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# Electro-Thermal Feedback Physics in Transition Edge Sensor

#### David Sadek

Radiation Impact on Superconducting Qubits @ Fermilab G4CMP Satellite Workshop

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## Outline

The Physics Problem

#### ODE derivation

Current equation Heat equation TES resistance Current effects

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# **TES thermal connection**

TES is thermally connected to the absorber by a thermal conductance G<sub>ta</sub>.



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# TES thermal connection

- TES is thermally connected to the absorber by a thermal conductance G<sub>ta</sub>.
- Absorber is thermally connected to the heat bath (fridge) by a thermal conductance G<sub>ab</sub>.



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# **TES thermal connection**

- TES is thermally connected to the absorber by a thermal conductance G<sub>ta</sub>.
- Absorber is thermally connected to the heat bath (fridge) by a thermal conductance  $G_{ab}$ .
- TESSim assumes that the absorber and the heat bath are identical. G<sub>ab</sub> is infinite.





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# TES electro-thermal connection

- ► TES is thermally connected to the absorber by a thermal conductance *G*<sub>ta</sub>.
- Absorber is thermally connected to the heat bath (fridge) by a thermal conductance  $G_{ab}$ .
- ► TESSim assumes that the absorber and the heat bath are identical. *G*<sub>ab</sub> is infinite.
- TES in SuperCDMS devices is kept under a constant voltage.
- Change in current changes the magnetic flux of the inductor L
- Magnetic flux changes are sensed and amplified by the SQUID



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## **Current equation**

Kirchoff's loop rule:

$$0 = -V_b + I_b R_b + (I_b - I_L) R_s$$



**TES Channel** 

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#### **Current equation**

Kirchoff's loop rule:



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### **Current equation**

#### Kirchoff's loop rule:



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## **Current equation**

$$\begin{split} \dot{I_L} &= \frac{V_{b0} - I_L \left( R_0 + \frac{1}{\sum_k G_k (A_k)} \right)}{L} \\ I_i &= I_L \frac{G_i}{\sum_k G_k}, \\ G_i &= \frac{1}{R_i} = \frac{2}{R_n} \cosh A_i e^{-A_i}, \\ V_{b0} &= V_b \frac{R_s}{R_b + R_s} \\ R_0 &= R_p + \frac{R_b R_s}{R_b + R_s} \end{split}$$



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#### Heat equation

$$\frac{dT}{dt}C = P_J + P_B + P_P$$



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#### Heat equation

$$\frac{dT_s}{dt}C = P_J + P_B + P_P$$
$$P_J = IR^2(T) = \frac{I}{G^2(T)}$$
$$P_B = K \left(T_b^n - T^n\right) = \Sigma \mathcal{V}_{eff} \left(T_b^n - T^n\right)$$



- $\Sigma$  is the electron-phonon coupling in the TES
- ▶ n = 5 for SuperCDMS detectors
- Phonon power comes from CrystalSim

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## Coupled ODEs

$$\dot{I_L} = \frac{V_{b0} - I_L \left(R_0 + \frac{1}{\sum_k G_k}\right)}{L} \\ \dot{A_i} = \frac{K \left(T_b^5 - T_i^5 \left(A_i\right)\right) + I_L^2 \frac{G_i(A_i)}{\left(\sum_k G_k(A_k)\right)^2} + P_{P,i}}{T_w C}.$$

Not the whole story!

Current effects have to be taken into account

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### **TES resistance model**



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# Current effects

- $\blacktriangleright$   $T_c$  is lowered due to a non-zero current.
- The effective  $T_c$  is modeled by

$$T_c^{\rm eff} = T_c \left(1 - \frac{|I|}{I_c}\right)^n$$

• For SuperCDMS detectors n = 2/3



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# Current effects

Current effects allow for the individual TESs to cross-talk through the matrix  $\frac{\partial T_i}{\partial A_i}$ 

$$\sum_{j} \frac{\partial T_i}{\partial A_j} \dot{A}_j = \frac{P_{B,i} + P_{J,i} + P_{P,i}}{T_w C} - \frac{\partial T_i}{\partial I_L} \dot{I}_L$$

 Thermal conductivity between neighboring TESs through the substrate is not included.



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#### Matrix inversion

• Applying chain rule the matrix  $\frac{\partial T_i}{\partial A_i}$  can be written as:

$$\frac{\partial T_i}{\partial A_j} = \delta_{ij} a_i + u_i v_j$$

• Which can be inverted using the Sherman-Morrison formula:

$$\left(\frac{\partial T_i}{\partial A_j}\right)^{-1} = \frac{\delta_{ij}}{a_i} - \frac{(u_i/a_i)(v_j/a_j)}{1 + \sum_k u_k v_k/a_k}$$
$$a_i = T_w + B_i I_L \frac{2}{R_n} e^{-2A_j}$$
$$B_i = \frac{2}{3} \frac{T_c}{I_c} \frac{1}{\sum_k G_k} \operatorname{sgn} I_i \left(1 - \frac{|I_i|}{I_c}\right)^{-1/3}$$
$$u_i = B_i I_i, \qquad v_j = -\frac{2}{R_n} e^{-2A_j}$$

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# ODEs final form

#### Putting all together

$$\dot{I_L} = \frac{V_{b0} - I_L \left(R_0 + \frac{1}{\sum_k G_k}\right)}{L}$$
$$\dot{A_i} = \frac{D_i}{a_i} - \frac{1}{1 + \sum_k u_k v_k / a_k} \frac{u_i}{a_i} \sum_j \frac{v_j}{a_j} D_j$$

Where

$$D_{i} = \frac{K\left(T_{b}^{5} - T_{i}^{5}\right) + \left(\frac{I_{L}}{\sum_{k} G_{k}}\right)^{2} G_{i} + P_{i}^{P}}{T_{w}C_{n}} + B_{i}G_{i}\frac{V_{b0} - I_{L}\left(R_{0} + 1/\sum_{k} G_{k}\right)}{L}$$

▶ We have n+1 coupled ODEs with no numerical matrix calculations.

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# ODE solver and limitations

- The ODEs are easily solved using CVODE solver in C + +.
- TESSim reports the pulse amplitude, temperature A,  $T_c$ ,  $T_c^{eff}$ , R, G and power.
- ► The ODE solver time step is smaller than the rise time to be able to capture all the changes.
- > The number of TESs in a detector can be huge  $\sim 1000$ , TESSim uses a lumping algorithm
- $\blacktriangleright~$  It lumps TESs into  $\sim 35$  pseudo-TESs that are bigger in size.
- Volumetric quantities (C, K,  $R_n$ ) are scaled accordingly.



# Pulse shape

Impulse response for a single TES

$$\delta I(t) = A \left( E_{ph} \right) \left( e^{-\frac{t}{\tau_{+}}} - e^{-\frac{t}{\tau_{-}}} \right)$$
$$\frac{1}{\tau_{\pm}} = \frac{1}{2\tau_{LR}} + \frac{1}{2\tau_{TES}}$$
$$\pm \frac{1}{2} \sqrt{\left( \frac{1}{\tau_{LR}} - \frac{1}{\tau_{TES}} \right)^2 - 4 \frac{R_0}{L} \frac{\mathcal{L}(2+\beta)}{\tau}}$$
$$\tau_{TES} = \frac{C}{G \left( 1 - \mathscr{L} \right)}, \quad \tau_{LR} = \frac{L}{R_L + (1+\beta)R_0}$$



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## Important TES model parameters

Parameter	Description	Units
fin efficiency	Contribution of proximity effect to volumetric quantities	
	Inductance	Н
$R_b$	Bias resistance	Ω
$R_s$	Shunt resistance	Ω
$R_p$	Parasitic resistance	$\Omega$
$R_n$	Normal-phase resistance of an individual TES	Ω
$I_c$	Critical current of an individual TES	А
$T_c$	TES critical temperature	К
$T_w$	Width of TES superconductivity transition curve	К
$\gamma$	Volumetric specific heat of the TES, $C=f_{sc}\mathcal{V}_{eff}\gamma T_c$	$J/(m^3K^2)$
Σ	Electron-phonon coupling constant, $K = \Sigma \mathcal{V}_{eff}$	$W/(m^3K^5)$

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HVeV run 2



#### HVeV run 4

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## **Electro-Thermal Oscillations**

- In HVeV and HV detectors, it was found that the baseline resolution worsens as the bias point is lowered
- ETO contributes a frequency component to the noise that degrades the resolution
- ► TESSim was used to check that the device in question lives in the ETO region.



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# Conclusion and Future Work

- ► TES coupled ODEs are solved numerically using BDF or Adams method from CVODE
- Sherman-Morrison formula is used to avoid matrix calculations
- TESSim is very successful in producing the correct pulse shape for single TESs and HVeV data
- Future code developments:
  - $\blacktriangleright$  3T model to include the thermal conductivity between the substrate and the heat bath
  - Thermal conductivity between neighboring TESs through the substrate

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