\sim$ MHz) signal is converted to microwave signal ( $\sim$ 6 GHz)



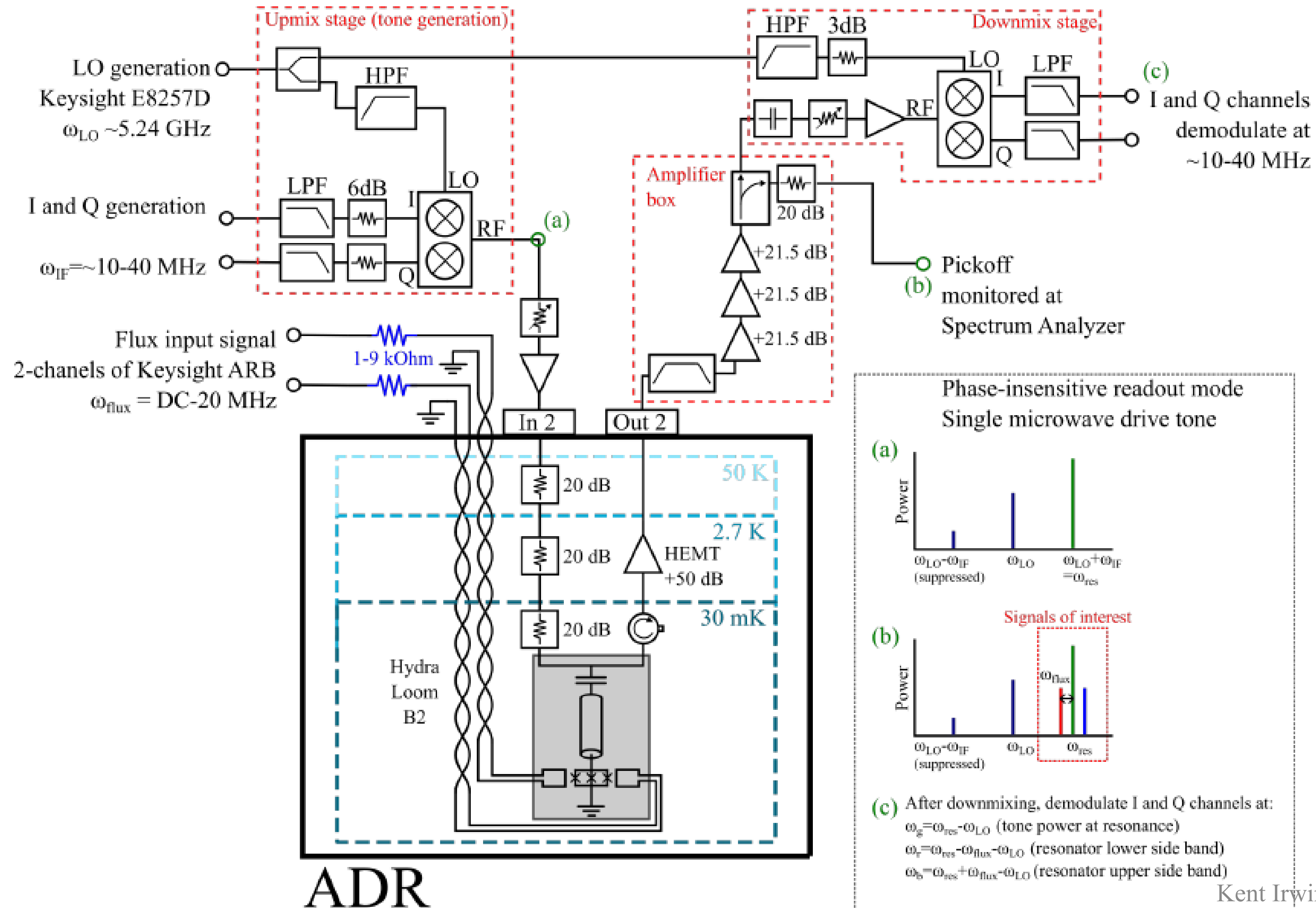
# Implementing the RQU



- Al-AlOx-Al on silicon substrate.
- Single deposition process using shadow evap.
- Symmetric 3-junction interferometer with larger central junction inductance.
- Single-turn flux signal input loops next to interferometer.
- Coplanar waveguide quarter-wave resonator.
- Capacitor couples the resonator to the transmission line.

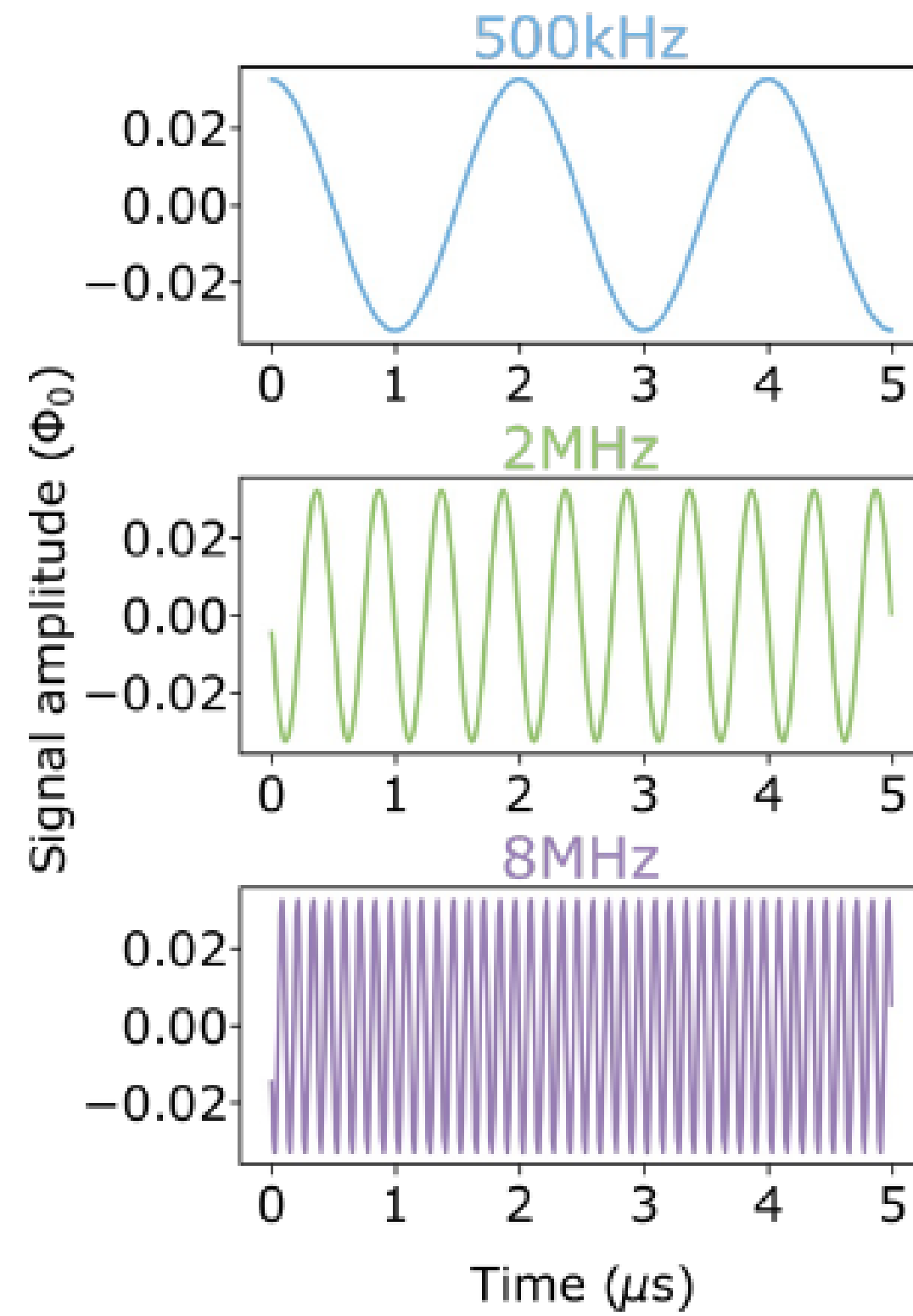


# RQU Readout Chain

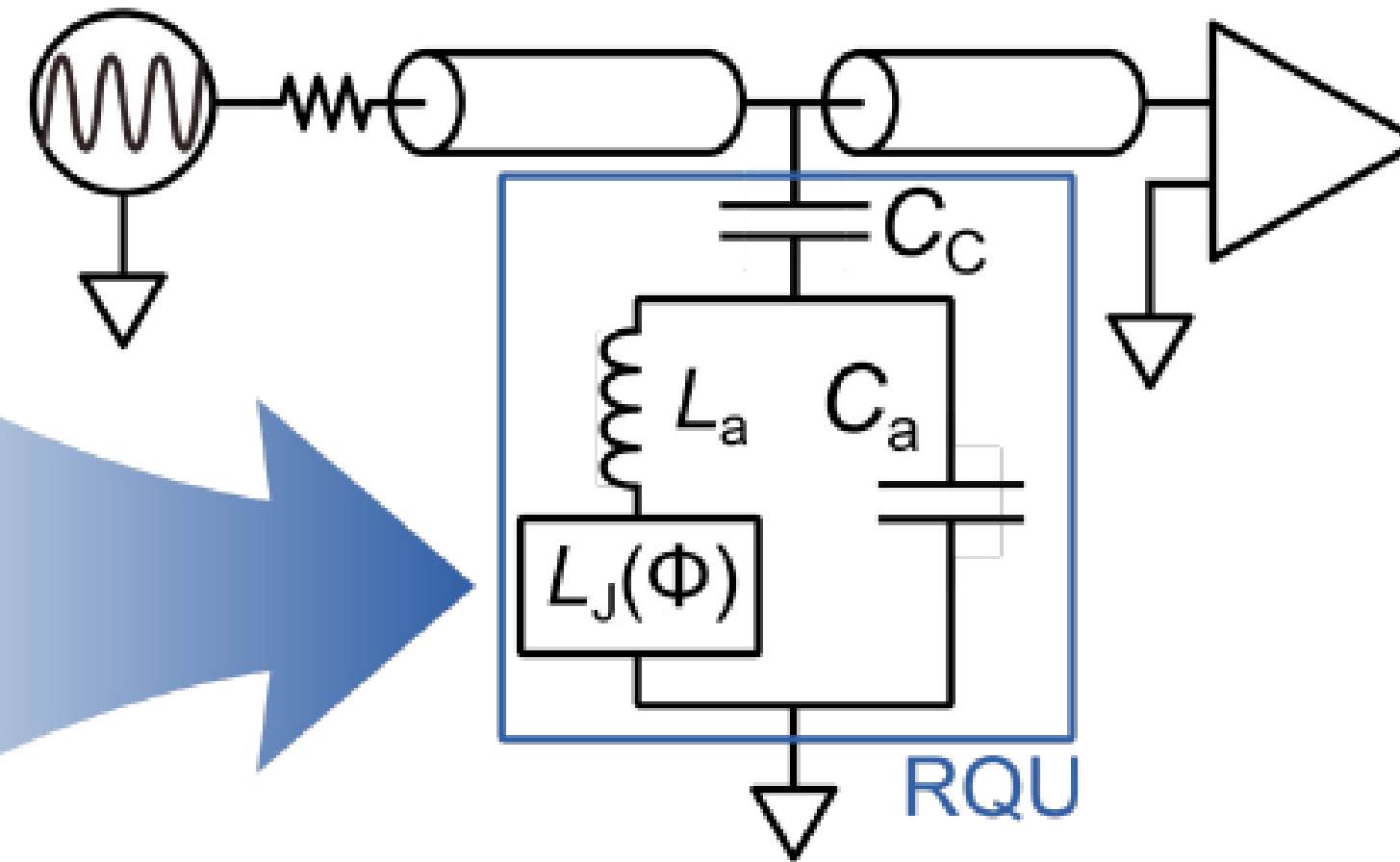


# Data Illustrating RF Upconversion

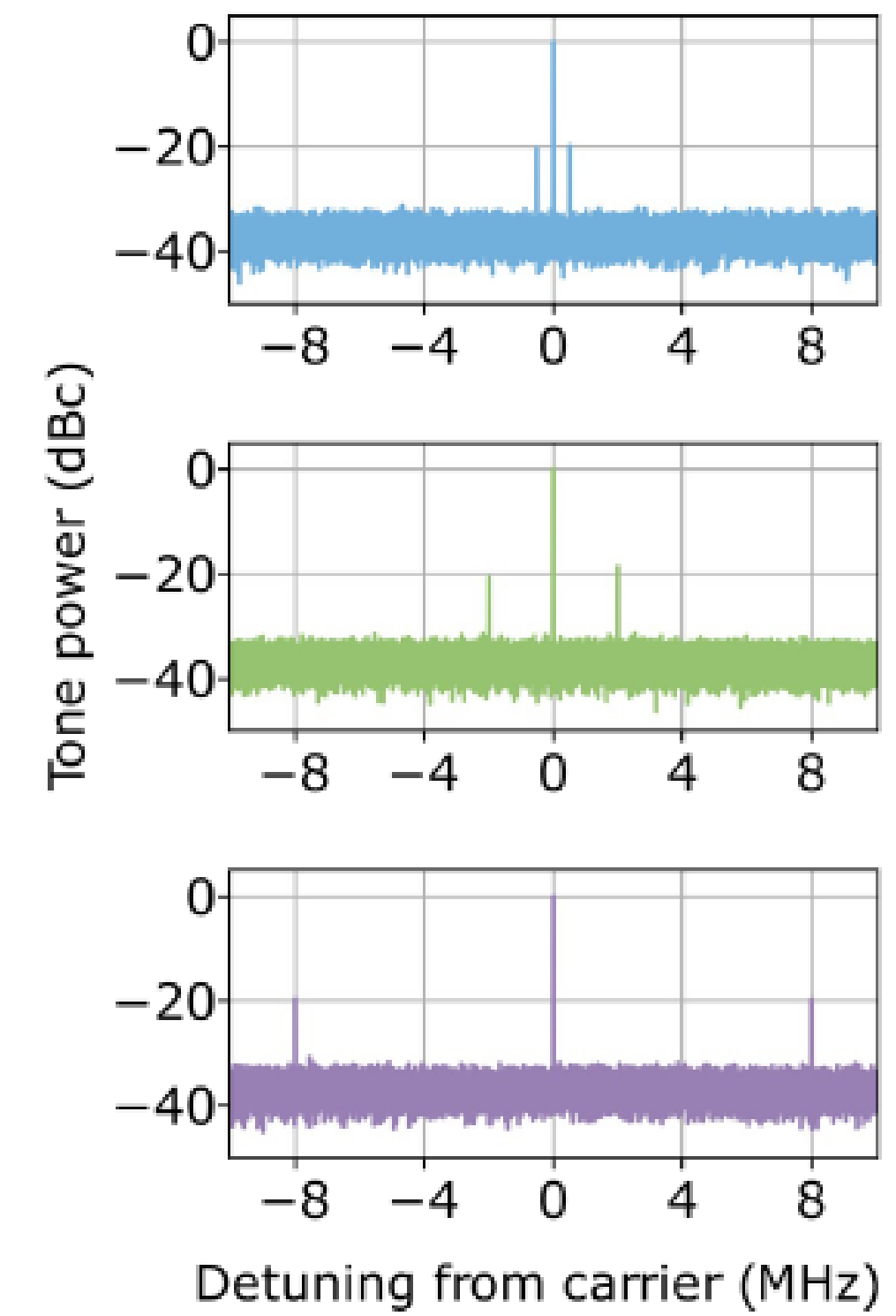
Inject low-f signal



Drive with microwave carrier



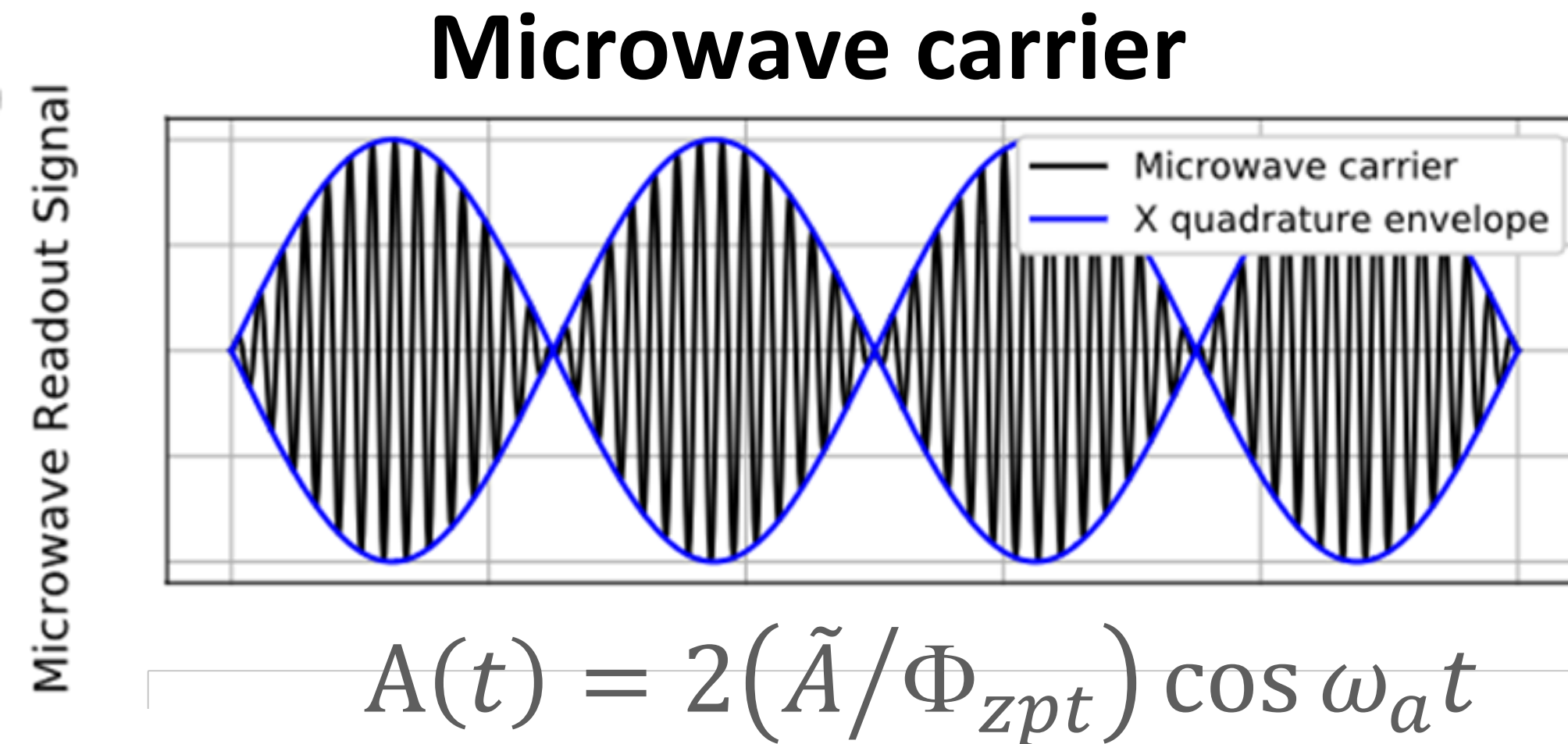
Sidebands carry low-f signal



- Data illustrating upconversion in RQU
- The signal information is upconverted to symmetric sidebands on the microwave carrier tone.

# Phase-Sensitive Upconversion

If the carrier tone is amplitude modulated in phase with the X-quadrature of the input signal, phase-sensitive amplification of only the X-quadrature is achieved.



Clerk, *New Journ. Phys.* **10**, 095010 (2008).

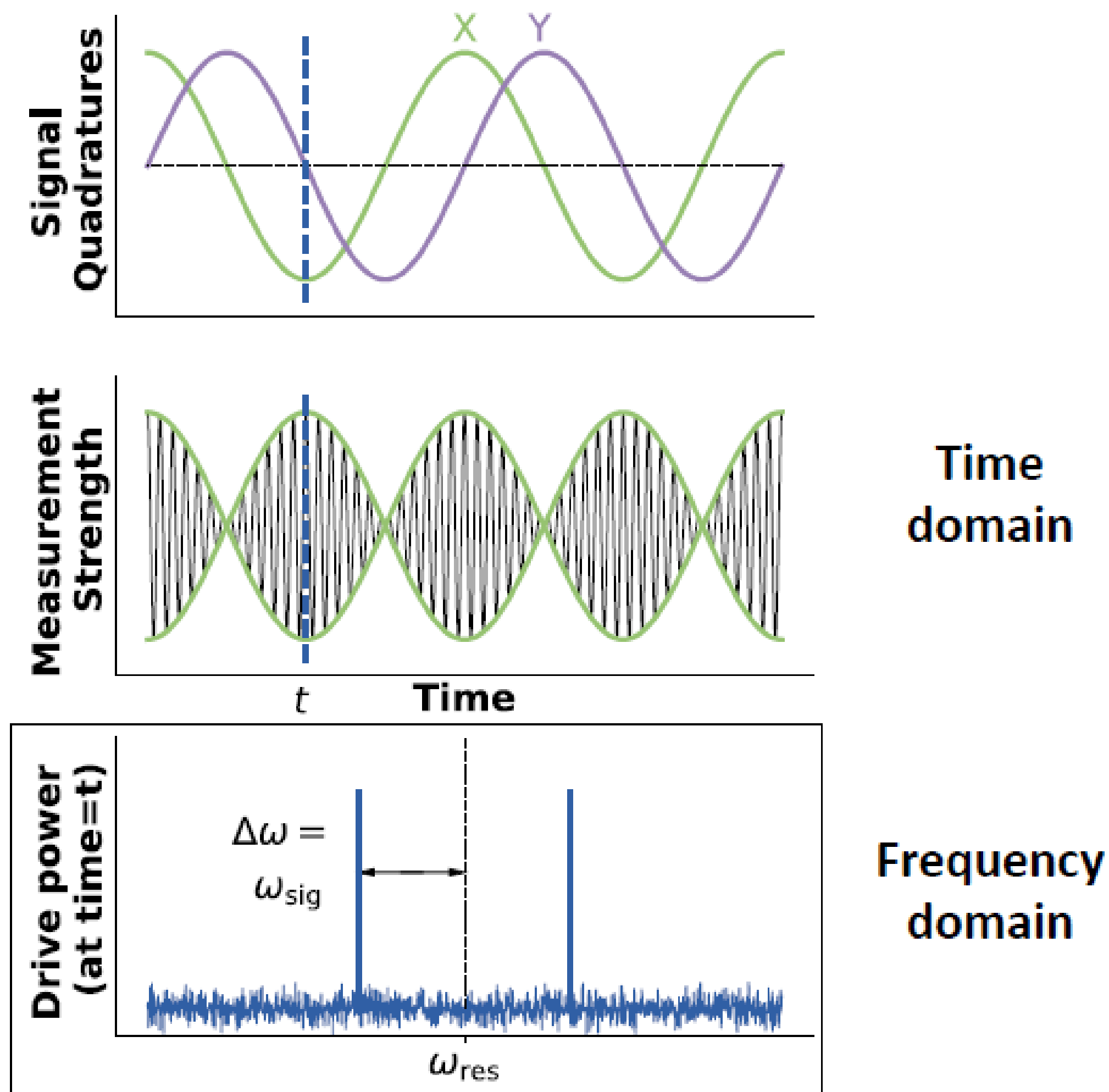
$$\hat{H} = \hbar\omega_a(\hat{a}^\dagger\hat{a} + 1/2) + \hbar\omega_b(\hat{b}^\dagger\hat{b} + 1/2) + \hat{H}_{\text{INT}}$$

$$\hat{H}_{\text{INT}} = -\hbar A \hat{F} \hat{\Phi} = -\sqrt{2}\hbar\tilde{A}\hat{F}[\hat{X}(1 + \cos(2\omega_a t)) + \hat{Y} \sin(2\omega_a t)]$$

If the carrier tone is amplitude modulated in phase with the X-quadrature of the input signal, phase-sensitive upconversion of only the X-quadrature is achieved (averaged over multiple cycles).

# How to achieve amplitude modulation

- Measurement strength depends on microwave drive amplitude.
- A smooth amplitude modulation envelope, averaged over time, will contain no information about the Y quadrature.
- Create the amplitude modulated drive by injecting two microwave tones detuned from the resonant frequency by  $\pm\omega_{\text{sig}}$



Simulated data for illustration

23

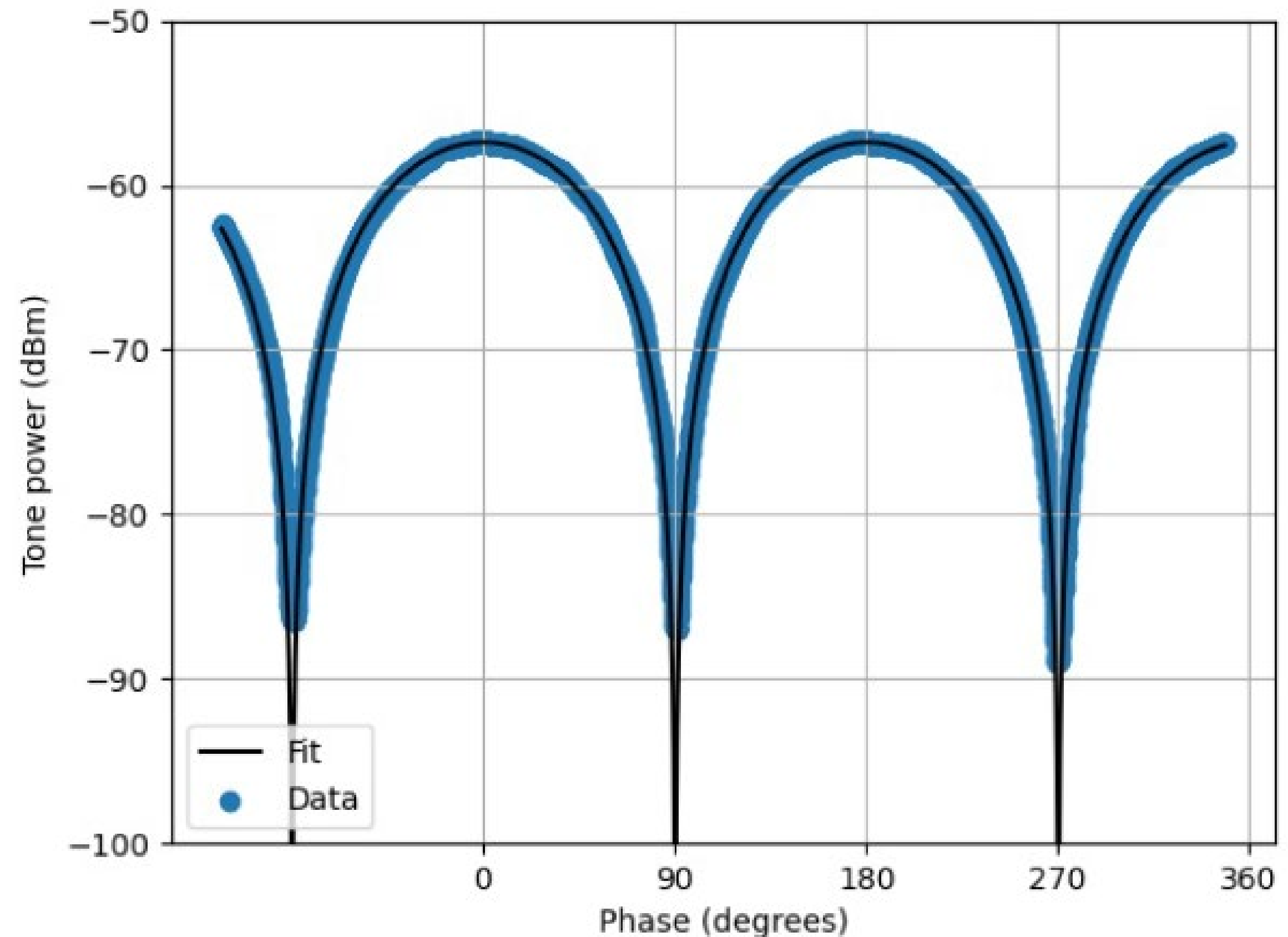
# Demonstrated RQU phase-sensitive gain

**Input:** 5 MHz flux signal into an RQU

**Carrier:** 5.26 GHz sinewave amplitude modulated at 5.26 GHz

**Measure:** output tone power as a function of phase shift between input sinewave and AM modulation

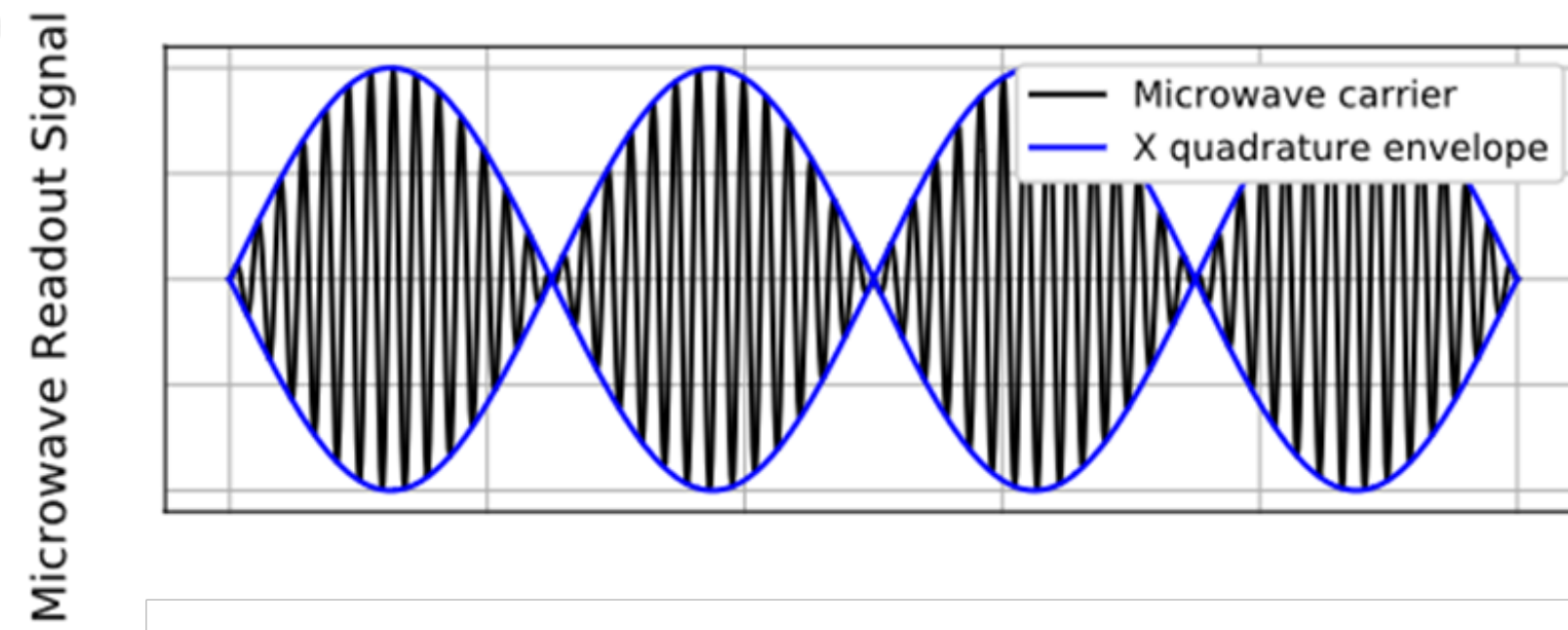
**Extinction ratio:** ratio of maximum gain to minimum gain:  
Curve fit to  $> 50$  dB  
Always measured to  $> 30$  dB





# Full Backaction Evasion

**Carrier tone modulated to measure only X quadrature**



- A backaction signal from the microwave resonator only does work on an LC resonator quadrature, on average, if it is 90 degrees out of phase.
- In this limit, if only the  $\hat{X}$  quadrature is measured, the backaction is injected preferentially into the  $\hat{Y}$  quadrature (which is not measured) - BAE
- If the Q of the microwave resonator is high enough (the “good cavity” limit), the sidebands are fully resolved

Microwave resonator linewidth

$$S_X(\omega) = \frac{\gamma}{(\omega - \omega_a)^2 + (\gamma/2)^2} [1/2 + n_{\text{th}} + n_{\text{leak}}] + S_{\text{IMP-X}}(\omega)$$

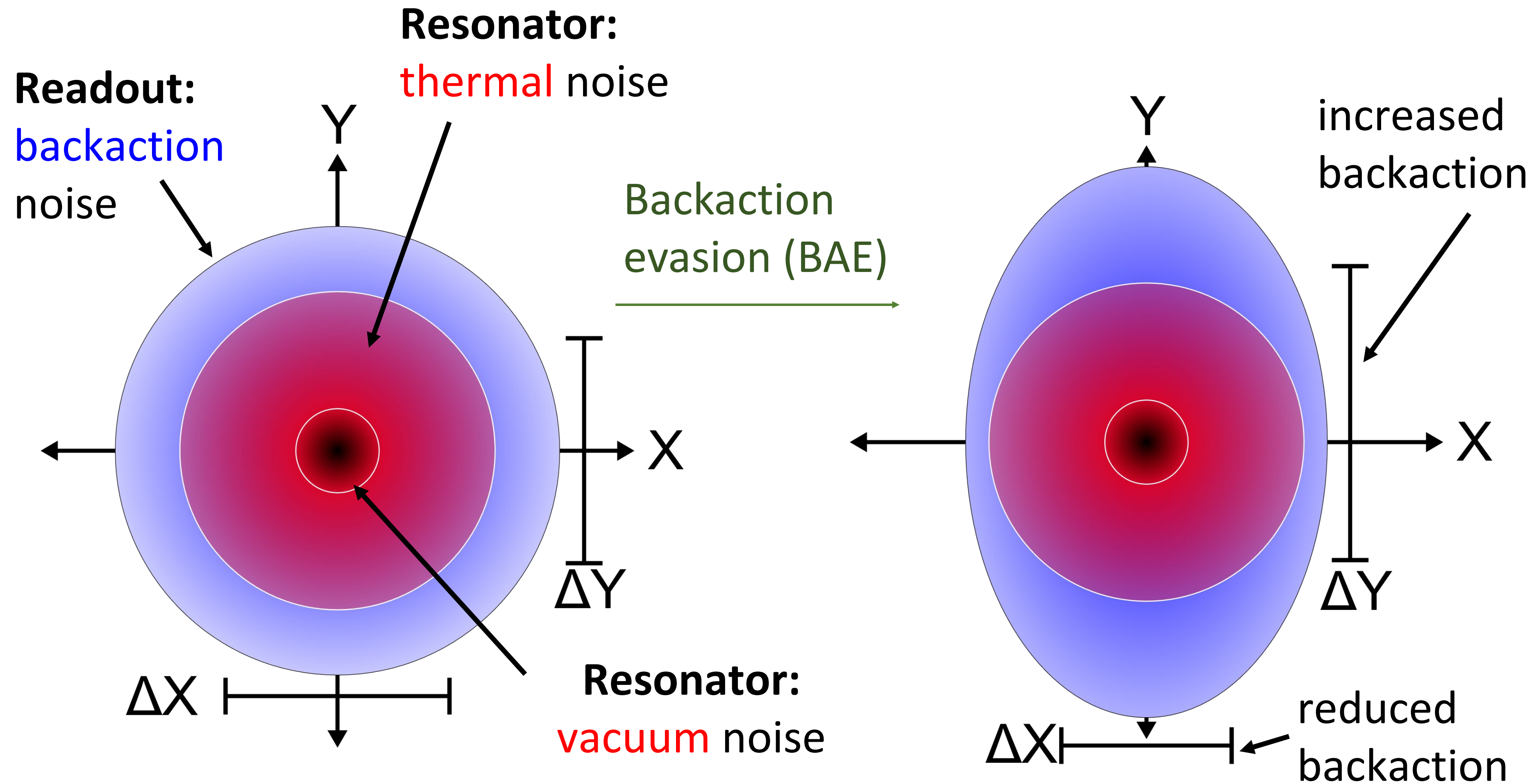
$$n_{\text{leak}} = \frac{n_{\text{BA}}}{32} \left( \frac{\kappa}{\omega_a} \right)^2$$

$$S_Y(\omega) = \frac{\gamma}{(\omega - \omega_a)^2 + (\gamma/2)^2} [1/2 + n_{\text{th}} + n_{\text{BA}} + n_{\text{leak}}] + S_{\text{IMP-Y}}(\omega)$$

Braginsky, Vorontsov, and Thorne. *Science* **209**, 547 (1980).

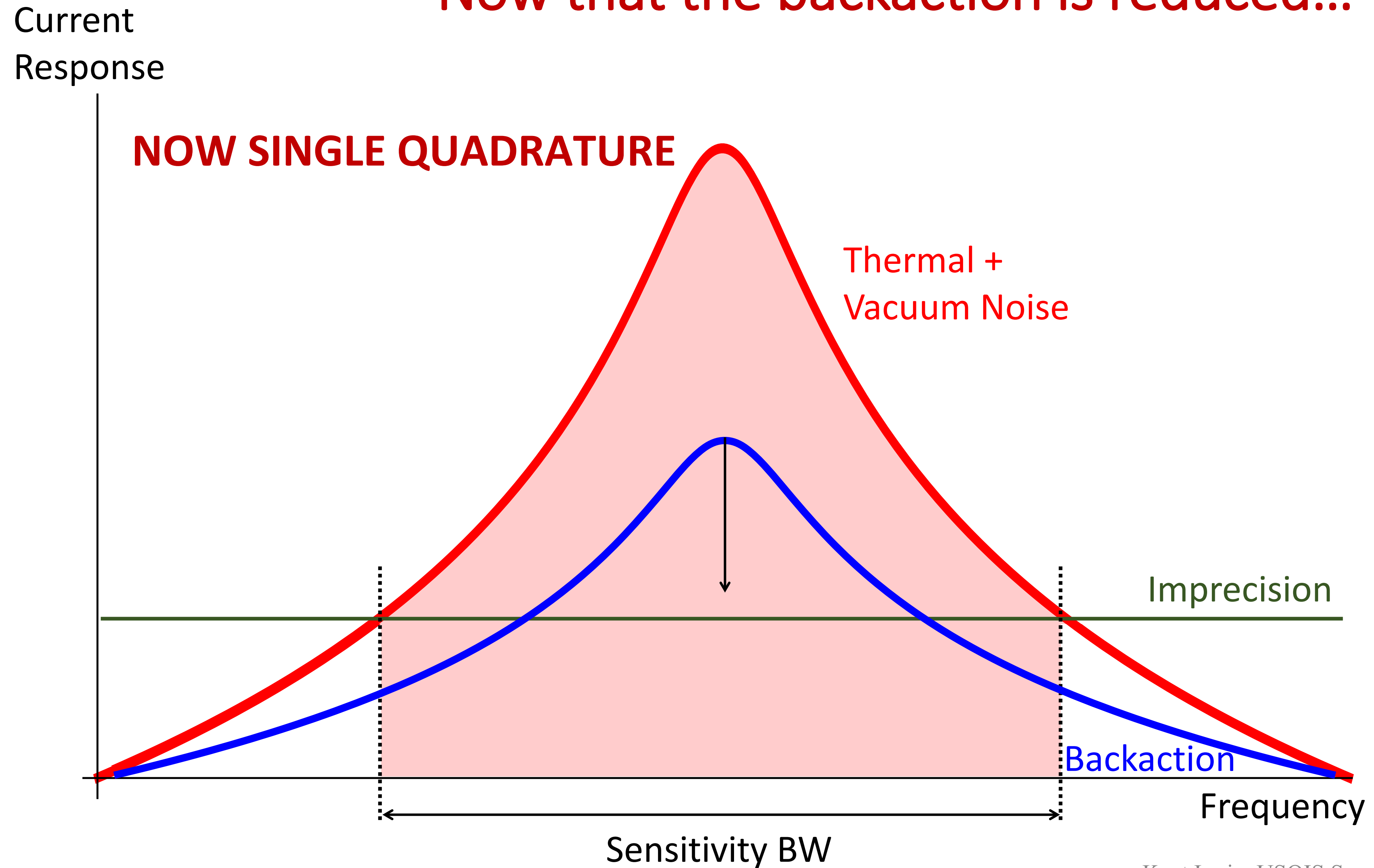
AA Clerk, F. Marquardt, and K. Jacobs, *New Journ. Phys.* **10**, 095010 (2008).

# Backaction evasion (BAE): reduced readout noise in one quadrature

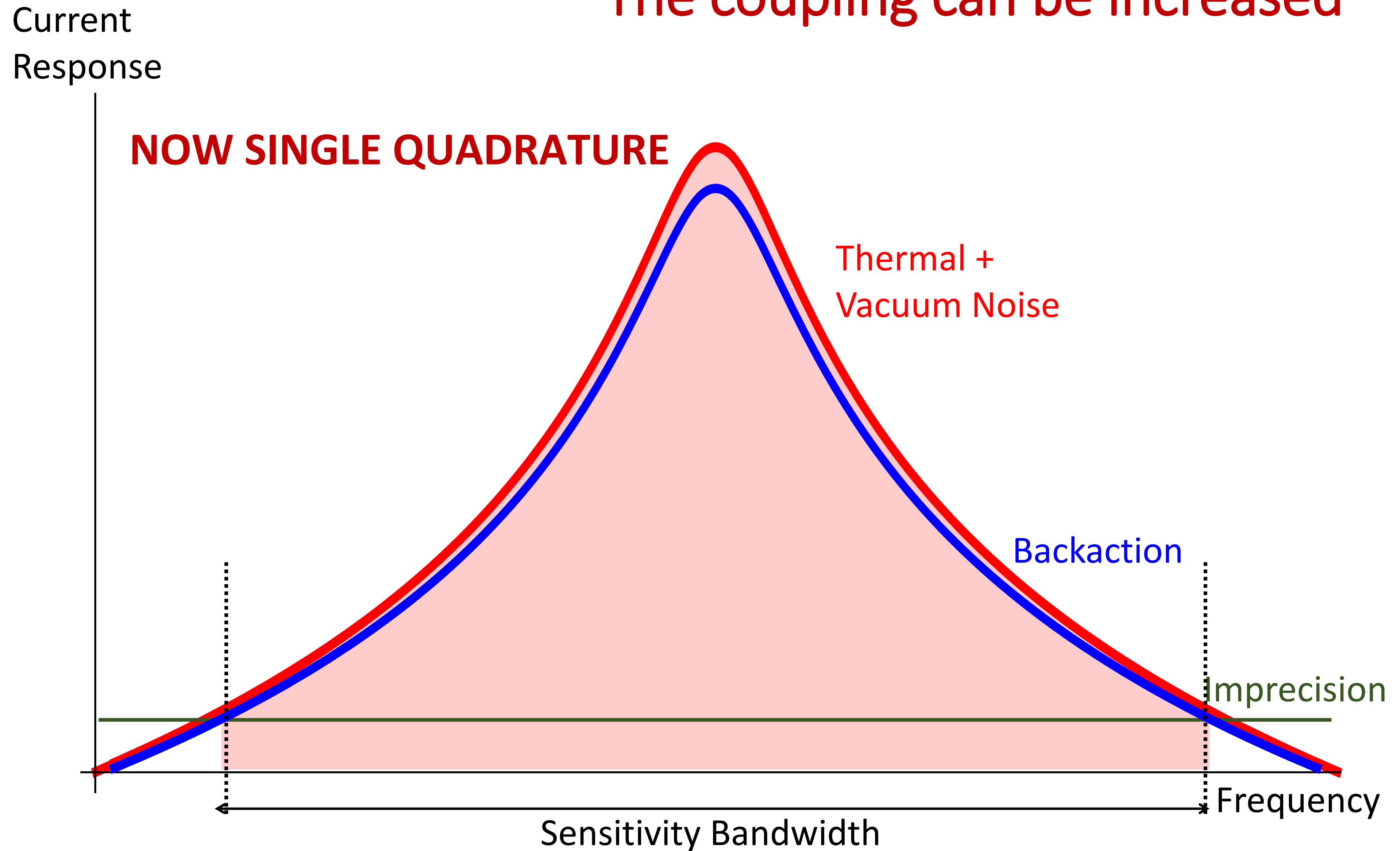


The reduction in the total X-quadrature noise appears unimpressive on resonance

# Now that the backaction is reduced...

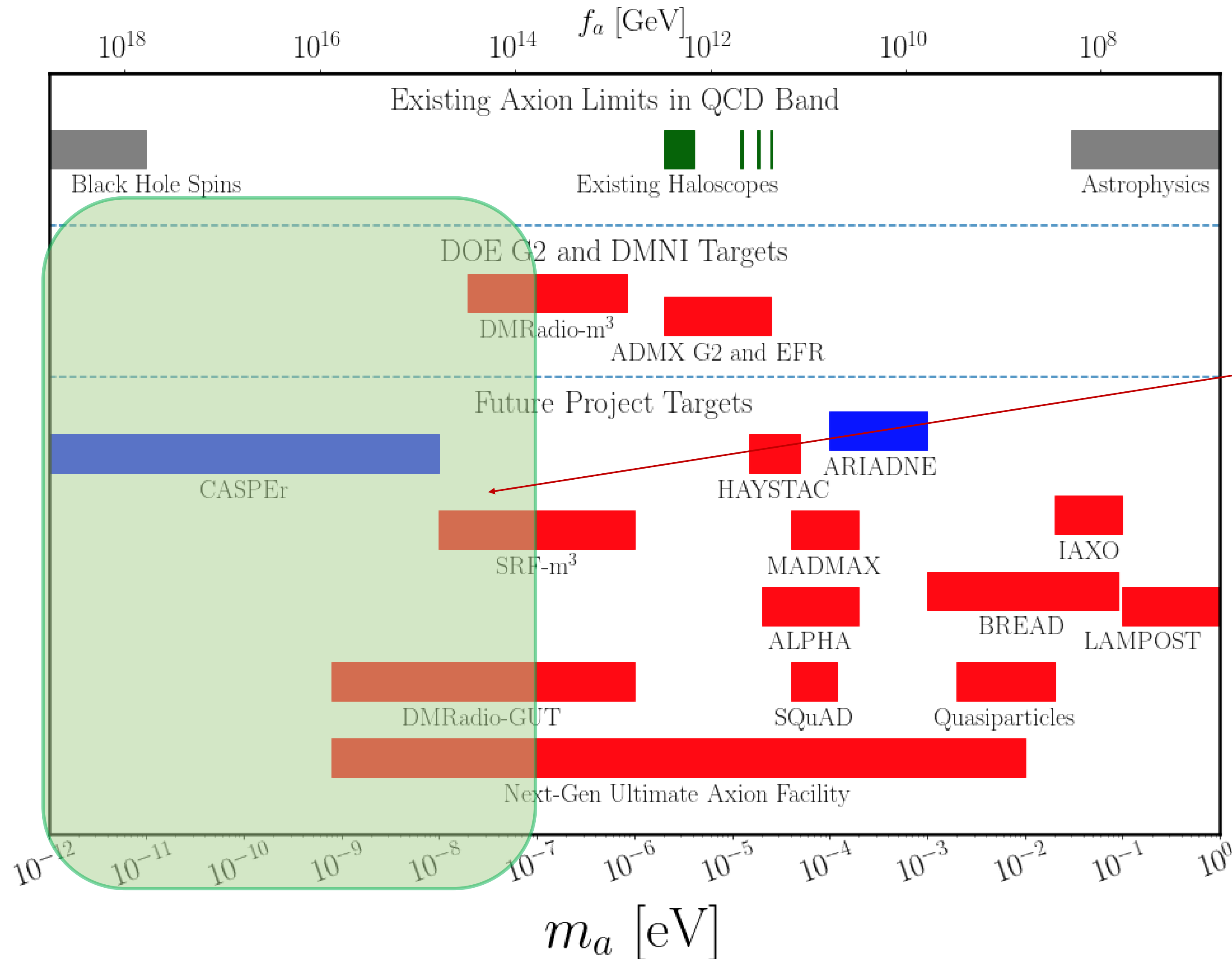


# The coupling can be increased



Sensitivity BW can be greatly increased

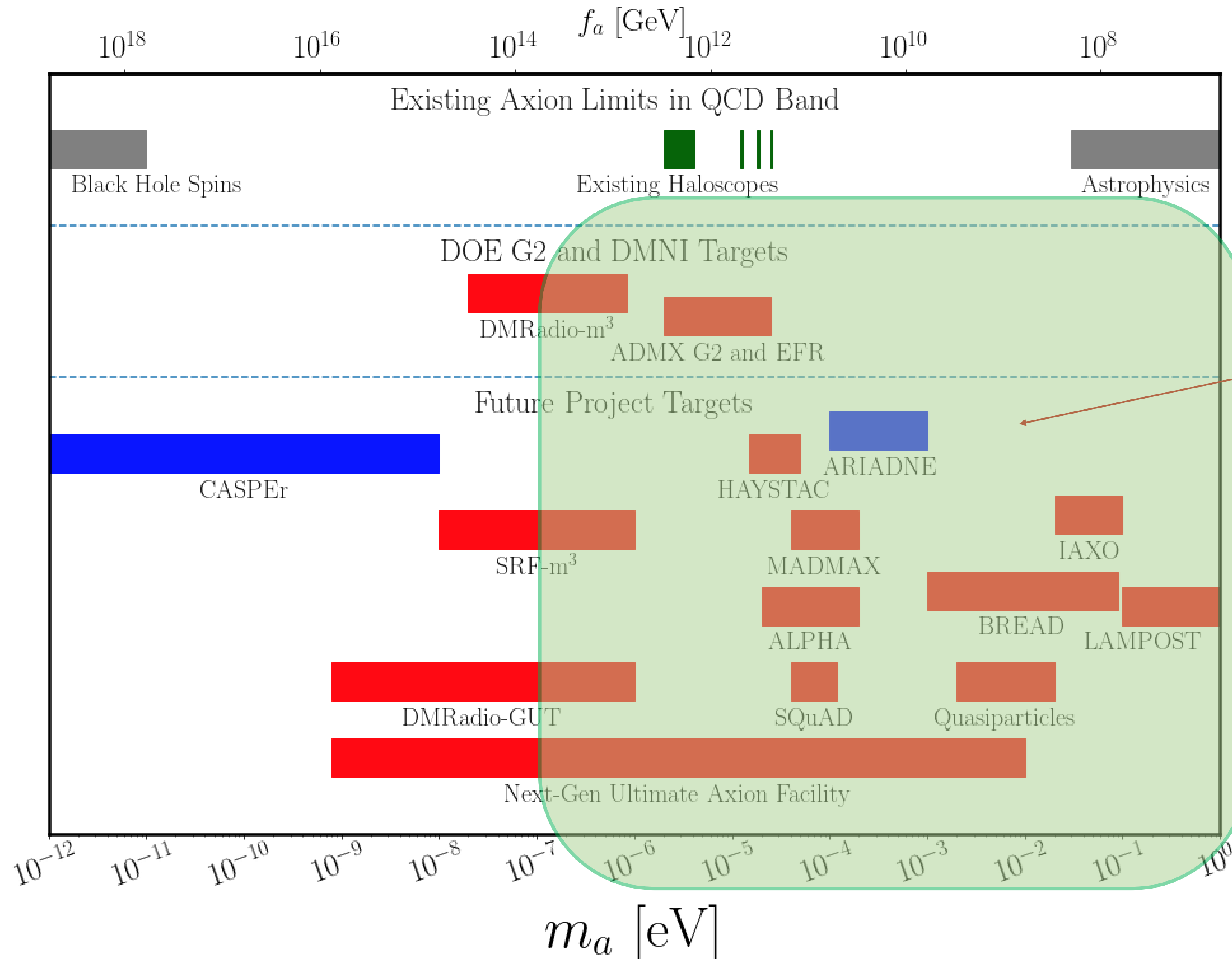
# SNOWMASS: a comprehensive search for QCD axions



QCD axions here:  
provably requires  
measurement  
below the SQL to  
probe the best  
models

Figure 3 of Snowmass CF2 summary  
[https://snowmass21.org/\\_media/cosmic/repv1\\_cf2.pdf](https://snowmass21.org/_media/cosmic/repv1_cf2.pdf)

# SNOWMASS: a comprehensive search for QCD axions

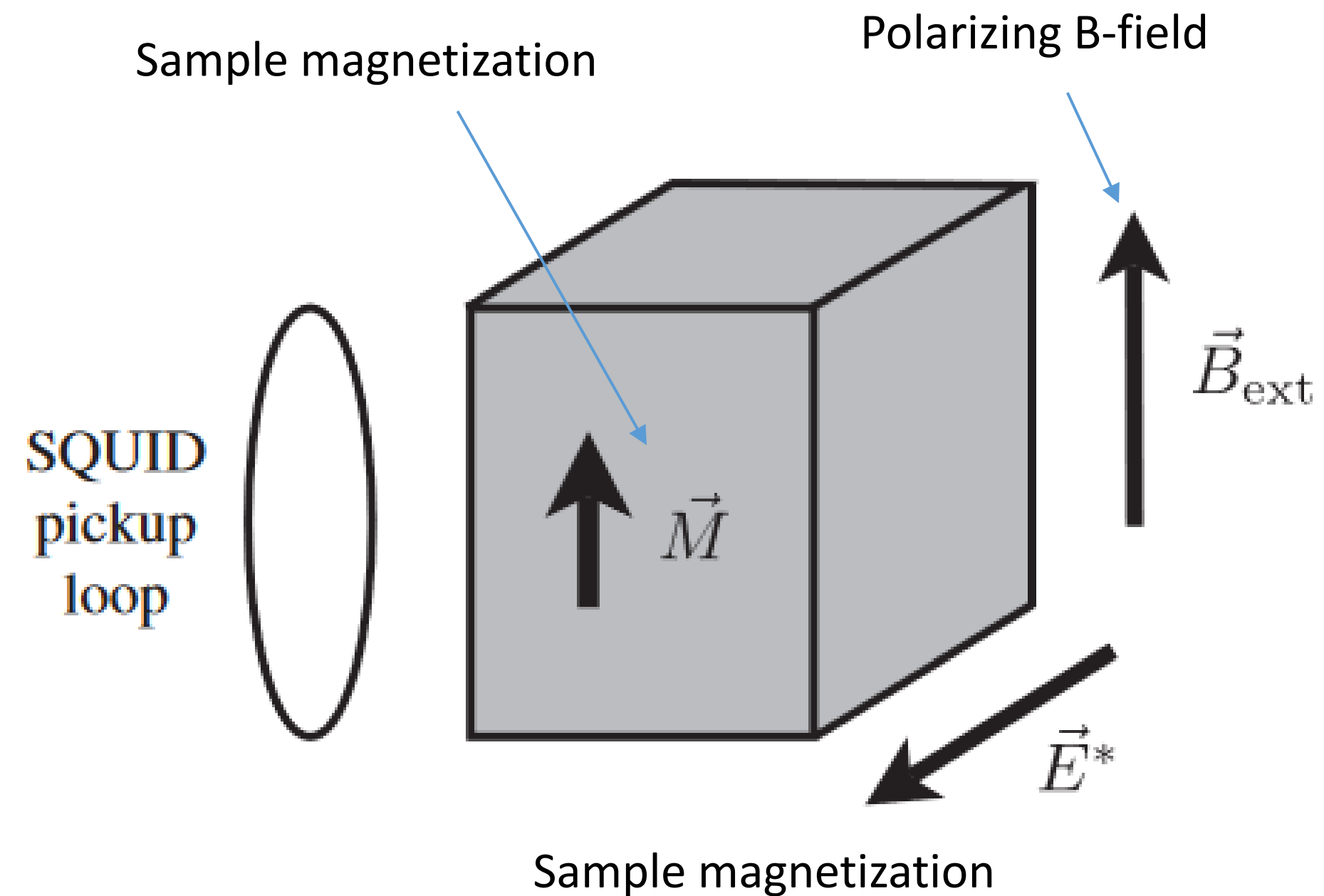


QCD axions here:  
New quantum  
sensing  
modalities  
required to cover  
all frequency  
space

Figure 3 of Snowmass CF2 summary  
[https://snowmass21.org/\\_media/cosmic/repv1\\_cf2.pdf](https://snowmass21.org/_media/cosmic/repv1_cf2.pdf)

# Searching for axions with NMR

- Axions couple to nuclear spins through the strong force, inducing an effective nuclear electric dipole moment which oscillates at  $f_{ax}$ .
- A spin-polarized sample of nuclear spins will resonantly precess if  $f_{ax}$  matches their Larmor frequency.
- The precessing magnetization can be detected with a SQUID magnetometer or quantum sensor.

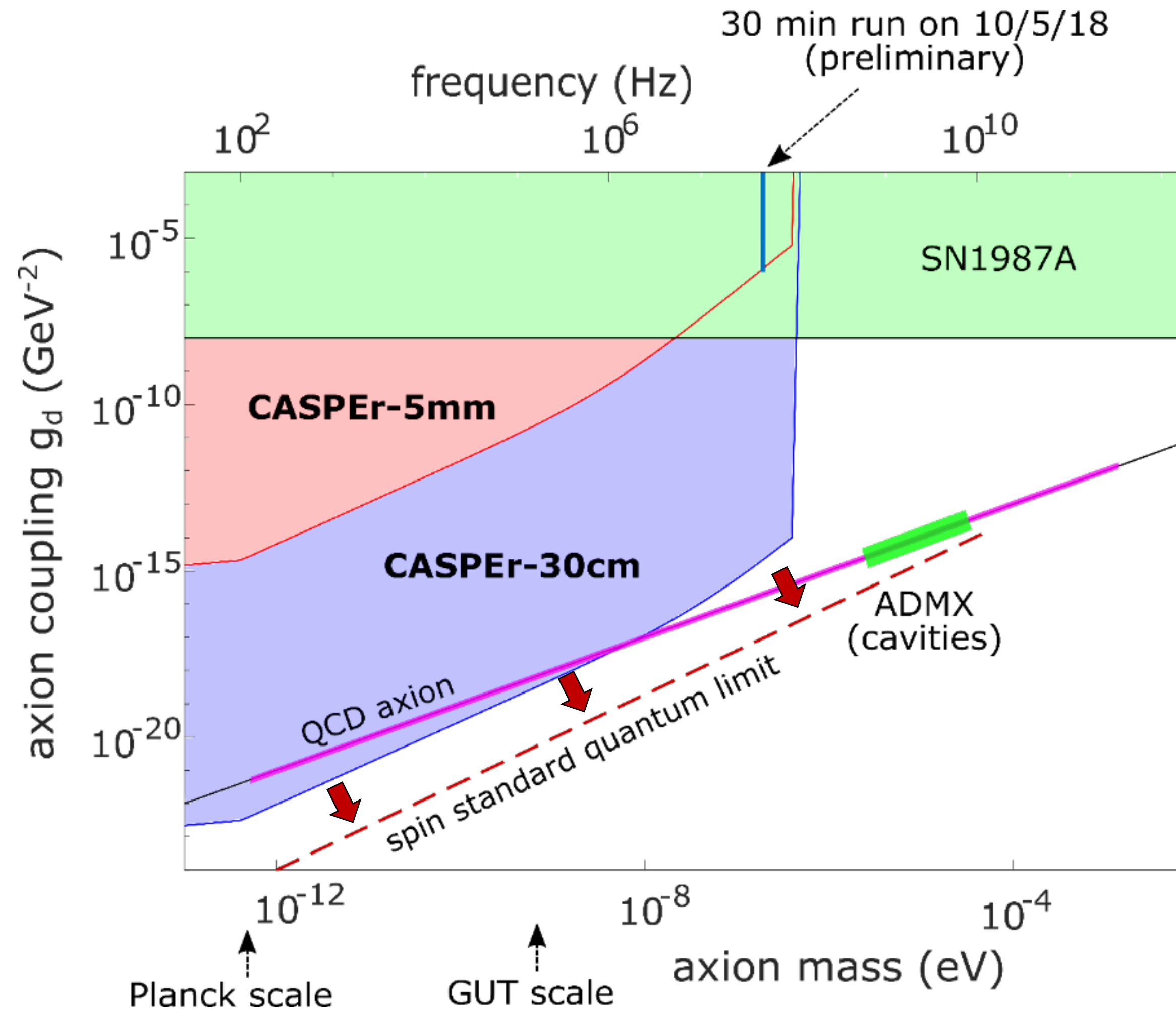


**CASPER-electric**



Garcon, Antoine, et al. "The Cosmic Axion Spin Precession Experiment (CASPER): a dark-matter search with nuclear magnetic resonance." *Quantum Science and Technology* 3.1 (2017): 014008.

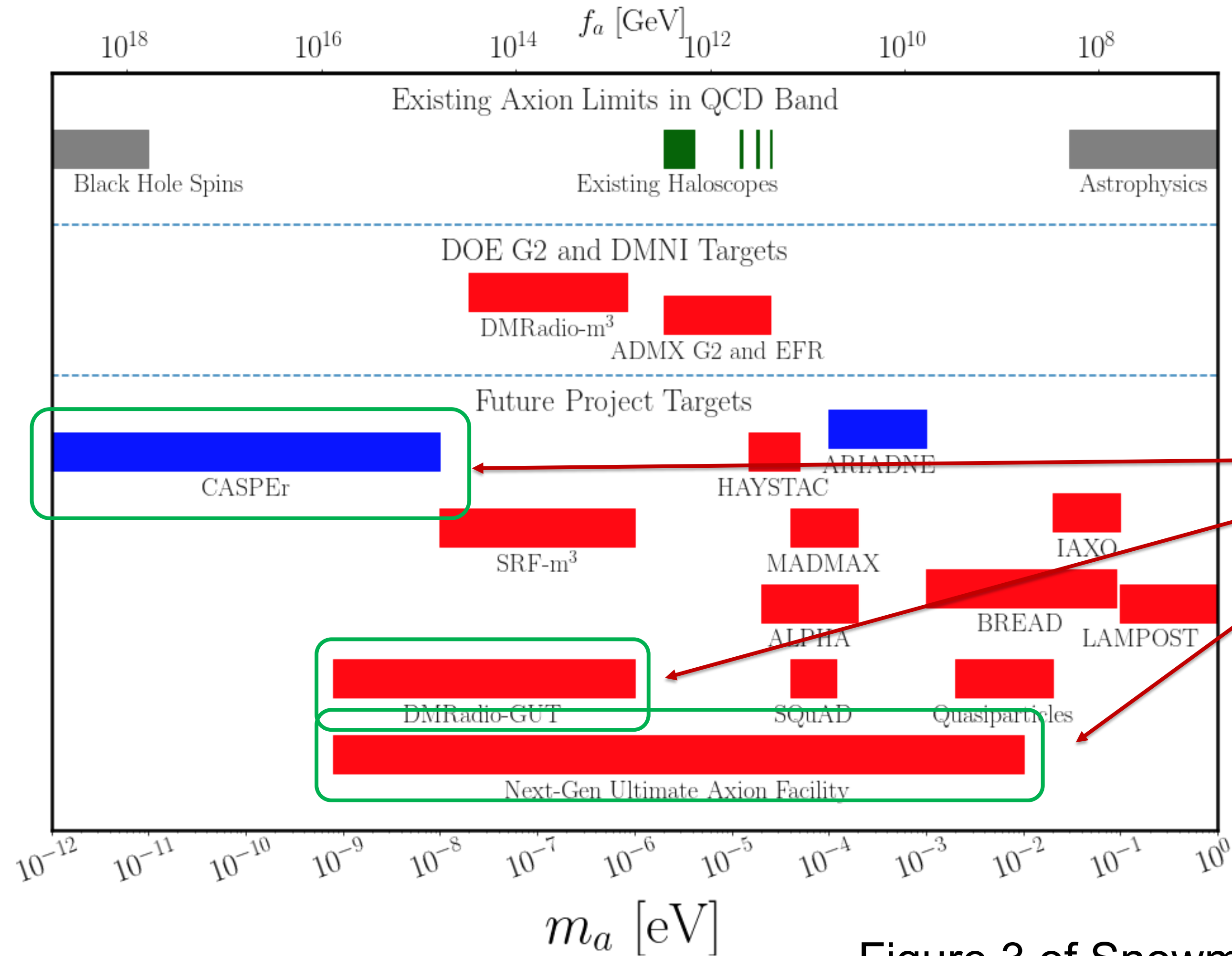
# Accelerating NMR search: CASPER-e



- RQUs can help CASPER's science reach to extend to the spin projection noise SQL
  - Spin squeezing can extend below the spin projection SQL



# Snowmass Graph: Axions and SLAC Quantum



BAE with RQU could enable these experiments

Figure 3 of Snowmass CF2 summary  
[https://snowmass21.org/\\_media/cosmic/repv1\\_cf2.pdf](https://snowmass21.org/_media/cosmic/repv1_cf2.pdf)

# Conclusions

- New breakthroughs in QIS are leading to a revolution in measurement.
- Compelling applications in fundamental physics, including dark matter, require measurement better than Standard Quantum Limits so that we can achieve results in *years* instead of *millennia*.
- Quantum measurement varies at different frequencies, and for different sources -> one size does not fit all, but the same principles underly all of them.
- Next lecture: Aaron Chou will talk about the quantum toolbox at microwave frequencies with qubits